

# **CMS79FT73x User Manual**

Enhanced flash memory 8-bit CMOS microcontrollers

Rev. 1.5.2

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# **1. Product Description**

#### 1.1 Features

- memory
  - Flash: 8Kx16
  - Universal RAM: 256x8
  - Dedicated RAM for built-in touch feature: 400x8
- 8 level stack buffer
- Clean instructions (68 instructions)
- look-up table
- built-in WDT timer
- built-in low voltage detection circuit
- Interrupt source
  - 3 timer interrupt
  - Interrupt for change in electrical level RB port
  - Other peripherals interrupt
- timer
  - 8-bit timer: TIMER0, TIMER2
  - 16-bit timer: TIMER1 which supports external
  - 32.768kHz crystal oscillator as timer clock source
- Capture, compare and PWM mod (CCP)
  - 10-bit PWM precision
  - 2 PWM circuit with configurable and adjustable period and duty cycle
  - Able to configurate between RB4/RB3 or RC6/RC7
- built-in 128-byte EEPROM
  - Can re-write/erase up to 100000 times
- built-in touch button detection mod, no need for external touch capacitor
  - Can pass dynamic/static 10V EMC test
  - Adjustable touch sensitivity
  - All pins can be configurated to touch channel

- Working voltage: 2.6V~5.5V@16MHz
   2.0V~5.5V@8MHz
- Working temperature: -40°C~85°C
- Multiple ways of oscillation
  - Internal RC: design frequency of 8MHz/16MHz
  - External XT: up to 16MHz
- Instructions period (single instruction or double instructions)
- built-in LED driver mod
  - Support up to 11 segments, 8 bits
  - Large driving current up to 150mA at COM port
  - Adjustable current of 2-30mA at SEG port
- built-in LCD1/2 Bias COM driver mod
  - Support up to 8 COM port
  - COM port driver current selection
- built-in MSSP communication mod (SPI/ I<sup>2</sup>C)
  - I<sup>2</sup>C supports master control/slave mode (7-bit addressing)
  - I<sup>2</sup>C slave mode supports broadcast call
  - SPI supports master/slave mode
- built-in 2 USART communication mod
  - Supports synchronous master/slave mode and asynchronous duplex mode
  - serial port 1 can be configurated between RB4/RB3 or RC6/RC7
- High precision 12-bit ADC
  - Built-in high precision 1.2V reference voltage

±1.5% @VDD=2.5V~5.5V T<sub>A</sub>=25°C

±2% @VDD=2.5V~5.5V T<sub>A</sub>=-40°C~85°C

- built-in LVD mod
  - Choice of voltage:
     2.2V/2.4V/2.7V/3.0V/3.3V/3.7V/4.0V/4.3V



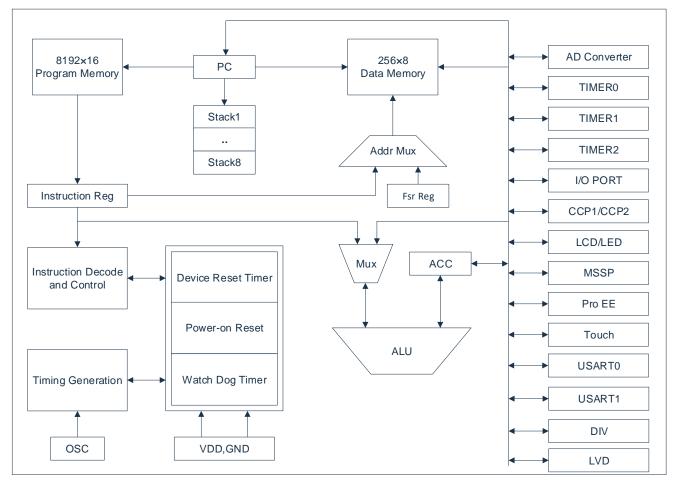
#### Product specification

PRODUCT	ROM	RAM	Pro EE	I/O	LED	LCD	ADC	Touch	USART	PACKAGE
CMS79FT736	8Kx16	256x8	128x8	18	8 <sub>seg</sub> x4 <sub>com</sub>	1/2Biasx4 <sub>com</sub>	12Bitx18	16	2	SOP20
CMS79FT738	8Kx16	256x8	128x8	26	11 <sub>seg</sub> x8 <sub>com</sub>	1/2Biasx8 <sub>com</sub>	12Bitx26	16	2	SOP28

Note: ROM----program memory Pro EE---program EEPROM



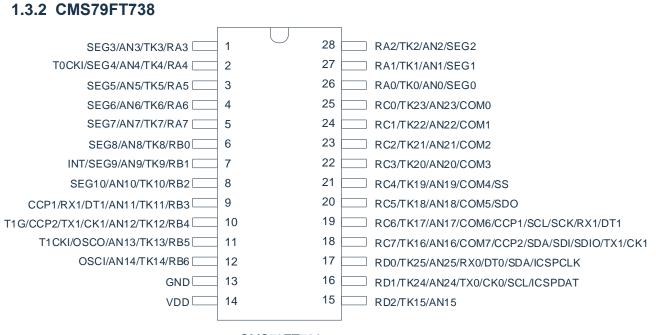
# 1.2 System Structure Diagram







CMS79FT736



CMS79FT738

Note:

1) RB4/RB3 and RC6/RC7's serial port function is congurated by CONFIG

2) RB4/RB3 and RC6/RC7's CCP function is congurated by CONFIG

#### Pin description:

	10.1				
Pin name	IO type	description			
VDD, GND	Р	Voltage input pin and ground			
OSCIN/OSCOUT	Р	Oscillator in/out pin			
RA0-RA7	I/O	Programmable in/ push-pull out pin, with pull-up resistance function			
RB0-RB6	I/O	Programmable in/push-pull out pin, with pull-up resistance, pull-down resistance, electrical level interrupt function			
RC0-RC7	I/O	Programmable in/ push-pull out pin, with pull-up resistance function			
RD0-RD2	I/O	Programmable in/ push-pull out pin, with pull-up resistance function			
ICSPCLK/ICSPDAT	I/O	Program clock/data pin			
TK0-TK25	-	touch button input pin			
AN0-AN23	I	12 位 ADC input pin			
SEG0-SEG10	0	LED driversegment ouput			
COM0-COM7	0	LED/LCD driver common port			
TOCKI	I	TIMER0 external clock input pin			
T1CKI	I	TIMER1 external clock input pin			
T1G	I	TIMER1 gate control input pin			
CCP1	I/O	capture/compare/PWM1			
CCP2	I/O	capture/compare/PWM2			
SCK	I/O	SPI clock input pin			
SDI	I	SPI data input pin			
SDO	0	SPI data output pin			
SS	I	SPI slave choice input pin			
SCL	I/O	l <sup>2</sup> Cclock input/output pin			
SDA	I/O	l <sup>2</sup> Cdata input/output pin			
TX0/CK0	I/O	USART0 asynchronous transmit output/synchronous clock input/output pin			
RX0/DT0	I/O	USART0 asynchronous receive input/synchronous data input/output pin			
TX1/CK1	I/O	USART1 asynchronous transmit output/synchronous clock input/output pin			
RX1/DT1	I/O	USART1 asynchronous receive input/synchronous data input/output pin			





# 1.4 System Configuration Register

System configuration register (CONFIG) is the initial FLASH choice of the MCU. It can only be burned by CMS burner. User cannot visit. It includes the following:

- 1. OSC (choice of oscillation)
  - INTRC Internal RC oscillation
  - ◆ XT External crystal oscillation
- 2. INTRC\_SEL (internal oscillation frequency)
  - INTRC8M F<sub>HSI</sub> choose internal 8MHz RC oscillation
  - INTRC16M F<sub>HSI</sub> choose internal 16MHz RC oscillation
- 3. WDT (watchdog choice)
  - ENABLE Enable watchdog timer
  - DISABLE
     Disable watchdog timer
- 4. PROTECT (encyption)
  - DISABLE Disable FLASH code encyption
  - ENABLE Enable FLASH code encryption, after which the read value from burning the simulator is uncertain.
- 5. LVR\_SEL (low voltage detection selection)
  - 2.0V
  - 2.6V
- 6. ICSPPORT\_SEL (simulation port selection)
  - ICSP ICSPCLK, DAT port keep as similation port, all functions disabled
  - NORMAL ICSPCLK, DAT port as normal port
- 7. USART1\_SEL (TX1/RX1) (USART1 port selection)
  - RC7/RC6 Select RC7 as TX1, RC6 as RX1,
  - RB4/RB3 Select RB4 as TX1, RB3 as RX1
- 8. CCP\_SEL (CCP port selection)
  - RC6/RC7 Select RC6 as CCP1, RC7 as CCP2
  - RB3/RB4 Select RB3 as CCP1, RB4 as CCP2
- 9. IIC\_SEL (IIC port selection)
  - RD1/RD0 Select RD1 as SCL, RD0 as SDA
  - RC6/RC7 Select RC6 as SCL, RC7 as SDA



# 1.5 Online Serial Programming

Can perform serial programming on MCU t the final application circuit. Programming is done through the following:

- Power wire
  - Ground wire
  - Data wire
  - Clock wire

This ensures users to use un-programmed devices to make circuit and only program the MCU just before the product being delivered. Therefore, the lastest version of firmware can be burned into the MCU.

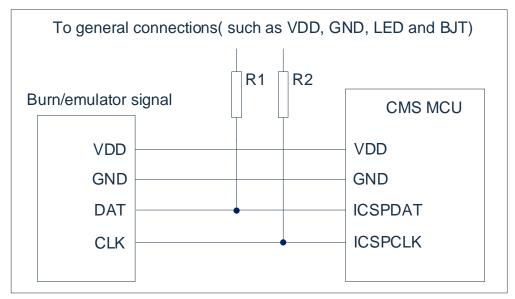


Fig 1-1: Typical connection for online serial programming

In the above figure, R1 and R2 are the electrical isolation devices, normally represented by resistor with the following resistance: R1 $\ge$ 4.7K, R2 $\ge$ 4.7K.

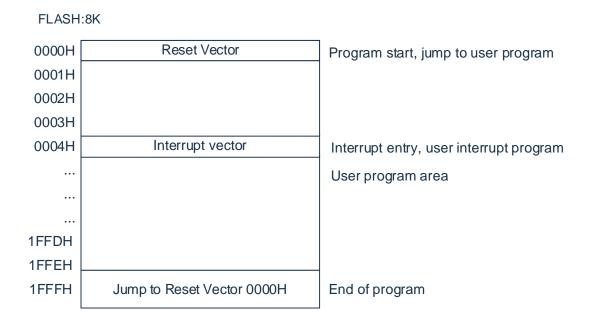


# 2. Central Processing Unit (CPU)

#### 2.1 Memory

#### 2.1.1 Program Memory

CMS79FT73x program memory space



#### 2.1.1.1 Reset Vector (0000H)

MCU has 1-byte long system reset vector (0000H). It has 3 ways to reset:

- Power-on reset
- Watchdog reset
- Low voltage reset (LVR)

When any above reset happens, program will start to execute from 0000H, system registerwill be recovered to default value. PD and TO from STATUS register can determine the which reset is performed from above. The following program illustrayes how to define the reset vector from FLASH.

Example. define re			
	ORG	0000H	; system reset vector
	JP	START	
	ORG	0010H	; start of user program
START:			
			; user program
	END		; program end
L			

Example: define reset vector



#### 2.1.1.2 Interrupt Vector

The address for interrupt vector is 0004H. Once the interrupt responds, the current value for program counter PCwill be saved to stack buffer and jump to 0004H to execute interruptservice program. All interruptwill enter 0004H. User will determine which interrupt to execute according to the bit of register of interrupt flag bit. The following program illustrate how to write interrupt service program.

Examplet denite i	nonapt tootoi,	interrupt program io	placed alter user program
	ORG	0000H	; systemresetvector
	JP	START	
	ORG	0004H	; start of user program
INT_START:			
	CALL	PUSH	; save ACC and STATUS
			; user interrupt program
INT_BACK:			
	CALL	POP	; back to ACC and STATUS
	RETI		; interrupt back
START:			
			; user program
	•••		
	END		; program end

Example: define interrupt vector, interrupt program is placed after user program

Note: MCU does not provide specific unstack and push instructions, so user needs to protect interrupt scene.

#### Example: interrupt-in protection

PUSH:			
L	_D	ACC_BAK,A	; save ACC to ACC_BAK
S	SWAPA	STATUS	; swap half-byte of STATUS
L	D	STATUS_BAK,A	; save to STATUS_BAK
F	RET		; back

#### Example: interrupt-out restore

POP:			
SV	VAPA S	STATUS_BAK ;	swap the half-byte data from STATUS_BAK to ACC
LD	) S	STATUS,A ;	pass the value in ACC to STATUS
SV	VAPR A	CC_BAK ;	swap the half-byte data in ACC_BAK
SV	VAPA A	CC_BAK ;	swap the half-byte data from ACC_BAK to ACC
RE	ΞT	-	back



#### 2.1.1.3 Look-up Table

Any address in FLASH can be use as look-up table.

Related instructions:

- TABLE [R] Pass the lower byte in table to register R, pass higher byte to TABLE\_DATAH.
- TABLEA Pass the lower byte in table to ACC, pass higher byte to TABLE\_DATAH.

related register:

- TABLE\_SPH (110H) Read/write register to indicate higher 5 bits in the table.
- TABLE\_SPL (111H) Read/write register to indicate lower 8 bits in the table.
- TABLE\_DATAH (112H) Read only register to save higer bit information in the table

Note: Write the table address into TABLE\_SPH and TABLE\_SP before using look-up. If main program and interrupt service programboth use look-up tablein structions, the value for TABLE\_SPH in the main program may change due to the look-up instructions from interrupt and hence cause error. Avoid using look-up table instruction in both main program and interrupt service. Dsiable the interrupt before using the look-up table instruction and enable interrupt after the look-up instructions are done.

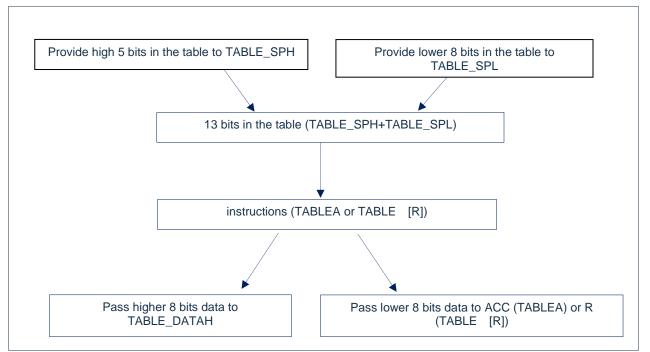


Fig 2-1: Flow chart for table usage



#### The following illustrates how to use the table in the program.

		;continue from user program
LDIA	02H	;lower bits address in the table
LD	TABLE_SPL,A	
LDIA	06H	; higher bits address in the table
LD	TABLE_SPH,A	
TABLE	R01	;table instructions, pass the lower 8 bits (56H) to R01
LD	A,TABLE_DATAH	;pass the higher 8 bits from look-up table (34H) to ACC
LD	R02,A	;pass the value from ACC (34H)to R02
		;user program
ORG	0600H	;start address of table
DW	1234H	;table content at 0600H
DW	2345H	;table content at 0601H
DW	3456H	;table content at 0602H
DW	0000H	;table content at 0603H

# 

#### 2.1.1.4 Jump Table

Jump table can achieve multi-address jump feature. Since the addition of PCL and ACC is the new value of PCL, multi-address jump is then achieved through adding different valueof ACC to PCL. If the value of ACC isn, then PCL+ACC represent the current address plus n. After the execution of the current instructions, the value of PCL will add 1 (refer to the following examples). If PCL+ACC overflows, then PC will not carry. As such, user can achieve multi-address jump through setting different values of ACC.

PCLATH is the PC high bit buffer register. Before operating on PCL, value must be given to PCLATH. Example: correct illustration of multi-address jump

		, ,		
FLASH address				
	LDIA	01H		
	LD	PCLATH,A	;must give value to PCLATH	
0110H:	ADDR	PCL	;ACC+PCL	
0111H:	JP	LOOP1	;ACC=0, jump to LOOP1	
0112H:	JP	LOOP2	;ACC=1, jump to LOOP2	
0113H:	JP	LOOP3	;ACC=2, jump to LOOP3	
0114H:	JP	LOOP4	;ACC=3, jump to LOOP4	
0115H:	JP	LOOP5	;ACC=4, jump to LOOP5	
0116H:	JP	LOOP6	;ACC=5, jump to LOOP6	

#### Example: wrong illustration of multi-address jump

FLASH address			
	CLR	PCLATH	
00FCH:	ADDR	PCL	;ACC+PCL
00FDH:	JP	LOOP1	;ACC=0, jump toLOOP1
00FEH:	JP	LOOP2	;ACC=1, jump toLOOP2
00FFH:	JP	LOOP3	;ACC=2, jump toLOOP3
0100H:	JP	LOOP4	;ACC=3, jump to0000H address
0101H:	JP	LOOP5	;ACC=4, jump to0001H address
0102H:	JP	LOOP6	;ACC=5, jump to0002H address

Note: Since PCI overflow will not carry to the higher bits, the program cannot be placed at the partition of the FLASH space when using PCL to achieve multi-address jump.



#### 2.1.2 Data Memory

#### List of data memory of CMS79FT73x

	address		address		address		address
INDF	00H	INDF	80H	INDF	100H	INDF	180H
TMR0	01H	OPTION_REG	81H	TMR0	101H	OPTION_REG	181H
PCL	02H	PCL	82H	PCL	102H	PCL	182H
STATUS	03H	STATUS	83H	STATUS	103H	STATUS	183H
FSR	04H	FSR	84H	FSR	104H	FSR	184H
PORTA	05H	TRISA	85H	RCSTA1	105H	LVDCON	185H
PORTB	06H	TRISB	86H	PORTB	106H	TRISB	186H
PORTC	07H	TRISC	87H	WPUA	107H	DIVS1	187H
PORTD	08H	TRISD	88H	WPUC	108H	DIVS0	188H
ANSEL0	09H	ANSEL1	89H	ANSEL2	109H	DIVCON	189H
PCLATH	0AH	PCLATH	8AH	PCLATH	10AH	PCLATH	18AH
INTCON	0BH	INTCON	8BH	INTCON	10BH	INTCON	18BH
PIR1	0CH	PIE1	8CH	EEDAT	10CH	DIVE3/DIVQ3	18CH
PIR2	0DH	PIE2	8DH	EEADR	10DH	DIVE2/DIVQ2	18DH
TMR1L	0EH	WPUD	8EH	EEDATH	10EH	DIVE1/DIVQ1	18EH
TMR1H	0FH	OSCCON	8FH	EEADRH	10FH	DIVE0/DIVQ0	18FH
T1CON	10H	WDTCON	90H	TABLE_SPH	110H		190H
TMR2	11H	SSPCON2	91H	TABLE_SPL	111H		191H
T2CON	12H	PR2	92H	TABLE_DATAH	112H		192H
SSPBUF	13H	SSPADD	93H	LEDCON0	113H		193H
SSPCON	14H	SSPSTAT	94H	LEDCON1	114H		194H
CCPR1L	15H	WPUB	95H	LEDADD	115H		195H
CCPR1H	16H	IOCB	96H	LEDDATA	116H		196H
CCP1CON	17H	WPDB	97H	SEGEN2	117H		197H
RCSTA0	18H	SPBRG1	98H	SEGEN1	118H		198H
TXREG0	19H	PWMCON	99H	SEGEN0	119H		199H
RCREG0	1AH	PWM1CYC	9AH	COMEN	11AH		19AH
CCPR2L	1BH	PWM2CYC	9BH	EECON1	11BH		19BH
CCPR2H	1CH	ADRESL	9CH	EECON2	11CH		19CH
CCP2CON	1DH	ADRESH	9DH	TXREG1	11DH		19DH
TXSTA0	1EH	ADCON0	9EH	RCREG1	11EH		19EH
SPBRG0	1FH	ADCON1	9FH	TXSTA1	11FH		19FH
Universal register 96 byte	20H	Universal register 80byte	AOH	Universal register 80byte	120H		1A0H
00 byte	6FH		EFH		16FH		1EFH
	70H	Factmomony	F0H	Factmann	170H	Fast memory	1F0H
		Fast memory spce		Fast memory space		Fast memory space	
	7FH	70H-7FH	FFH	70H-7FH	17FH	70H-7FH	1FFH
BANK0		BANK1		BANK2		BANK3	-

Data memory consists of 512×8 bits. It can be devided into to space: special function register and universal data memory. Most of data memory are able to write/read data, only some data memory are read-only. Special register address is from 00H-1FH, 80-9FH, 100-11FH, 180-19FH.



Address	Name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	Reset value	
00H	INDF	Look-up for t	ook-up for this unit will use FSR, not physical register.								
01H	TMR0	TIMER0 data	a register							xxxxxxx	
02H	PCL	Lower bit of	program counter	r						00000000	
03H	STATUS	IRP	RP1	RP0	то	PD	Z	DC	С	00011xxx	
04H	FSR	memory poir	nters for indirect	addressing of	data memory					xxxxxxx	
05H	PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	xxxxxxx	
06H	PORTB		RB6	RB5	RB4	RB3	RB2	RB1	RB0	-xxxxxx	
07H	PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	xxxxxxx	
08H	PORTD						RD2	RD1	RD0	xxx	
09H	ANSEL0				Analog input co	ntrol register0				00000000	
0AH	PCLATH				Write buffer of	higher 5 bits of	f program cou	nter		00000	
0BH	INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	00000000	
0CH	PIR1		ADIF	<b>RC0IF</b>	TX0IF	SSPIF	CCP1IF	TMR2IF	TMR1IF	-0000000	
0DH	PIR2	LVDIF			EEIF	BCLIF	TX1IF	RC1IF	CCP2IF	0—00000	
0EH	TMR1L	Data register	r of 16-bits TIME	R1 register lov	wer bit					xxxxxxx	
0FH	TMR1H	Data register	Data register of 16-bits TIMER1 register higher bit								
10H	T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	00000000	
11H	TMR2	TIMER2 mod	d register							00000000	
12H	T2CON		TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-0000000	
13H	SSPBUF	Synchronous	s serial port rece	ive buffer /trar	nsmit register					xxxxxxx	
14H	SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	00000000	
15H	CCPR1L	Lower bit of	capture/compare	e/PWM registe	r1					xxxxxxx	
16H	CCPR1H	Higher bit of	capture/compar	e/PWM registe	er1					xxxxxxx	
17H	CCP1CON			DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	000000	
18H	RCSTA0	SPEN0	RX9EN0	SREN0	CREN0	RCIDL0	FERR0	OERR0	RX9D0	00001000	
19H	TXREG0	USART0 trai	nsmit data regist	er						00000000	
1AH	RCREG0	USART0 rec	eive data registe	er						00000000	
1BH	CCPR2L	Lower bit of	capture/compare	e/PWM registe	r2					XXXXXXXX	
1CH	CCPR2H	Higher bit of	capture/compar	e/PWM registe	er2					xxxxxxx	
1DH	CCP2CON			DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	000000	
1EH	TXSTA0	CSRC0	TX9EN0	TXEN0	SYNC0	SCKP0		TRMT0	TX9D0	00000-10	
1FH	SPBRG0	BRG07	BRG06	BRG05	BRG04	BRG03	BRG02	BRG01	BRG00	00000000	



Address	Name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	Reset value
80H	INDF	Look-up for t	this unit will use	FSR, not phys	ical register.	-	-	-		xxxxxxx
81H	OPTION_REG	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0	11111011
82H	PCL	Lower bit of	Lower bit of program counter							
83H	STATUS	IRP	RP1	RP0	ТО	PD	Z	DC	С	00011xxx
84H	FSR	memory poir	nters for indirect	addressing of	data memory					xxxxxxx
85H	TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	11111111
86H	TRISB		TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	-1111111
87H	TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	11111111
88H	TRISD						TRISD2	TRISD1	TRISD0	111
89H	ANSEL1				analog input cor	ntrol register1				00000000
8AH	PCLATH				Write buffer of	higher 5 bits o	f program cou	nter		00000
8BH	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	00000000
8CH	PIE1		ADIE	<b>RC0IE</b>	TX0IE	SSPIE	CCP1IE	TMR2IE	TMR1IE	-0000000
8DH	PIE2	LVDIE			EEIE	BCLIE	TX1IE	RC1IE	CCP2IE	0—00000
8EH	WPUD						WPUD2	WPUD1	WPUD0	000
8FH	OSCCON		IRCF2	IRCF1	IRCF0				SCS	-1100
90H	WDTCON								SWDTEN	0
91H	SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	01000000
92H	PR2	TIMER2 per	iod register							11111111
93H	SSPADD	Synchronous	s serial port (I <sup>2</sup> C	mode) addres	s register					00000000
93H	SSPMSK	MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0	11111111
94H	SSPSTAT		CKE	D/A	Р	S	R/W	UA	BF	-0000000
95H	WPUB		WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0	-0000000
96H	IOCB		IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0	-0000000
97H	WPDB		WPDB6	WPDB5	WPDB4	WPDB3	WPDB2	WPDB1	WPDB0	-0000000
98H	SPBRG1	BRG17	BRG16	BRG15	BRG14	BRG13	BRG12	BRG11	BRG10	00000000
99H	PWMCON		CYC2EN	CK	2[1:0]		CYC1EN	CK1	[1:0]	-000-000
9AH	PWM1CYC	PWM1 perio	d data register							11111111
9BH	PWM2CYC	PWM2 perio	d data register							11111111
9CH	ADRESL	Lower bit of	A/D result regist	er						XXXXXXXX
9DH	ADRESH	Higher bit of	A/D result regist	ter						xxxxxxx
9EH	ADCON0	ADCS1	ADCS0	CHS3	CHS2	CHS1	CHS0	GO/ DONE	ADON	00000000
9FH	ADCON1	ADFM	CHS4							00



address	name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	Reset value	
100H	INDF	Look-up for this	Look-up for this unit will use FSR , not physical register.								
101H	TMR0	TIMER0 mod re	TIMER0 mod register								
102H	PCL	Lower bit of pro	gram counter (PC	C)						00000000	
103H	STATUS	IRP	RP1	RP0	ТО	PD	Z	DC	С	00011xxx	
104H	FSR	memory pointer	s for indirect add	ressing of data	a memory					xxxxxxx	
105H	RCSTA1	SPEN1	RX9EN1	SREN1	CREN1	RCIDL1	FERR1	OERR1	RX9D1	00001000	
106H	PORTB		RB6	RB5	RB4	RB3	RB2	RB1	RB0	-xxxxxx	
107H	WPUA	WPUA7	WPUA6	WPUA5	WPUA4	WPUA3	WPUA2	WPUA1	WPUA0	00000000	
108H	WPUC	WPUC7	WPUC6	WPUC5	WPUC4	WPUC3	WPUC2	WPUC1	WPUC0	00000000	
109H	ANSEL2			an	alog input cont	rol register2				0	
10AH	PCLATH				Write buffer	of higher 5 bits	of program co	ounter		00000	
10BH	INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF	00000000	
10CH	EEDAT	EEDAT7	EEDAT6	EEDAT5	EEDAT4	EEDAT3	EEDAT2	EEDAT1	EEDAT0	xxxxxxx	
10DH	EEADR	EEADR7	EEADR6	EEADR5	EEADR4	EEADR3	EEADR2	EEADR1	EEADR0	00000000	
10EH	EEDATH	EEDATH7	EEDATH6	EEDATH5	EEDATH4	EEDATH3	EEDATH2	EEDATH1	EEDATH0	xxxxxxx	
10FH	EEADRH				EEADRH4	EEADRH3	EEADRH2	EEADRH1	EEADRH0	00000	
110H	TABLE_SPH				Pointers for I	higher 5 bits of	the table			xxxxx	
111H	TABLE_SPL	Pointers for low	bits in table							xxxxxxx	
112H	TABLE_DAT AH	Data for high bit	ts in table							xxxxxxx	
113H	LEDCON0	LCDEN	LEDEN	COMSEL1	COMSEL0	LEDCLK3	LEDCLK2	LEDCLK1	LEDCLK0	00000000	
114H	LEDCON1		LEDF	SEGOUT1	SEGOUT0					-000	
115H	LEDADD	LEDCS		COMSEL2			LEDADD[4:0]			0-000000	
116H	LEDDATA	LED data regist	er							xxxxxxx	
117H	SEGEN2	current control r	egister2 for SEG	port driver						0000	
118H	SEGEN1	current control r	egister1 for SEG	port driver						000	
119H	SEGEN0	current control r	egister0 for SEG	port driver						00000000	
11AH	COMEN	control register	for COM port							00000000	
11BH	EECON1	EEPGD				WRERR	WREN	WR	RD	0x000	
11CH	EECON2	EEPROM contr	ol register2 (not p	ohysical registe	er)						
11DH	TXREG1	USART1 transn	nit data register							00000000	
11EH	RCREG1	USART1 receiv	e data register							00000000	
11FH	TXSTA1	CSRC1	TX9EN1	TXEN1	SYNC1	SCKP1		TRMT1	TX9D1	00000-10	



Address	Name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	Reset value	
180H	INDF	Look-up for th	Look-up for this unit will use FSR , not physical register.								
181H	OPTION_RE	RBPU	RBPU INTEDG TOCS TOSE PSA PS2 PS1 PS0							11111011	
	G										
182H	PCL	Lower bit of pr	ower bit of program counter (PC)								
183H	STATUS	IRP	RP1	RP0	то	PD	Z	DC	С	00011xxx	
184H	FSR	memory point	ers for indirect a	ddressing of a	data memory					xxxxxxx	
185H	LVDCON	LVD_RES		-		LVD_SEL2	LVD_SEL1	LVD_SEL0	LVDEN	000000	
186H	TRISB		TRISB6	TRISB5	TRISB4	TRISB73	TRISB2	TRISB1	TRISB0	-1111111	
187H	DIVS1	Higher 8 bits	for divisor							00000000	
188H	DIVS0	Lower 8 bits f	or divisor							00000000	
189H	DIVCON	DIVEN	CAL_END						DIV_CLK	010	
18AH	PCLATH				Write buffer of h	igher 5 bits of p	program count	er		00000	
18BH	INTCON	GIE	PEIE	TOIE	INTE	RBIE	TOIF	INTF	RBIF	00000000	
18CH	DIVE3/DIVQ3	dividend or qu	uotient BIT[31:24	4]						00000000	
18DH	DIVE2/DIVQ2	dividend or qu	uotient BIT[23:16	6]						00000000	
18EH	DIVE1/DIVQ1	dividend or qu	uotient BIT15:8]							00000000	
18FH	DIVE0/DIVQ0	dividend or qu	uotient BIT[7:0]							00000000	



#### 2.2 Addressing Mode

#### 2.2.1 Direct Addressing

Operate on RAM through accumulator (ACC)

Example: pass the value in ACC to 30H register

LD

30H,A

Example: pass the value in 30H register to ACC

LD	A,30H	

#### 2.2.2 Immediate Addressing

Pass the immediate value to accumulator (ACC).

Example: pass immediate value 12H to ACC

LDIA 12H

#### 2.2.3 Indirect Addressing

Data memory can be direct or indirect addressing. Direct addressing can be achieved through INDF register, INDF is not physical register. When load/save value in INDF, address is the value in FSR register (lower 8 bits) and IRP bit in STATUS register (9<sup>th</sup> bit), and point to the register of this address. Therefore, after setting the FSR register and the IRP bit of STATUS register, INDF register can be regarded as purpose register. Read INDF (FSR=0) indirectly will produce 00H. Write INDF register indirectly will casue an empty action. The following example shows how indirect addressing works.

Example: application	Example: application of FSR and INDF								
LI	DIA	30H							
LI	D	FSR,A	;Points to 30H for indirect addressing						
С	LRB	STATUS, IRP	;clear the 9 <sup>th</sup> bit of pointer						
С	CLR	INDF	;clear INDF, which mean clear the 30H address RAM tha FSR points to						

Example: clear RAM (20H-7FH) for indirect addressing:

	LDIA	1FH	
	LD	FSR,A	;Points to 1FH for indirect addressing
	CLRB	STATUS, IRP	
LOOP:			
	INCR	FSR	;address add 1, initial address is 30H
	CLR	INDF	;clear the address where FSR points to
	LDIA	7FH	
	SUBA	FSR	
	SNZB	STATUS,C	;clear until the address of FSR is 7FH
	JP	LOOP	



### 2.3 Stack

Stack buffer of the chip has 8 levels. Stack buffer is not part of data memorynor program memory. It cannot be written nor read. Operation on stack buffer is through stack pointers, which also cannot be written nor read. After system resets, SP points to the top of the stack. When sub-program happens or interrupts happens, value in program counter (PC)will be transferred to stack buffer. When return from interrupt or return from sub-program, value is transferred back to PC. The following digram illustrates its working principle.

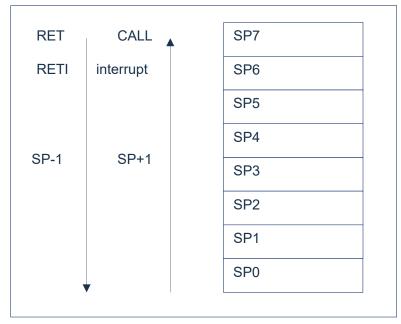


Fig 2-2: stack buffer working principle

Stack bufferwill follow one principle: 'first in last out'

Note: stackbuffer has only 8 levels, if the stack is full and interrupt happens which can not be screened out, then only the indication bit of the interrupt will be noted down. The response for the interrupt will be suppressed until the pointer of stack starts to decrease. This feature can prevent overflow of the stack caused by interrupt. Similarly, when stack is full and sub-program happens, then stack will overflow and the contents which enter the stack first will be lost, only the last 8 return address will be saved.



# 2.4 Accumulator (ACC)

#### 2.4.1 General

ALU is the 8-bit arithmetic-logic unit. All math and logic related calculations in MCU are done by ALU. It can perform addition, subtraction, shift and logical calculation on data; ALU can also control STATUS to represent the status of the product of the calculation.

ACC register is an 8-bit register to store the product of calculation of ALU. It does not belong to data memory. It is in CPU and used by ALU during calculation. Hence it cannot be addressed. It can only be used through the instructions provided.

#### 2.4.2 ACC Applications

Example: use ACC for data transfer

LD	A,R01	;pass the value in register R01 to ACC	
LD	R02,A	;pass the value in ACC to register R02	

#### Example: use ACC for immediate addressing

LDIA	30H	;load the ACC as 30H
ANDIA	30H	;run 'AND' between value in ACC and immediate number 30H,save the result in ACC
XORIA	30H	; run 'XOR' between value in ACC and immediate number 30H,save the result in ACC

Example: use ACC as the first operand of the double operand instructions

HSUBA R0	;ACC-R01, save the result in ACC
HSUBR R0	;ACC-R01, save the result in R01

#### Example: use ACC as the second operand of the double operand instructions

SUBA	R01	;R01-ACC, save the result in ACC
SUBR	R01	; R01-ACC, save the result in R01



## 2.5 Program Status Register (STATUS)

STATUS register includes:

- status of ALU.
- Reset status.
- Selection bit of Data memory (GPR and SFR)

Just like other registers, STATUS register can be the target register of any other instruction. If A instructions that affects Z, DC or C bit that use STATUS as target register, then it cannot write on these 3 status bits. These bits are cleared or set to 1 according to device logic. TO and PD bit also cabbot be written. Hence the instructions which use STATUS as target instruction may not result in what is predicted.

For example, CLRSTATUS will clear higher 3 bits and set the Z bit to 1. Hence the value of STATUS will be 000u u1uu (u will not change.). Hence, it is recommended to only use CLRB, SETB, SWAPA and SWAPR instructions to change STATUS register because these will not affect any status bits.

	-								
03H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	
Read/write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
Reset value	0	0	0	1	1	Х	Х	Х	
Bit7	IRP	: Selection	bit of registe	er memory (fo	or indirect ad	dressing)			
	1=	= Bank2 an	d Bank3 (10	0h-1FFh);		0,			
	0=	= Bank0 an	d Bank1 (00	h-FFh).					
Bit6~Bit5	RP [1:0]	: Selection	bit of memo	ry;					
	00	: Select Ba	nk 0;						
	01	: Select Ba	nk 1;						
	10	: Select Ba	nk 2;						
	11	: Select Ba	nk 3.						
Bit4	TO	: Time out	Time out bit;						
	1=	= Power on	Power on or CLRWDT instructions or STOP instructions;						
	0=	<ul> <li>WDT time</li> </ul>	WDT time out.						
Bit3	PD	: Power do	Power down;						
	1=		Power on or CLRWDT instructions;						
	0=								
Bit2	Z		ult inzero;						
	1=								
	0=								
Bit1	DC								
	1=		• • • •	•			Lower 4 bits in		
D:10	0=		• • • •	ns to higher	Dits or Dorror	nappens in	Lower 4 bits in	the result.	
Bit0	C	,		ot the bighter	thitorna ha	may honres	o.		
	1=			•		rrow happen			
	0=		When no carry happens at the highest bit or borrow happens						

program status register STATUS (03H)



TO and PD bit can reflect the reason for reset of chip. The following is the events which affects the TO and PD and the status of TO nad PD after these events.

events	ТО	PD
Power on	1	1
WDT overflow	0	Х
STOP instructions	1	0
CLRWDT instructions	1	1
sleep	1	0

Events which affect TO/PD

PD	Reset reason
0	WDT overflow awaken MCU
1	WDT overflow non-sleep status
1	Power on
	PD 0 1 1

TO/PD status after reset



# 2.6 **Pre-scaler (OPTION\_REG)**

OPTION\_REG registercan be read or written. Each control bit for configuration is as follow:

- TIMER0/WDT pre-scaler
- TIMER0
- PORTB pull up resistance control

#### pre-scaler OPTION\_REG (81H)

81H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
OPTION_REG	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0
Read/write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	1	1	1	1	1	0	1	1

Bit7	RBPU:	PORTB pull up enable bit							
	1=	Forbi	Forbidden PORTB pull up.						
	0=	Enab	Enable PORTB pull up according to latched value of ports.						
Bit6	INTEDG:	Edge	selectio	n bit for	triggering interrupt				
	1=	INT p	in rising	edge trig	ggered interrupt				
	0=	INT p	in falling	g edge tri	ggered interrupt				
Bit5	T0CS:	Selec	tion bit	for TIME	R0 clock source.				
	0=	Intern	al instru	uctions p	eriod clock (Fsys/4).				
	1=	transi	tion edg	je on TOO	CKI pin				
Bit4	T0SE:	Edge	selectio	n bit for	TIMER0 clock source				
	0=	Increa	ase whe	n T0CKI	pin signal transite from low	/ to high			
	1=	Increa	ase whe	n T0CKI	pin signal transite from hig	h to low			
Bit3	PSA:	pre-so	caler all	ocation					
	0=	pre-so	pre-scaler allocates to TIMER0 mod						
	1=	pre-so	caler all	ocates to	WDT				
Bit2~Bit0	PS2~PS0:	config	configuration bit for pre-allocation parameters.						
		PS2	PS1	PS0	TMR0 frequency ratio	WDTfrequency ratio			
		0	0	0	1:2	1:1			
		0	0	1	1:4	1:2			
		0	1	0	1:8	1:4			
		0	1	1	1:16	1:8			
		1	0	0	1:32	1:16			
		1	0	1	1:64	1:32			
		1	1	0	1:128	1:64			
		1	1	1	1:256	1:128			

Pre-scaler register is a 8-bit counter. When surveil on register WDT, it is a post scaler; when it is used as timer or counter, it is called pre-scaler. There is only 1 physical scaler and can only be used for WDT or TIMER0, but not at the same time. This means that if it is used for TIMER0, the WDT cannot use pre-scaler and vice versa.

When used for WDT, CLRWDT instructions will clear pre-scaler and WDT timer When used for TIMER0, all instruction related to writing TIMER0 (such as : CLR TMR0, SETB TMR0,1 .etc )will clear pre-scaler.



Whether TIMER0 or WDT uses pre-scaler is full controlled by software. This can be changed dynamically. To avoid unintended chip reset, when switch from TIMER0 to WDT, the following instructions should be executed.

CLRB	INTCON,GIE	; Disable enable bit for interrupt to avoid entering interrupt during the following time series
LDIA	B'00000111'	
ORR	OPTION_REG,A	; set pre-scaler as its max value
CLR	TMR0	; clear TMR0
SETB	OPTION_REG,PSA	; set pre-scaler to allocate to WDT
CLRWDT		; clear WDT
LDIA	B'xxxx1xxx'	; set new pre-scaler
LD	OPTION_REG,A	
CLRWDT		; clear WDT
SETB	INTCON,GIE	; when interrupt is needed, enable bis is turned on here

When switch from WDT to TIMER0 mod, the following instructions should be executed.

CLRWDT		;clear WDT
LDIA	B'00xx0xxx'	;set new pre-scaler
LD	OPTION_REG,A	

Note: in order for TIMER0 to have 1:1 pre-scalling, pre-scaller can be allocated to WDT through PSA position 1 of selection register.

# 2.7 Program Counter (PC)

program counter (PC)controls the instruction sequence in programe memory FLASH, it can addressin the whole range of FLASH. After obtaining instruction code, PC will increase by 1 and point to the address of the next instruction code. When executing jump, passing value to PCL, sub-program, initializing reset, interrupt, interrupt return, sub-program return and other actions, PC will load the address which is related to the instruction, rather than the address of the next instruction.

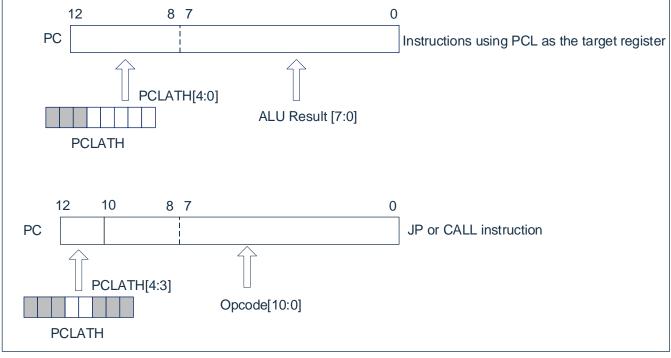
When encountering condition jump instructions and the condition is met, the instruction read during the current instruction will be discarded and an empty instruction period will be inserted. After this, the correct instruction can be obtained. If not, the next instruction will follow the order.

Program counter (PC) is 13 Bit, user can access lower 8 bits through PCL (02H). The higher 5 bits cannot be accessed. It can hold address for 8K×16Bit program. Passing a value to PCL will cause a short jump which range until the 256 addressof the current page.

Note: When using PCL for short jump, it is needed to pass some value to PCLATH

reset	PC=0000;
interrupt	PC=0004 (original PC+1will be add to stack automatically);
CALL	PC=program defined address (original PC+1will be add to stack automatically);
RET、RETI、RET i	PC=value coming out from stack;
Operating on PCL	PC[12:8] unchange, PC[7:0]=user defined value;
JP	PC=program defined value;
Other instructions	PC=PC+1;

The following are the value of PC under special conditions.



The following example illustrate the precautions when using JP or CALL



ORG	00H		
			Target address LABEL1 located at 300H address, current value of PCLATH is 00H under the same 2K
	JP	LABEL1	range. Hence there is no need to change PCLATH before executing JP instructions.
ORG	 300H		
LABEL1:			
	LDIA LD	08H PCLATH,A	Target address LABEL2 located as 900H address, current value of PCLATH is 00H, under the different 2K
	JP	LABEL2	range. Hence it needed to change PCLATH before executing JP instructions.
ORG LABLE4:	 7FEH		
LADEL4.	NOP		;7FEH
	NOP		;7FFH
	NOP		;800H
	LDIA LD	08H PCLATH,A	Target address LABEL5 located as 880H address,
	LD	FULATH,A	current value of PCLATH is 00H (program is executed normally, when PC change from 7FFH to 800H,
	JP	LABLE5	PCLATH will not change), under the different 2K range.
			Hence it needed to change PCLATH before executing JP instructions.
ORG	 880H		
LABLE5:			
	NOP RET		
ORG LABLE2:	 900H		
LADLLZ.	NOP		
			Target address LABEL3 located at E00H address, current value of PCLATH is 08H under the same 2K
	CALL	LABLE3	range. Hence there is no need to change PCLATH before executing CALL instructions.
	LDIA	00H	Target address LABEL4 located as 7FE address, current
	LD	PCLATH,A	value of PCLATH is 08H, under the different 2K range.
	CALL	LABLE4	Hence it needed to change PCLATH before executing CALL instructions.
	NOP		
ORG	 0E00H		
LABLE3:	SEO011		
	NOP		
	RET		



## 2.8 Watchdog Timer (WDT)

Watchdog timer is a self-oscillated RC oscillation timer. There is no need for any external devices. Even the main clock of the chip stops working, WDT can still function/ WDT overflow will cause reset.

#### 2.8.1 WDT Period

WDT and TIMER0 share 8-bit pre-scaler. After all reset, default overflow period fo WDT is 128ms. The way to calculate WDT overflow is 16ms\*pre-scalling parameter. If WDT period needs to be changed, you can configure OPTION\_REG register. The overflow period is affected by environmental temperature, voltage of the power source and other parameter.

"CLRWDT" and "STOP" instructions will clear counting value inside the WDT timer and pre-scaler (when pre-scaler is allocated to WDT). WDT generally is used to prevent the system and MCU program from being out of control. Under normal condition, WDT shouldbe cleared by "CLRWDT" instructions before overflow to prevent reset being generated. If program is out of control for some reason such that "CLRWDT" instructions is not able to execute before overflow, WDT overflow will then generate reset to make sure the system restarts. If reset is generated by WDT overflow, then 'TO'bit of STATUS will be cleared to 0. User can judge whether the reset is caused by WDT overflow according to this.

Note:

- 1) If WDT is used, 'CLRWDT' instructions must be placed somewhere is the program to make sure it is cleared before WDT overflow. If not, chip will keep resetting and the system cannot function normally.
- 2) It is not allowed to clear WDT during interruptso that the main program 'run away' can be detected.
- 3) There should be 1 clear WDT in the main program. Try not to clear WDT inside the sub program, so that the protection feature of watchdog timer can be used largely.
- Different chips has slightly different overflow time in watchdog timer. When setting clear time for WDT, try to leave extra time for WDT overflow time so that unnecessary WDT reset can be avoided.

#### 2.8.2 Watchdog Timer Control Register WDTCON

#### WDTCON (90H)

90H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
WDTCON								SWDTEN
R/W								R/W
Reset value								0

Bit7~Bit1	Not u

Bit0

Not used, read as 0

SWDTEN: Software enable or disable watchdog timer bit

1= Enable WDT

0= Disable WDT (reset value)

Note: if WDT configuration bit in CONFIG equals 1, then WDT is always enabled and is unrelated to the status of control bit of SWDTEN. if WDT configuration bit in CONFIG equals 0, then it is able to disable WDT using the control bit of SWDTEN.



# 3. System Clock

## 3.1 General

When clock signalis input from OSCIN pin (or generated by internal oscillation), 4 non-overlapping orthogonal clock signals called Q1, Q2, Q3, Q4 are produced. Inside IC, each Q1 makes program counter (PC)increase 1, Q4 obtain this instruction from program memory unit and lock it inside instructions register. Compile and execute the instruction obtained between next Q1 and Q4, which means that 4 clock period for 1 executed instruction. The following diagram illustrate the time series of clock and execution of instruction period.

1 instruction period contains 4 Q period. The execution of instructions has pipeline structure. Obtaining instructions only require 1 instruction period, compiling and executing use another instruction period. Since pipeline structure is used, the effective executing time for every instructionis 1 instruction period. If 1 instruction casue PC addressto change (such as JP), then the pre-loaded instruction code is useless and 2 instruction period is needed to complete this instruction. This is why every operation on PC consumes 2 clock period.

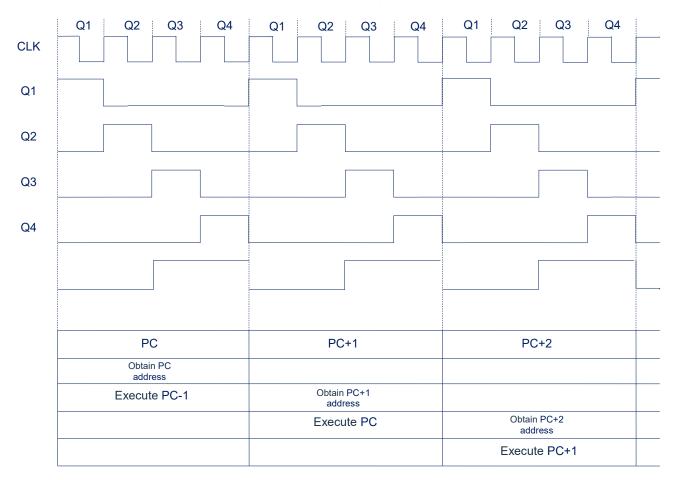


Fig 3-1: time series for clock and instruction period



#### Following is the relationship between working frequency of system and the speed of instructions:

System frequency (F <sub>SYS</sub> )	Double instruction period	Single instruction period
1MHz	8µs	4µs
2MHz	4µs	2µs
4MHz	2µs	1µs
8MHz	1µs	500ns



## 3.2 System Oscillator

Chip has 2 ways of oscillation, internal RC oscillation and external XT oscillation.

#### 3.2.1 Internal RC Oscillation

Default oscillation is internal RC oscillation. Its frequency is 8MHz or 16MHz, which is set by OSCCON register.

When select internalRC as oscillator 时, OSCIN and OSCOUT can be used as normal I/O ports.

#### 3.2.2 External XT Oscillation

Select OSC to be XT during burning process, chip works under external XT oscillation mode.此时 internal RC oscillation stops working and OSCIN and OSCOUT are oscillation ports.

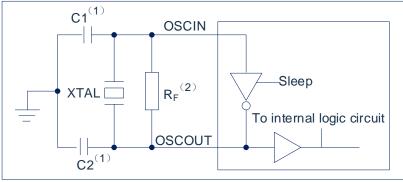


Fig 3-2: Typical XT oscillation

Recommend parameters:

Туре	Frequency	Recommend R <sub>F</sub>	Recommend C1 ~ C2
XT	4MHz	1ΜΩ	10Pf~47pF
XT	8MHz	1ΜΩ	10Pf~47pF
XT	16MHz	1ΜΩ	10Pf~47pF

## 3.3 Reset Time

Reset Time is the time for chip to change from reset to stable oscillation. The value is about 16ms.

Note: Reset time exists for both power on reset and other resets.



## 3.4 Oscillator Control Register

Oscillator control (OSCCON)register controls the system clock and frequency selection. Oscillator tune register OSCTUNE can tune the frequency of internal oscillationin the software.

#### OSCCON (8FH)

8FH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
OSCCON		IRCF2	IRCF1	IRCF0				SCS
R/W		R/W	R/W	R/W				R/W
reset value		1	1	0				0

Bit7	Not used, read 0	
Bit6~Bit4	IRCF<2:0>:	Slection bit for frequency division of Internal oscillator
	111=	Fsys = Fhsi /1
	110=	F <sub>SYS</sub> = F <sub>HSI</sub> /2 (default)
	101=	Fsys = Fhsi /4
	100=	Fsys = Fhsi /8
	011=	Fsys = Fhsi /16
	010=	Fsys = Fhsi /32
	001=	Fsys = Fhsi /64
	000=	Fsys = 32kHz (LFINTOSC).
Bit3~Bit1	Not used	
Bit0	SCS:	Selection bit for system clock
	1=	Internal oscillator as system clock
	0=	clock source defined by CONFIG

Note: F<sub>HSI</sub> as internal oscillator has frequency of 8MHz/16MHz; F<sub>SYS</sub> is the working frequency of the system.



## 3.5 Clock Block Diagram

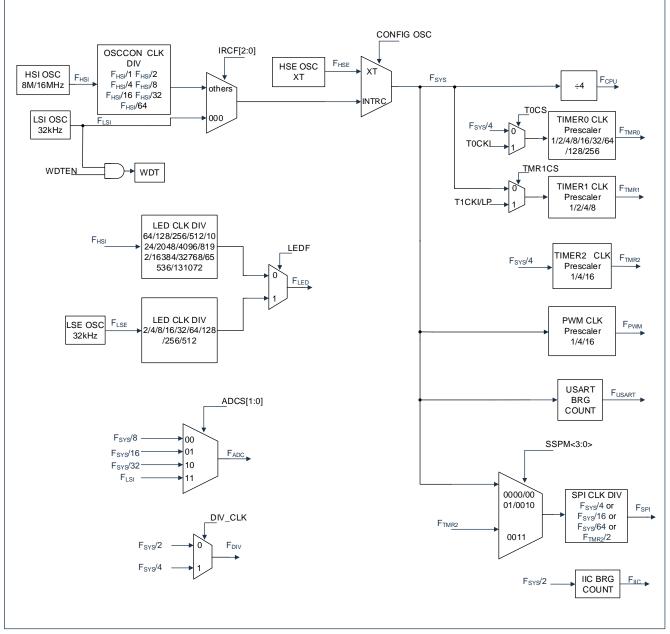


Fig 3-2: clock block diagram



# 4. Reset

Chip has 3 ways of reset:

- Power on reset;
- Low voltage reset;
- Watchdog overflow reset under normal working condition.

When any reset happens, all system registers reset to defaultcondition, program stops executing and PC is cleared. When finishing resetting, program executes from reset vector 0000H. TO nad PD bit from STATUS can provide information for system reset (see STATUS). User can control the route of the program according to the status of PD and TO.

Any reset requires certain respond time. System provides completed reset procedures to make sure the reset is processed normally.

## 4.1 **Power on Reset**

Power on reset is highly related to LVR. Power on process of the sysems should be increasing, after passing some time, the normal electrical level is then reached. The normal time series for power on is as follows:

- Power on: system detects the voltage of the source to increase and wait for it to stabilize;
- System initialization: all system registerset to initial value;
- Oscillator starts working: oscillator starts to provide system clock;
- Executing program: power on process ends, program starts to be executed.



## 4.2 Power off Reset

#### 4.2.1 General

Power off reset is used for voltage drop caused by external factors (such as interference or change in external load). Voltage drop may enter system dead zone. System dead zone means power source cannot satisfy the minimal working voltage of the system.

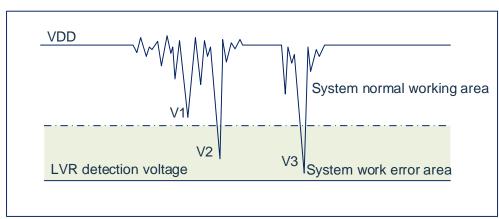


Fig 4-1: power off reset

The above is a typical power off reset case. VDD is under serious interference and the voltage is dropped to a low value. The system works normally above the dotted line and the system enters an unknown situation below the dotted line. This zone is called dead zone. When VDD drops to V1, system still works normally. When VDD drops to V2 and V3, system enters the dead zone and may cause error.

System will enter the dead zone under the following situation:

- DC:
  - Battery provides the power under DC. When the voltage of the battery is too low or the driver of MCU is over-loaded, system voltage may drop and enter the dead zone. Here, power source will not drop further to LVD detection voltage, hence system remains staying at the dead zone.
- AC:
  - When the system is powered by AC, voltage of DC is affected by the noise in AC source. When
    external over-loaded, such as driving motor, this action will also interfer the DC source. VDD
    drops below the minimal working voltage due to interference, system may enter untableworking
    condition.
  - Under AC condition, system power on/off take long time. Power on protection can ensure the system to power on normally, but power off situation is similar to DC case, when AC source is off, VDD drops and may enter dead zone easily.

As illustrated in the above diagram, the normal working voltage is higher than the system reset volateg, at the same time, reset voltage is decided by LVR. When the execution speed increases, the minimal working voltage should increase. However, the system reset volatage is fixed, hence there is a dead zone between the minimal working voltage and system reset voltage.





#### 4.2.2 Improvements for Power off Reset

Suggestions to improve the power off reset:

- Choose higher LVR voltage;
- Turn on watchdog timer;
- Lower working frequency of the system;
- Increase the gradient of the voltage drop.

#### Watchdog timer

Watchdog timer is used to make sure the program is run normally. When system enter the dead zone or error happens, watchdog timer overflow and system reset.

#### Lower the working speed of the system

Higher the working frequency, higher the minimal working voltage system. Dead zone is increase when system works at higher frequency. Therefore, lower the working speed can lower the minimal working voltage and then decrease the probability of entering the dead zone.

#### Increase the gradient of the voltage drop

This method is used under AC. Voltage drops slowly under AC and cause the system to stay longer at the dead zone. If the system is power on at this moment, error may happen. It is then suggested to insert a resistor between power source and ground to ensure the MCU pass the dead zone and enter the reset zone faster.



## 4.3 Watchdog Reset

Watchdog reset is a protection for the system. Under normal condition, program clear the the watchdog timer. If error happends and system is under unknown status, watchdog timer overflow and then system reset. After watchdofg reset, system restarts and enter normal working condition.

Time series for watchdog reset:

- Watchdog timer status: system detects watchdog timer. If overflow, then system reset;
- Initialization: all system registerset to default;
- oscillator starts working: oscillator starts to provide system clock;
- program: reset ends, program starts to be executed.

For applications of watchdog timer, see chapters at 2.8



# 5. Sleep Mode

## 5.1 Enter Sleep Mode

System can enter sleep modewhen executing STOP instructions. If WDT enabled, then:

- WDT is cleared and continue to run.
- PD bit in STATUS register is cleared.
- TO bit set to 1.
- Turn off oscillator driver device.
- ◆ I/O port keep at the status before STOP (driver is high level, low lower, or high impedence).

Under sleep mode, to avoid current consumption, all I/O pin should keep at VDD or GND to make sure no extermal circuit is consuming the current from I/O pin. To avoid input pin, suspend and invoke current, high impedence I/O should be pulled to high or low level externally. Internal pull up resistance should also be considered.

## 5.2 Awaken from Sleep Mode

Awaken through any of the following events:

- 1. Watchdog timer awake (WDT force enable)
- 2. PORTB electrical level interrupt or peripherals interrupt

The above 2 events are regards as the extention of the execution of the program. TO and PD bit in STATUS register are used to find the reason for reset. PD is set to 1 when power on and clear to 0 when STOP instruction is executing.TO is cleared when WDT awaken happens.

When executes STOP instructions, next instruction (PC+1) is withdrawed frist. If it is intended to awaken the system using interrupt, the corresponding enable bit should be set to 1 for the interrupt. Awaken is not related to GIE bit. If GIE is cleared, system will continue to execute the instruction after STOP instruction, and then junp to interrupt address (0004h) to execute. To avoid instruction after STOP instruction being executed, user should put one NOP instruction after STOP instruction. When system is awaken from sleep mode, WDT will be cleared to 0 and has nothing to do with the reason for awakening.

## 5.3 Interrupt Awakening

When forbidden overall interrupt (GIE clear), and there exist 1 interrupt source with its interrupt enable bit and indication bit set to 1, one event from the following will happen:

- If interrupt happens before STOP instructions, then STOP instruction is executed as NOP instructions. Hence, WDT and its pre-scaler and post-scaler will not be cleared, and TO bit will not be set to 1, PD will not be cleared to 0.
- If interrupt happens during or after STOP instruction, then system is awaken from sleep mode. STOP will be executed before system being fully awaken. Hence, WDT and its pre-scaler, post-scaler will be cleared to, TO bit set to 1 and PD bit cleared to 0. Even if the indication bit is 0 before executing the STOP instruction, it can be set to 1 before STOP instruction is finished. To check whether STOP is executed, PD bit can be checked, if is 1, then STOP instruction is executed as NOP. Before executing STOP instruction, 1 CLRWDT instruction must be executed to make sure WDT is cleared.



## 5.4 Sleep Mode Application

Before system enters sleepmode, if user wants small sllep current, please check all I/O status. If suspended I/O port is required by user, set all suspended ports as output to make sure each I/O has a fixed status and avoid increasing sleep current when I/O is input; turn off AD and other peripherals mod; WDT functions can be turned off to decrease the sleep current.

I	eddies for entering sleep		
SLEEP_MODE:			
	CLR	INTCON	; disable interrupt
	LDIA	B'00000000'	
	LD	TRISA,A	
	LD	TRISB,A	;all I/O set as output
	LD	TRISC,A	
	LD	TRISE,A	
			;turn off other functions
	LDIA	0A5H	
	LD	SP_FLAG,A	;set sleep status memory register
	CLRWDT		;clear WDT
	STOP		;execute STOP instruction

example: procedures for entering sleep mode

## 5.5 Sleep Mode Awaken Time

When MCU is awaken from sleep mode, oscillation reset time is needed. The specific relationship is shown in the table below:

System main clock source	System clock frequency (IRCF<2:0>)	Twait
	Fsys=Fhsi	Twait=1032*1/F <sub>HSI</sub> +16*1/F <sub>HSI</sub>
Internal high-speed RC oscillation	Fsys= F <sub>HSI</sub> /2	T <sub>WAIT</sub> =1032*2/F <sub>HSI</sub> +16*1/F <sub>HSI</sub>
(F <sub>HSI</sub> )		
	Fsys= F <sub>HSI</sub> /64	Twait=1032*64/F <sub>HSI</sub> +16*1/F <sub>HSI</sub>
Interna llow speedRCoscillation		
(Flfintosc)		T <sub>WAIT</sub> =15/F <sub>LFINTOSC</sub>
XT oscillation (F <sub>XT</sub> )		T <sub>WAIT</sub> =2048/F <sub>XT</sub>



# 6. I/O Port

Chip has 4 I/O port: PORTA、PORTB、PORTC、PORTD (max. of 26 I/O).read/write port data register can directly read/write these ports.

Port	Bit	Pin Description	I/O
	0	Schmitt trigger input, push-pull output, AN0, TK0, LEDdriverSEG 🛛	I/O
	1	Schmitt trigger input, push-pull output, AN1, TK1, LEDdriverSEG	I/O
	2	Schmitt trigger input, push-pull output, AN2, TK2, LEDdriverSEG	I/O
PORTA	3	Schmitt trigger input, push-pull output, AN3, TK3, LEDdriverSEG	I/O
FURIA	4	Schmitt trigger input, push-pull output, AN4, TK4, LEDdriverSEG [], T0CKI	I/O
	5	Schmitt trigger input, push-pull output, AN5, TK5, LEDdriverSEG 🛛	I/O
	6	Schmitt trigger input, push-pull output, AN6, TK6, LEDdriverSEG 🛛	I/O
	7	Schmitt trigger input, push-pull output, AN7, TK7, LEDdriverSEG 🛛	I/O
	0	Schmitt trigger input, push-pull output, AN8, TK8, LEDdriverSEG 🛛	I/O
	1	Schmitt trigger input, push-pull output, AN9, TK9, LEDdriverSEG [], externalinterruptinput	I/O
	2	Schmitt trigger input, push-pull output, AN10, TK10, LEDdriverSEG 🛛	I/O
PORTB	3	Schmitt trigger input, push-pull output, AN11, TK11, CCP, RX1/DT1	I/O
	4	Schmitt trigger input, push-pull output, AN12, TK12, CCP, TX1/CK1, T1G	I/O
	5	Schmitt trigger input, push-pull output, AN13, TK13, T1CKI, OSCO	I/O
	6	Schmitt trigger input, push-pull output, AN14, TK14, OSCI	I/O
	0	Schmitt trigger input, push-pull output, AN23, TK23, LCD/LEDdriverCOM	I/O
	1	Schmitt trigger input, push-pull output, AN22, TK22, LCD/LEDdriverCOM	I/O
	2	Schmitt trigger input, push-pull output, AN21, TK21, LCD/LEDdriverCOM	I/O
PORTC	3	Schmitt trigger input, push-pull output, AN20, TK20, LCD/LEDdriverCOM 🛛	I/O
TORIC	4	Schmitt trigger input, push-pull output, AN19, TK19, LCD/LEDdriverCOM	I/O
	5	Schmitt trigger input, push-pull output, AN18, TK18, LCD/LEDdriverCOM	I/O
	6	Schmitt trigger input, push-pull output, AN17, TK17, LCD/LEDdriverCOM [], CCP, RX1/DT1	I/O
	7	Schmitt trigger input, push-pull output, AN16, TK16, LCD/LEDdriverCOM [], CCP, TX1/CK1	I/O
	0	Schmitt trigger input, push-pull output, AN25, TK25, program clock input, RX0/DT0	I/O
PORTD	1	Schmitt trigger input, push-pull output, AN24, TK24, program data input/output, TX0/CK0	I/O
	2	Schmitt trigger input, push-pull output, AN15, TK15	I/O

< Table 6-1: port configuration summary >



## 6.1 I/O Port Structure

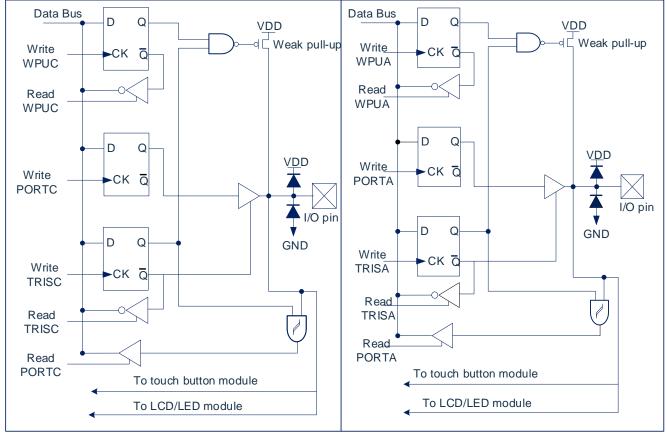


Fig 6-1: I/Oport structure (1)



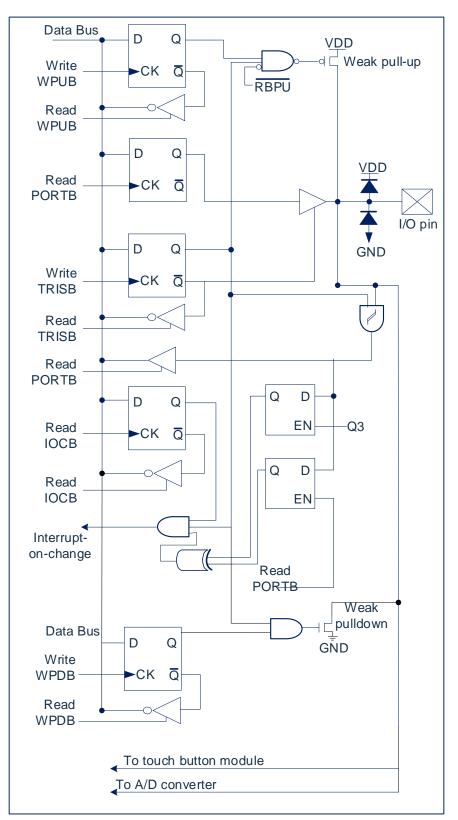


Fig 6-2: I/O port structure (2)

## 6.2 PORTA

#### 6.2.1 PORTA Data and Direction Control

PORTA is 8 Bit bi-directional port. Its corresponding data direction register is TRISA. Setting 1 bit of TRISA to be 1 can configure the corresponding pin to be input. Setting 1 bit of TRISA to be 0 can configure the corresponding pin to be output.

Reading PORTA register reads the pin status. Writing PORTA write to port latch. All write operation are read-change-write. Hence, write 1 port means read the pin electrical level of the port, change the value and write the value intoport latch. Even when PORTA pin is used as analog input, TRISA registerstill control the direction of PORTA pin. When use PORTA pin as analog input, user must make sure the bits in TRISA register are kept as 1.

Registes related to PORTA ports are PORTA、TRISA、WPUA、ANSEL0 and etc.

PORTA	data	register	PORTA	(05H)	)
	aata	regiotor			,

05H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
reset value	Х	Х	Х	Х	Х	Х	Х	Х

Bit7~Bit0

#### PORTA<7:0>: PORTAI/O pinbit;

- TRISAx=1
  - 1= Port pin level>VIH;
  - 0= Port pin level<VIL.

TRISAx=0

- 1= Port output high level;
- 0= Port output low level.

#### PORTA direction register TRISA (85H)

85H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
TRISA	TRISA6	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
reset value	1	1	1	1	1	1	1	1

Bit7~Bit0

TRISA<7:0>: PORTA;

1= PORTA pin set to be input;

0= PORTA pin set to be output

#### example: procedure for PORTA

LDIA	B'11110000'	;set PORTA<3:0> as output port, PORTA<7:4>as input port
LD	TRISA,A	
LDIA	03H	;PORTA<1:0>output high level, PORTA<3:2>output low level
LD	PORTA,A	;since PORTA<7:4>are input ports, 0 or 1 does not matter



#### 6.2.2 PORTA Pull Up Resistance

Each PORTA pin has an internal weak pull up that can be individually configured. The control bits WPUA<7:0> enable or disable each weak pull up. When the portpin is configured as output, its weak pull up will be automatically cut off.

107H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
WPUA	WPUA7	WPUA6	WPUA5	WPUA4	WPUA3	WPUA2	WPUA1	WPUA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0

#### PORTA pull up resistance register WPUA (107H)

Bit7~Bit0 WPUA<7:0>: Weak pull up register bit

1= Enable pull up

0= Disable pull up

Note: If pin is configured as output, weak pull up will be automatically disabled

#### 6.2.3 PORTA Analog Control Selection

The ANSEL0 register is used to configure the input mode of I/Opin to analog mode. Setting the appropriate bit in ANSEL0 to 1 will cause all digital read operations of the corresponding pin to return to 0 and make the analog function of the pin work normally. The state of the ANSEL0 bit has no effect on the digital output function. The pin with TRIS cleared and ANSEL0 set to 1 will still be used as a digital output, but the input mode will become an analog mode. This can cause unpredictable results when performing read-modify-write operations on the affected port.

	0	0	(	/				
09H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ANSEL0	ANS7	ANS6	ANS5	ANS4	ANS3	ANS2	ANS1	ANS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0

PORT A analog selection register ANSEL0 (09H)

Bit7~Bit0

ANS<7:0>: Analog selection bit, select the digital or analog function of pin AN<7:0>

<sup>1=</sup> Analog input

<sup>0=</sup> Digital I/O



## 6.3 PORTB

#### 6.3.1 PORTB Data and Direction

PORTB is a 7Bit wide bi-directional port. The corresponding data direction register is TRISB. Set a bit in TRISB to 1 (=1) to make the corresponding PORTBpin as the input pin. Clearing a bit in TRISB (=0) will make the corresponding PORTB pin as the output pin.

Reading the PORTB register reads the pin status and writing to the register will write the port latch. All write operations are read-modify-write operations. Therefore, writing a port means to read the pin level of the port first, modify the read value, and then write the modified value into the portdata latch. Even when the PORTB pin is used as an analog input, the TRISB register still controls the direction of the PORTB pin. When using the PORTB pin as an analog input, the user must ensure that the bits in the TRISB register remain set as 1.

Related registers with PORTB port include PORTB, TRISB, WPUB, IOCB, WPDB, ANSEL1, etc. PORTB data register PORTB (06H)

06H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PORTB		RB6	RB5	RB4	RB3	RB2	RB1	RB0
R/W		R/W						
Reset value		Х	Х	Х	Х	Х	Х	Х

Bit7 Bit6~Bit0	Not used PORTB<6:0>: TRISBx=1	PORTB I/O pin bit
	1=	Port pin level >VIH;
	0=	Port pin level <v৷∟< td=""></v৷∟<>
	TRISBx=0	
	1=	Port output high level;
	0=	Port output low level

#### PORTB direction register TRISB (86H)

	0		,					
86H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
TRISB		TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
R/W		R/W						
reset value		1	1	1	1	1	1	1

Bit7 Bit6~Bit0 Not used

TRISB<6:0>: PORTB tri-state control bit

1= PORTB pin configured as input

0= PORTB pin configured as output

#### example: PORTB port procedure

CLR	PORTB	;clear data register
LDIA	B'00110000'	;set PORTB<5:4> as input port, others as output port
LD	TRISB,A	





#### 6.3.2 PORTB Pull up Resistance

Each PORTB pin has an internal weak pull up that can be individually configured. The control bits WPUB<6:0> enable or disable each weak pull up. When the port pin is configured as output, its weak pull up will be automatically cut off. At power-on reset, weak pull up is prohibited by the RBPU bit of the OPTION\_REG register.

95H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
WPUB		WPUB6	WPUB5	WPUB4	WPUB3	WPUB2	WPUB1	WPUB0
R/W		R/W						
Reset value		0	0	0	0	0	0	0

PORTB pull up resistance register WPUB (95H)

Bit7	Not used	
Bit6~Bit0	WPUB<6:0>:	Weak pull up register bit
	1=	Enable pull up

0= Disable pull up

#### Note:

To individually enable any pull up, the global RBPU bit of OPTION\_RE Gregister must be cleared.
 If the pin is configured as output or analog input, weak pull up will be automatically prohibited.

#### 6.3.3 PORTB Analog Selection Control

The ANSEL1 register is used to configure the input mode of I/Opin to analog mode. Setting the appropriate bit in ANSEL1 to 1 will cause all digital read operations of the corresponding pin to return to 0 and make the analog function of the pin work normally. The state of the ANSEL1 bit has no effect on the digital output function. The pin whose TRIS is cleared and ANSEL1 is set to 1 is still used as a digital output, but the input mode will become an analog mode. This can cause unpredictable results when executing read-modify-write operations on the affected port.

89H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ANSEL1		ANS14	ANS13	ANS12	ANS11	ANS10	ANS9	ANS8
R/W		R/W	R/W	R/W	R/W	R/W	R/W	R/W
reset value		0	0	0	0	0	0	0

PORTB analog selection register ANSEL1 (89H)

Bit7 Bit6~Bit0 Reference PORTD analog Selection Control

ANS<14:8>: Analog selection bits, select the analog or digital functions of pin AN<14:8>.

1= Pin set as analog input.

0= Digital I/O

# 

#### 6.3.4 PORTB Level Chage Interrupt Interrupt

All PORTB pins can be individually configured as level change interrupt pins. The control bit IOCB<6:0> allows or disables the interrupt function of each pin. Disable pin level change interrupt function when power on reset.

For the pin that has allowed level change interrupt, compare the value on the pin with the old value latched when PORTB was read last time. Perform a logical OR operation with the output "mismatch" of the last read operation to set the PORTB level change interrupt flag (RBIF) in the INTCON register as 1.

This interrupt can wake up the device from sleep mode, and the user can clear the interrupt in the interrupt service program in the following ways:

-Read or write to PORTB. This will end the mismatch state of the pin level.

-Clear the flag bit RBIF.

The mismatch status will continuously set the RBIF flag bit as 1. Reading or writing PORTB will end the mismatch state and allow the RBIF flag to be cleared. The latch will keep the last read value from the undervoltage reset. After reset, if the mismatch still exists, the RBIF flag will continue to be set as 1.

Note: If the level of the I/O pin changes during the read operation (beginning of the Q2 cycle), the RBIF interrupt flag bit will not be set as 1. In addition, since reading or writing to a port affects all bits of the port, special care must be taken when using multiple pins in interrupt-on-change mode. When dealing with the level change of one pin, you may not notice the level change on the other pin.

96H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
IOCB		IOCB6	IOCB5	IOCB4	IOCB3	IOCB2	IOCB1	IOCB0
R/W		R/W						
Reset value		0	0	0	0	0	0	0

#### PORTB level change interrupt register IOCB (96H)

Bit7 Not used

Bit6~Bit0

IOCB<6:0>

> Control bit of level change interrupt of PORTB

- 1= enable level change interrupt
- 0= disable level change interrupt



#### 6.3.5 PORTB Pull Down Resistance

Each PORTB pin has an internal weak pull-down that can be individually configured. The control bits WPDB<6:0> enable or disable each weak pull-down. When the port pin is configured as output, its weak pull-down will automatically cut off.

#### PORTB pull down resistance registerWPDB (97H)

97H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
WPDB		WPDB6	WPDB5	WPDB4	WPDB3	WPDB2	WPDB1	WPDB0
R/W		R/W						
resetvalue		0	0	0	0	0	0	0

Bit7

Not used

Bit6~Bit0

WPDB<6:0>: We

0>: Weak pull-down register bit1= Enable pull down

0= Disable pull down

Note: If the pin is configured as output or analog input, weak pull-down will be automatically disabled.



#### 6.4 PORTC

#### 6.4.1 PORTC Data and Direction

PORTC is an 8-bit wide bidirectional port. The corresponding data direction register is TRISC. Set a certain position in TRISC to 1 (=1) to make the corresponding PORTC pin as the input pin. Clearing a bit in TRISC (=0) will make the corresponding PORTC pin as the output pin.

Reading the PORTC register reads the pin status and writing to the register will write the port latch. All write operations are read-modify-write operations. Therefore, writing a port means reading the pin level of the port first, modifying the read value, and then writing the modified value to the port data latch.

07H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	Х	Х	Х	Х	Х	Х	Х	Х

PORTC data register PORTC (07H)

Bit7~Bit0

PORTC<7:0> PORTC I/O pin bit

TRISCx=1

1= Port pin level >VIH; 0= Port pin level <VIL

TRISCx=0

1= Port output high level;

0= Port output low level

#### PORTC direction register TRISC (87H)

87H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	1	1	1	1	1	1	1	1

Bit7~Bit0

#### TRISC<7:0>: Control bit of PORTC tri-state

- 1= PORTC pin configured as input
- 0= PORTC pin configured as output

#### Example: procedure for PORTC

1 1		
CLR	PORTC	;clear data register
LDIA	B'01110000'	;set PORTC<3:0> as output, PORTC<6:4> as input
LD	TRISC,A	



#### 6.4.2 PORTC pull up resistance

Each PORTC pin has an internal weak pull up that can be individually configured. The control bits WPUC<7:0> enable or disable each weak pull up. When the port pin is configured as output, its weak pull up will be automatically cut off.

	rocistanco	register WPUC	· (109L)
F OIXTOPUII UP	resistance	IEGISIEI WFUC	

108H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
WPUC	WPUC7	WPUC6	WPUC5	WPUC4	WPUC3	WPUC2	WPUC1	WPUC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 WPUC<7:0>: Weak pull up register bit

1= Enable pull up

0= Disable pull up

Note: If the pin is configured as output or analog input, weak pull up will be automatically disabled.

#### 6.4.3 PORTC analog control selection

The ANSEL2 register is used to configure the input mode of I/Opin to analog mode. Setting the appropriate bit in ANSEL2 to 1 will cause all digital read operations of the corresponding pin to return to 0 and make the analog function of the pin work normally. The state of the ANSEL2 bit has no effect on the digital output function. The pin with TRIS cleared and ANSEL2 set to 1 is still used as a digital output, but the input mode will become an analog mode. This can cause unpredictable results when performing read-modify-write operations on the affected port.

109H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
ANSEL2	ANS23	ANS22	ANS21	ANS20	ANS19	ANS18	ANS17	ANS16	
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
Reset value	0	0	0	0	0	0	0	0	

PORTB analog selection register ANSEL2 (109H)

Bit7~Bit0 ANS

ANS<23:16>: Analog selection bit, select the analog or digital function of pin AN<23:16>

respectively.

1= Pin set as Analog input

0= Set as Digital I/O



## 6.5 PORTD

#### 6.5.1 **PORTD** data and direction

PORTD is a 3-bit wide bidirectional port. The corresponding data direction register is TRISD. Set a certain position in TRISD to 1 (=1) to make the corresponding PORTD pin as the input pin. Clearing a bit in TRISD (=0) will make the corresponding PORTD pin as the output pin.

Reading the PORTD register reads the pin status and writing to the register will write the port latch. All write operations are read-modify-write operations. Therefore, writing a port means reading the pin level of the port first, modifying the read value, and then writing the modified value to the port data latch.

08H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PORTD						RD2	RD1	RD0
R/W						R/W	R/W	R/W
Reset value						Х	Х	Х

PORTD data register PORTD (08H)

Bit7~Bit3 Bit2~Bit0 Not used PORTD<2:0>: PORTD I/O pin bit

TRISDx=1

1= Port pin level >VIH;

0= Port pin leve<VIL TRISDx=0

1= Port output high level;

0= Port output low level

#### PORTD direction register TRISD (88H)

88H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
TRISD						TRISD2	TRISD1	TRISD0
R/W						R/W	R/W	R/W
Reset value						1	1	1

Bit7~Bit3 Bit2~Bit0

#### Not used TRISD<2:0>: Control bit of PORTD tri-state 1= PORTD pin configured as input

0= PORTD pin configured as output

#### example: procedure for PORTD

CLR	PORTD	;clear data register
LDIA	B'01111111'	;set PORTD<2:0> as input
LD	TRISD,A	



#### 6.5.2 **PORTD** pull up resistance

Each PORTD pin has an internal weak pull up that can be individually configured. The control bits WPUD<2:0> enable or disable each weak pull up. When the port pin is configured as output, its weak pull up will be automatically cut off.

#### PORTD pull up resistance register WPUD (8EH)

8EH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
WPUD						WPUD2	WPUD1	WPUD0
R/W						R/W	R/W	R/W
Reset value						0	0	0

Bit7~Bit3	Not used	
Bit2~Bit0	WPUD<2:0>:	Weak pull up register bit
	1=	Enable pull up

- 0= Disable pull up

Note: If the pin is configured as output or analog input, weak pull up will be automatically prohibited. The AN25/AN24 channel corresponding to RD0/RD1 has no analog selection register.

#### 6.5.3 PORTD Analog Selection Control

The ANSEL1 register is used to configure the input mode of I/Opin to analog mode. Setting the appropriate bit in ANSEL1 to 1 will cause all digital read operations of the corresponding pin to return to 0 and make the analog function of the pin work normally. The state of the ANSEL1 bit has no effect on the digital output function. The pin whose TRIS is cleared and ANSEL1 is set to 1 is still used as a digital output, but the input mode will become an analog mode. This can cause unpredictable results when executing read-modify-write operations on the affected port.

89H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ANSEL1	ANS15							
R/W	R/W							
reset value	0							

PORTB	analog	selection	register ANSEL1	(89H)
1 01110	analog	0010001011	I Oglotor / TTOEL	(0011)

Bit7

ANS<15>: Analog selection bits, select the analog or digital functions of pin AN<15>.

1= Pin set as analog input.

0= Digital I/O

Bit6~Bit0 Reference PORTB analog Selection Control



## 6.6 I/O Usage

## 6.6.1 Write I/O Port

The chip's I/O port register, like the general universal register, can be written through data transmission instructions, bit manipulation instructions, etc.

_			
	LD	PORTA,A	;pass value of ACC to PORTA
	CLRB	PORTB,1	;clear PORTB.1
	CLR	PORTC	;clear PORTC
	SET	PORTA	;set all output port of PORTA as 1
	SETB	PORTB,1	;set PORTB.1as 1

#### 6.6.2 Read I/O Port

Example: write I/O	port program
--------------------	--------------

LD	A,PORTA	;pass value of PORTA to ACC
SNZB	PORTA,1	; check whether PORTA, port 1 is 1, if it is 1, skip the next statement
SZB	PORTA,1	; check if PORTA, 1 port is 0, if 0, skip the next statement

Note: When the user reads the status of an I/O port, if the I/O port is an input port, the data read back by the user will be the state of the external level of the port line. If the I/O port is an output port then the read value will be the data of the internal output register of this port.



## 6.7 Precautions for I/O Port Usage

When operating the I/O port, pay attention to the following aspects:

- 1. When I/O is converted from output to input, it is necessary to wait for several instruction periods for the I/O port to stabilize.
- 2. If the internal pull up resistor is used, when the I/O is converted from output to input, the stable time of the internal level is related to the capacitance connected to the I/O port. The user should set the waiting time according to the actual situation. Prevent the I/O port from scanning the level by mistake.
- 3. When the I/O port is an input port, its input level should be between "VDD+0.7V" and "GND-0.7V". If the input port voltage is not within this range, the method shown in the figure below can be used.

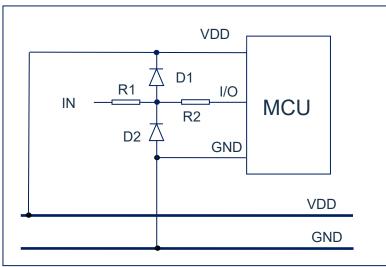


Fig 6-3: The input voltage is not within the specified range

4. If a longer cable is connected to the I/O port, please add a current limiting resistor near the chip I/O to enhance the MCU's anti-EMC capability.



# 7. Interrupt

## 7.1 Interrupt General

The chip has the following interrupt source:

- TIMER0 overflow interrupt
- TIMER2 match interrupt
- PORTB level change interrupt
- CCP1/CCP2 interrupt
- USART0/1receive/transmit interrupt
- LVD interrupt

- TIMER1 overflow interrupt
- INT interrupt
- A/D interrupt
- MSSP interrupt
- Program EEPROM write interrupt

The interrupt control register (INTCON) and the peripherals interrupt request register (PIR1, PIR2) record various interrupt requests in their respective flag bits. The INTCON register also includes various interrupt enable bits and global interrupt enable bits.

The global interrupt enable bit GIE (INTCON<7>) allows all unmasked interrupts when set to 1, and prohibits all interrupts when cleared. Each interrupt can be prohibited through the corresponding enable bits in the INTCON, PIE1, and PIE2 registers. GIE is cleared when reset.

Executing the "return from interrupt" instructions, RETI, will exit the interrupt service program and set the GIE bit to 1, thereby re-allowing unshielded interrupt.

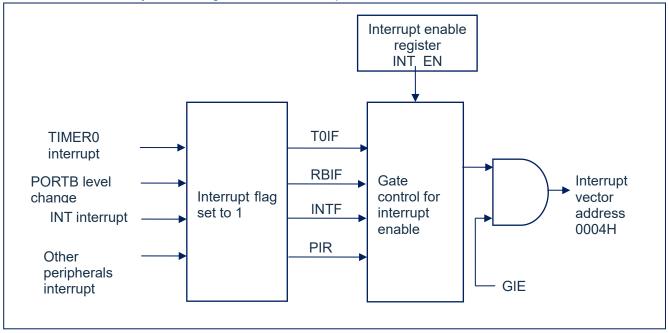


Fig 7-1: interrupt theory

## 7.2 Interruptcontrol Register

#### 7.2.1 Interrupt Control Register

The interrupt control registerINTCON is a readable and writable register, including the allowable and flag bits for TMR0 register overflow and PORTB port level change interrupt.

When an interrupt condition occurs, regardless of the state of the corresponding interrupt enable bit or the global enable bit GIE (in the INTCON register), the interrupt flag bit will be set to 1. The user software should ensure that the corresponding interrupt flag bit is cleared before allowing an interrupt.

0BH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
INTCON	GIE	PEIE	TOIE	INTE	RBIE	T0IF	INTF	RBIF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0

Interrupt control register INTCON (0BH)

Bit7	GIE:	Global interrupt enable bit;
	1=	Enable all unshielded interrupt;
	0=	Disable all interrupt
Bit6	PEIE:	Peripherals interrupt enable bit;
	1=	Enable all unshielded peripherals interrupt;
	0=	Disable all peripherals interrupt.
Bit5	T0IE:	TIMER0 overflow interrupt enable bit;
	1=	Enable TIMER0 interrupt;
	0=	Disable TIMER0 interrupt
Bit4	INTE:	INT external interrupt enable bit;
	1=	Enable INT external interrupt;
	0=	Disable INT external interrupt
Bit3	RBIE:	PORTB level change interruptenable bit (1);
	1=	Enable PORTB level change interrupt;
	0=	Disable PORTB level change interrupt
Bit2	T0IF:	TIMER0 overflow interrupt enable bit (2);
	1=	TMR0 register overflow already (must clear through software);
	0=	TMR0 register not overflow
Bit1	INTF:	INT external interrupt flag bit;
	1=	INT external interrupt happens (must clear through software);
	0=	INT external interrupt not happen
Bit0	RBIF:	PORTB level change interrupt flag bit;
	1=	The level of at least one pin in the PORTB port has changed (must clear through
		software);
	0=	None of the PORTB universal I/O pin status has changed.

#### Note:

- 1) The IOCB register must also be enabled, and the corresponding port must be set to input state.
- 2) The T0IF bit is set as 1when TMR0 rolls over to 0. Reset will not change TMR0 and should be initialized before clearing the T0IF bit.



## 7.2.2 Peripherals Interrupt Enable Register

The peripherals interrupt enable register has PIE1 and PIE2. Before allowing any peripherals interrupt, the PEIE bit of the INTCON register must be set to 1.

	0	
Peripherals interr	rupt enable regist	ter PIE1 (8CH)

1	1	5	(	/				
8CH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PIE1		ADIE	RC0IE	TX0IE	SSPIE	CCP1IE	TMR2IE	TMR1IE
R/W		R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value		0	0	0	0	0	0	0

Bit7	Not used, read 0	
Bit6	ADIE:	A/D converter (ADC)interrupt enable bit;
	1=	enable ADC interrupt;
	0=	disable ADC interrupt
Bit5	RC0IE:	USART0 receive interrupt enable bit;
	1=	enable USART0 receive interrupt;
	0=	disable USART0 receive interrupt.
Bit4	TX0IE:	USART0 transmit interrupt enable bit;
	1=	enable USART0 transmit interrupt;
	0=	disable USART0 transmit interrupt.
Bit3	SSPIE:	Main synchronous serial port (MSSP)interrupt enable bit;
	1=	enable MSSP interrupt;
	0=	disable MSSP interrupt.
Bit2	CCP1IE:	CCP1 interrupt enable bit;
	1=	enable CCP1 interrupt;
	0=	disable CCP1 interrupt.
Bit1	TMR2IE:	TIMER2 and PR2 match interrupt enable bit;
	1=	enable TMR2 and PR2 match interrupt;
	0=	disable TMR2 and PR2 match interrupt.
Bit0	TMR1IE:	TIMER1 overflow interrupt enable bit;
	1=	enable TIMER1 overflow interrupt;
	0=	disable TIMER1 overflow interrupt.



#### Peripherals interrupt enable registerPIE2 (8DH)

8DH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PIE2	LVDIE			EEIE	BCLIE	RC1IE	TX1IE	CCP2IE
R/W	R/W			R/W	R/W	R/W	R/W	R/W
Reset value	0			0	0	0	0	0

Bit7	LVDIE	LVDenable bit;
	1=	enable LVDinterrupt;
	0=	disable LVDinterrupt.
Bit6~Bit5	Not used.	
Bit4	EEIE:	Program EEPROM write operation interrupt enable bit;
	1=	enable program EEPROM write operationinterrupt;
	0=	disable program EEPROM write operation interrupt.
Bit3	BCLIE:	bus conflict interrupt enable bit;
	1=	enable bus conflict interrupt;
	0=	disable bus conflict interrupt.
Bit2	RC1IE:	USART1 receive interrupt enable bit;
	1=	enable USART1 receive interrupt;
	0=	disable USART1 receive interrupt.
Bit1	TX1IE:	USART1t ransmit interrupt enable bit;
	1=	enable USART1 transmit interrupt;
	0=	disable USART1 transmit interrupt.
Bit0	CCP2IE:	CCP2 interrupt enable bit;
	1=	enable CCP2 interrupt;
	0=	disable CCP2 interrupt.



#### 7.2.3 Peripherals Interrupt Request Register

The peripherals interrupt request register is PIR1 and PIR2. When an interrupt condition occurs, regardless of the state of the corresponding interrupt enable bit or the global enable bit GIE, the interrupt flag bit will be set to 1. The user software should ensure that the interrupt is set before allowing an interrupt. The corresponding interrupt flag bit is cleared.

1		1 5	( = =	/				
0CH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PIR1		ADIF	RC0IF	TX0IF	SSPIF	CCP1IF	TMR2IF	TMR1IF
R/W		R/W	R	R	R/W	R/W	R/W	R/W
Reset value		0	0	0	0	0	0	0

Peripherals interrupt request register PIR1 (0CH)

Bit7	Not used, read 0	
Bit6	ADIF:	A/D converter interrupt flag bit;
	1=	A/D conversion complete (must clear through software);
	0=	A/D conversion not complete or not start.
Bit5	RC0IF:	USART0 receive interrupt flag bit;
	1=	USART0 receive buffer full (clear through reading RCREG);
	0=	USART0 receive buffer empty.
Bit4	TX0IF:	USART0 transmit interrupt flag bit;
	1=	USART0 transmit buffer empty (clear through TXREG);
	0=	USART0 transmit buffer full.
Bit3	SSPIF:	Main synchronous serial port (MSSP)interruptf lag bit.
	1=	The MSSP interrupt condition is met. Before returning from the interrupt service
		program, it must clear through software. The conditions for making this bit 1 are::
		- SPI;
		- transmit/receive happens;
		- I <sup>2</sup> C slave/master control;
		<ul> <li>transmit/ receive happens;</li> </ul>
		- I <sup>2</sup> C master control;
		<ul> <li>The start condition that occurs is done by MSSP mod;</li> </ul>
		<ul> <li>The stop condition that occurs is completed by MSSP mod;</li> </ul>
		<ul> <li>The restart condition that occurs is done by MSSPmod;</li> </ul>
		<ul> <li>The respond condition that occurs is done by MSSPmod;</li> </ul>
		- The start condition occurs when the MSSP mod is idle (multi-host system);
		- The stop condition occurs when the MSSP mod is idle (multi-host system);.
	0=	No MSSP interrupt condition is met.
Bit2	CCP1IF:	CCP1 interrupt flag bit.
	Capture mode:	1= Capture for TMR1 register happens (must clear through software);
		0= Capture for TMR1 register not happen
	Compare mode:	
		1= Compare match for TMR1 register happens (must clear through software);
		0= Compare match for TMR1 register not happen.
	PWM mode:	Not used under this mode.
Bit1	TMR2IF:	TIMER2 and PR2 match interrupt flag bit.
	1=	TIMER2 and PR2 match heppens (must clear through software);
	0=	TIMER2 and PR2 not match.
Bit0	TMR1IF:	TIMER1 overflow interrupt flag bit.
	1=	
	0=	TMR1 register not overflow.



0DH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
PIR2	LVDIF			EEIF	BCLIF	TX1IF	RX1IF	CCP2IF	
R/W	R/W			R/W	R/W	R	R	R/W	
Reset value	0			0	0	0	0	0	
Bit7	LVDIF 1=		rruptflag bit; errupt happer	ns:					
	0=		rrupt not hap						
Bit6~Bit5	Not used.								
Bit4	EEIF:	Program	EEPROM v	vrite operation	interrupt flag	bit;			
	1=	write o	operation cor	mplete (must o	clear through	software);			
	0=	write o	write operation not complete or not start.						
Bit3	BCLIF:	bus conf	bus conflict interrupt flag bit;						
	1=	When co	onfigured as	I <sup>2</sup> Cmaster cor	ntrol mode, bu	us conflict hap	pens in MSSI	⊃;	
	0=	No bus o	conflict.						
Bit2	TX1IF:	USART1	transmit int	errupt flag bit;	•				
	1=	USART1	l transmit bu	ffer empty (cle	ear through w	riting TXREG	1);		
	0=	USART1	transmit bu	ffer full.					
Bit1	RC1IF:	USART1	receive inte	errupt flag bit;					
	1=	USART1	receive buf	fer full (clear t	hrough readir	ng RCREG1);			
	0=	USART1	receive buf	fer empty.					
Bit0	CCP2IF:	CCP2 in	terrupt flag b	oit.					
(	Capture mode:								
				register happe		ar through sof	tware);		
		0= captu	re of TMR1	register not ha	appen.				
	Compare								
	mode:			f TMR1 regist	••••		ough software	);	
		-		f TMR1 regist	er not hepper	۱.			
PWM mode: Not used under this mode.									

#### Peripherals interrupt request registerPIR2 (0DH)



#### **Protection Methods for Interrupt** 7.3

After an interrupt request occurs and is responded, the program goes to 0004H to execute the interrupt sub-routine. Before responding to the interrupt, the contents of ACC and STATUS must be saved. The chip does not provide dedicated stack saving and unstack recovery instructions, and the user needs to protect ACC and STATUS by himself to avoid possible program operation errors after the interrupt ends.

Example: Stack prote	ection for ACC and	I STATUS	
	ORG	0000H	
	JP	START	;start of user program address
	ORG	0004H	
	JP	INT_SERVICE	;interrupt service program
	ORG	0008H	
START:			
INT_SERVICE:			
PUSH:			;entrance for interruptservice program, save ACC and STATUS
	LD	ACC_BAK,A	;save the value of ACC (ACC_BAK needs to be defined)
	SWAPA	STATUS	
	LD	STATUS_BAK,A	;save the value of STATUS (STATUS_BAK needs to be defined)
	•••		
POP:			;exit for interrupt serice program, restore ACC and STATUS
	SWAPA	STATUS_BAK	
	LD	STATUS,A	;restore STATUS
	SWAPR	ACC_BAK	;restore ACC
	SWAPA	ACC_BAK	
	RETI		

#### **Interrupt Priority and Multi-interrupt Nesting** 7.4

The priority of each interrupt of the chip is equal. When an interrupt is in progress, it will not respond to the other interrupt. Only after the "RETI" instructions are executed, the next interrupt can be responded to.

When multiple interrupts occur at the same time, the MCU does not have a preset interrupt priority. First, the priority of each interrupt must be set in advance; second, the interrupt enable bit and the interrupt control bit are used to control whether the system responds to the interrupt. In the program, the interrupt control bit and interrupt request flag must be checked.



# 8. TIMER0

## 8.1 TIMER0 General

TIMER0 is composed of the following functions:

- 8-bit timer/counter register (TMR0);
- 8-bit pre-scaler (shared with watchdog timer);
- Programmable internal or external clock source;
- Programmable external clock edge selection;
- overflow interrupt.

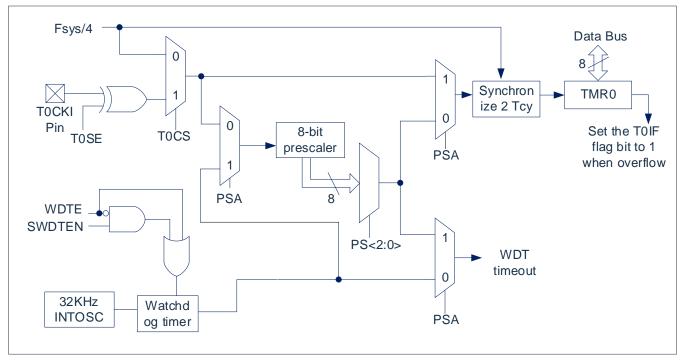


Fig 8-1: TIMER0/WDT mod structure

Note:

- 1. TOSE, TOCS, PSA, PS<2:0> are the bits in OPTION\_RE Gregister.
- 2. SWDTEN is a bit in the WDTCON register.
- 3. WDTE bit is in CONFIG.



## 8.2 Working Principle for TIMER0

The TIMER0 mod can be used as an 8-bit timer or an 8-bit counter.

#### 8.2.1 8-bit Timer Mode

When used as a timer, the TIMER0 mod will be incremented every instruction period (without pre-scaler). The timer mode can be selected by clearing the T0CS bit of the OPTION\_REG register to 0. If a write operation is performed to the TMR0 register, the next two Each instruction period will be prohibited from incrementing. The value written to the TMR0 register can be adjusted so that a delay of two instruction periods is included when writing TMR0.

#### 8.2.2 8-bit Counter Mode

When used as a counter, the TIMER0 mod will increment on every rising or falling edge of the T0CKI pin. The incrementing edge depends on the T0SE bit of the OPTION\_REG register. The counter mode can be selected by setting the T0CS bit of the OPTION\_REG register to 1.

#### 8.2.3 Software Programmable Pre-scaler

TIMER0 and watchdog timer (WDT) share a software programmable pre-scaler, but they cannot be used at the same time. The allocation of the pre-scaler is controlled by the PSA bit of the OPTION\_REG register. To allocate the pre-scaler to TIMER0, the PSA bit must be cleared to 0.

TIMER0mod has 8 selections of prescaler ratio, ranging from 1:2 to 1:256. The prescaler ratio can be selected through the PS<2:0> bits of the OPTION\_REG register. To make TIMER0 mod have a 1:1 prescaler, the pre-scaler must be assigned to the WDT mod.

The pre-scaler is not readable and writable. When the pre-scaler is assigned to the TIMER0 mod, all instructions written to the TMR0 register will clear the pre-scaler. When the pre-scaler is assigned to the WDT, the CLRWDT instructions will also clear the pre- scaler and WDT.

#### 8.2.4 Switch Prescaler Between TIMER0 and WDT Module

After assigning the pre-scaler to TIMER0 or WDT, an unintentional device reset may occur when switching the prescaler. To change the pre-scaler from TIMER0 to WDT mod, the following instructions must be executed sequence.

Modify pre-scale	er (TMR0-WDT)		
	CLRB	INTCON,GIE	; Turn off the interrupt enable bit to avoid entering the interrupt program when the following specific time series is executed
	LDIA	B'00000111'	
	ORR	OPTION_REG,A	;set pre-scaler to max. value
	CLR	TMR0	;clear TMR0
	SETB	OPTION_REG,PSA	;set pre-scaler allocate to WDT
	CLRWDT		;clear WDT
	LDIA	B'xxxx1xxx'	;set new pre-scaler
	LD	OPTION_REG,A	
	CLRWDT		;clear WDT
	SETB	INTCON,GIE	;if the program needs to use interrupt, turn on the enable bit here



To change the pre-scaler from WDT to TIMER0 mod, the following sequence of instructions must be executed.

## Modify pre-scaler (WDT-TMR0)

CLRWDT		;clear WDT
LDIA	B'00xx0xxx'	;set new pre-scaler
LD	OPTION_REG,A	



#### 8.2.5 TIMER0 Interrupt

When the TMR0 register overflows from FFh to 00h, a TIMER0 interrupt is generated. Every time the TMR0 register overflows, regardless of whether TIMER0 interrupt is allowed, the T0IF interrupt flag bit of the INTCON register will be set to 1. The T0IF bit must be cleared in software. TIMER0 interrupt enable bit is the T0IE bit of the INTCON register.

Note: Because the timer is turned off in sleep mode, the TIMER0 interrupt cannot wake up the processor.



## 8.3 TIMER0 Related Register

There are two registers related to TIMER0, 8-bit timer/counter (TMR0), and 8-bit programmable control register (OPTION\_REG).

TMR0 is an 8-bit readable and writable timer/counter, OPTION\_REG is an 8-bit write-only register, the user can change the value of OPTION\_REG to change the working mode of TIMER0, etc. Please refer to the application of 0 prescaler register (OPTION\_REG).

01H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
TMR0								
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	Х	Х	Х	Х	Х	Х	Х	Х

#### 8-bit timer/counter TMR0 (01H)

#### OPTION\_REG register (81H)

81H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
OPTION_REG	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0
Read/write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	1	1	1	1	1	0	1	1

Bit7	RBPU:	PORTB pull up enable bit.							
	1=	Disable PORTB pull up.							
	0=	Enab	le POR <sup>-</sup>	TB pull u	p according to latch value	in ach port.			
Bit6	INTEDG:	Interrupt edge selection bit.							
	1=	The r							
	0=	The fa	alling ed	dge of the	e INT pin triggers interrupt.				
Bit5	T0CS:								
	1=	Trans	ition ed	ge of T0	CKI pin.				
	0=	Intern	a linstru	uction pe	riod clock (Fsys/4).				
Bit4	T0SE:	TIME	R0 cloc	k source	edge selection bit.				
	1=	Increi	ment wł	nen the T	OCKI pin signal transitions	from high to low.			
	0=	Increment when the T0CKI pin signal transitions from low to high.							
Bit3	PSA:	pre-scaler allocation bit.							
	1=	pre-scaler allocated to WDT.							
	0=	pre-scaler allocated toTIMER0 mod.							
Bit2~Bit0	PS2~PS0:	Pre-allocated parameter configuration bits.							
		PS2	PS1	PS0	TMR0 Frequency division ratio	WDT Frequency division ratio			
		0	0	0	1:2	1:1			
		0	0	1	1:4	1:2			
		0	1	0	1:8	1:4			
		0	1	1	1:16	1:8			
		1	0	0	1:32	1:16			
		1	0	1	1:64	1:32			
		1	1	0	1:128	1:64			
		1 1 1 1:256 1:128							



# 9. TIMER1

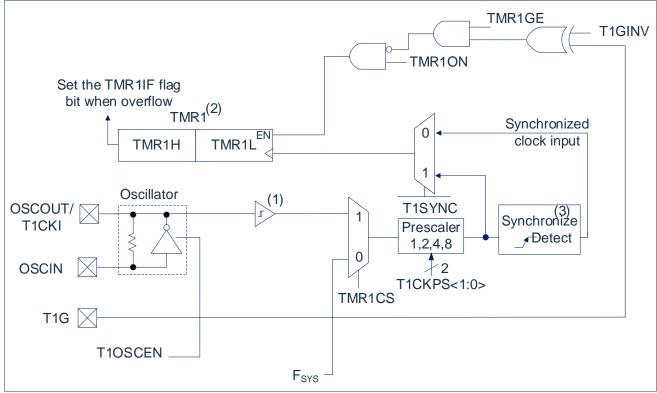
## 9.1 TIMER1 General

TIMER1 mod is a 16-bit timer/counter with the following characteristics:

• 16-bit timer/counter register (TMR1H: TMR1L)

Special event trigger function (with ECCP)

- 3-bit pre-scaler
- Synchronous or asynchronous operation
- Wake up when overflow (external clock asynchronous mode only)
- Programmable internal or external clock source
- Optional LP oscillator
- Through T1Gpingate control TIMER1 (enable counting)
- overflow interrupt
- Time base with capture/compare function



### Fig 9-1: TIMER1 structure

### Note:

- 1. The ST buffer is in low power mode when using the LP oscillator, but in high speed mode when using T1CKI.
- 2. The Timer1 register increments on the rising edge.
- 3. Do not perform synchronous during sleep.



# 9.2 Working Principle for TIMER1

TIMER1 mod is a 16-bit incremental counter accessed through a pair of register TMR1H: TMR1L. Writing to TMR1H or TMR1L can directly update the counter.

When used with internal clock source, this mod can be used as a counter. When used with external clock source, this mod can be used as a timer or counter.

# 9.3 Clock Source Selection

The TMR1CS bit of the T1CON register is used to select the clock source. When TMR1CS=0, the frequency of the clock source is FSYS. When TMR1CS=1, the clock source is provided by external.

clock source	TMR1CS			
F <sub>SYS</sub>	0			
T1CKIpin	1			

## 9.3.1 Internal Clock Source

After selecting the internal clock source, the TMR1H:TMR1L register will increase in frequency with a multiple of FSYS. The specific multiple is determined by the TIMER1 pre-scaler.



### 9.3.2 External Clock Source

After selecting the external clock source, TIMER1mod can be used as a timer or counter.

When counting, TIMER1 is incremented on the rising edge of external clock inputT1CKI. In addition, the clock in counter mode can be synchronous or asynchronous with the microcontroller system clock.

If you need an external clock oscillator, TIMER1 can use LP oscillator as clock source.

In counter mode, when one or more of the following conditions occur, a falling edge must be passed before the counter can count up for the first time on the subsequent rising edge (see Figure 9-2):

- Enable TIMER1.
- A write operation was performed on TMR1H or TMR1L.
- When TIMER1 is disabled, T1CKI is high; when TIMER1 is re-enabled, T1CKI is low.

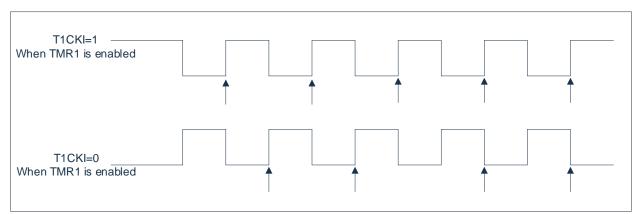


Fig 9-2: incremental edge of TIMER1

### Note:

- 1) The arrow indicates that the counter is incrementing.
- 2) In the counter mode, a falling edge must be passed before the counter can perform the first increment technique on the subsequent rising edge.



## 9.4 TIMER1 Pre-scaler

TIMER1 has four selections of prescaler ratios, allowing the input clock to be divided by 1, 2, 4 or 8. The T1CKPS bit of the T1CON register controls the prescaler counter. The prescaler counter cannot be directly read or written; but, the prescaler counter can be cleared by writing to TMR1H or TMR1L.

## 9.5 TIMER1 Oscillator

A built-in low-power 32.768KHz oscillator is connected between the T1OSI (input) pin and T1OSO (amplifier output) pin. Set the T1OSCEN control bit of the T1CON register to 1 to enable the oscillator. This oscillator will be in sleep mode Continue to run, but TIMER1 must be selected as the asynchronous counting mode.

The TIMER1 oscillator is exactly the same as the LP oscillator. The user must provide a software delay to ensure the normal oscillation of the oscillator.

When the TIMER1 oscillator is enabled, PORTB5 and PORTB are set as analog inputs.

Note: The oscillator can be used after a period of start-up and stabilization time. Therefore, before enabling TIMER1, set T1OSCEN to 1 and pass an appropriate delay.

## 9.6 TIMER1 Working Principle Under Aasynchronous Counter Mode

If the control bit T1SYNC in the T1CON register is set to 1, the external clock input will not be synchronous. The timer continues to count up asynchronously with the internal phase clock. The timer will continue to run in the sleep state, and will generate an interrupt during overflow, thereby waking up Processor. However, you should be especially careful when using software to read/write timers (see Section 9.6.1 "Read and Write to TIMER1 in Asynchronous Counter Mode").

### Note:

- 1) When switching from synchronous operation to asynchronous operation, an increment may be missed.
- 2) When switching from asynchronous operation to synchronous operation, a false increment may occur.

## 9.6.1 Read and Write Operations to TIMER1 In Asynchronous Counter Mode

When the timer uses an external asynchronous clock to work, the read operation of TMR1H or TMR1L will ensure that it is valid (the hardware is responsible). But users should keep in mind that reading two 8-bit values to read a 16-bit timer has its own problems. This is because the timer may overflow between two read operations.

For write operations, it is recommended that the user stop the timer before writing the required value. When the register is counting up, writing data to the timer register may cause write contention. This will cause unpredictability in the register pair TMR1H:TMR1L Value.



# 9.7 TIMER1 Gate Control

Software can configure the TIMER1 gate control signal source as T1G pin, which allows the device to directly use T1G to time external events.

Note: The TMR1GE bit of the T1CON register must be set to 1 to use the gate control signal of TIMER1.

You can use the T1GINV bit of the T1CON register to set the polarity of the TIMER1gate control signal. The gate control signal can come from T1Gpin. This bit can configure TIMER1 to time the high-level time or low-level time between events.

## 9.8 TIMER1 Interrupt

After a pair of TIMER1 registers (TMR1H:TMR1L) count up to FFFFH, the overflow returns to 0000H. When TIMER1 overflows, the TIMER1 interrupt flag bit of the PIR1 register is set to 1. To allow the overflow interrupt, the user should set the following bit to 1:

- TIMER1 interrupt enable bit in PIE1 register;
- PEIE bit in INTCON register;
- GIE bit in INTCON register.

Clear the TMR1IF bit in the interrupt service program to clear the interrupt.

Note: Before allowing the interrupt again, the register pair TMR1H:TMR1L and the TMR1IF bit should be cleared.

# 9.9 TIMER1 Working Principle During Sleep

TIMER1 can work in sleep mode only when it is set to asynchronous counter mode. In this mode, the external crystal or clock source can be used to make the counter count up. The timer can wake up the device through the following settings:

- The TMR1ON bit in the T1CON register must be set to 1;
- The TMR1IE bit in the PIE1 register must be set to 1;
- The PEIE bit in the INTCON register must be set to 1.

The device will be woken up at overflow and execute the next instruction. If the GIE bit in the INTCON register is 1, the device will call the interrupt service routine (0004h).

# 9.10 ECCP Capture/Compare Time Base

The ECCP mod uses the TMR1H:TMR1L pair of registers as the time base for working in capture or compare mode.

- -In capture mode, the value of the register pair TMR1H:TMR1L is copied to the register pair CCPRxH:CCPRxL when a configuration event occurs.
- -In compare mode, when the value in the CCPRxH:CCPRxL pair of registers matches the value in the TMR1H:TMR1L pair of registers, an event will be triggered. This event can be used to trigger special events.

For more information, please refer to the "capture/compare/PWM mod (CCP1 and CCP" chapter).



# 9.11 ECCP Special Events Flip-flop

If ECCP is configured to trigger a special event, flip-flop will clear the MR1H:TMR1L pair of registers. This special event will not cause TIMER1interrupt. You can still configure ECCPmod to generate an ECCP interrupt.

In this working mode, the CCPRxH: CCPRxL pair register actually becomes the period register of TIMER1.

To use special event flip-flop, you should make TIMER1 and FSYS synchronous.TIMER1 working in asynchronous mode can cause loss of special event trigger signal.

When the operation of writing to TMR1H or TMR1L occurs at the same time as the signal triggered by a special event from ECCP, the writing operation has priority.

For more information, please refer to the "capture/compare/PWMmod (CCP1 and CCP" chapter).

## 9.12 TIMER1 Control Register

	lorregister			-		-	-			
10H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0		
T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N		
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		
reset value	0	0	0	0	0	0	0	0		
Bit7 T1GINV: 1= 0= Bit6 TMR1GE:			TIMER1 gate control signal polarity bit; TIMER1 gate control signal is active high (TIMER1 counts when the gate control signal is high level); The TIMER1 gate control signal is active low (TIMER1 counts when the gate control signal is low). TIMER1 gate control enable bit. If TMR1ON=0, ignore this bit. If TMR1ON=1: 1= TIMER1 counting is controlled by TIMER1gate control function; 0=TIMER1always counts.							
Bit5~Bit4 T1CKPS<1:0>: 11= 10= 01=		11= 1: 10= 1: 01= 1:	IMER1 input c :8; :4; :2; :1.		y ratio selectic					
Bit3	T1	1= E	LP oscillator enable controlbit; Enable LP oscillatoras the clock source of TIMER1; Disable LP oscillator.							
Bit2	Т	Т	MR1CS=1:	1= not synch 0= synchrono	synchronous c ronous with ex ous with exterr TIMER1 uses	ternalclockir nal clock inpu	ıt.			
Bit1	Т	MR1CS: T 1= F tr	TMR1CS=0: ignore this bit, TIMER1 uses internal clock. TIMER1 clock source selection bit; From LP oscillator clock source or clock source from T1CKI pin (rising edge trigger); Internal clock source Fsys.							
Bit0	T	1= E	IMER1enable nable TIMER1 visable TIMER	;						

TIMER1control register T1CON (10H)



# **10. TIMER2**

## 10.1 TIMER2 General

TIMER2 mod is an 8-bit timer/counter with the following characteristics:

- 8-bit timer register (TMR2);
- 8-bit period register (PR2);
- Interrupt when TMR2 matches PR2;
- Software programmable prescaler ratio (1:1, 1:4 and 1:16);
- Software programmable postscaler ratio (1:1 to 1:16).

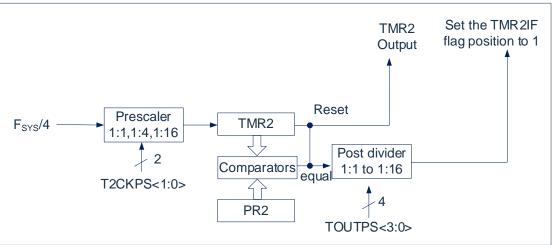


Fig 10-1: TIMER2 structure



## **10.2 Working Principle of TIMER2**

The input clock of the TIMER2 mod is the system instruction clock (FSYS/4). The clock is input to the TIMER2 pre-scaler. There are several division ratios to choose from: 1:1, 1:4 or 1:16. pre-scaler the output is then used to increment TMR2register.

Continue to compare the values of TMR2 and PR2 to determine when they match. TMR2 will increase from 00h until it matches the value in PR2. When a match occurs, the following two events will occur:

- TMR2 is reset to 00h in the next increment period;
- TIMER2 post-scaler increments.

The matching output of the TIMER2 and PR2 comparator is then input to the post-scaler of TIMER2. The post-scaler has a prescaler ratio of 1:1 to 1:16 to choose from. The output of the TIMER2 post-scaler is used to make PIR1 The TMR2IF interrupt flag bit of the register is set to 1.

Both TMR2 and PR2 registers can be read and written. At any reset, TMR2 register is set to 00h and PR2 register is set to FFh.

Enable TIMER2 by setting the TMR2ON bit of the T2CON register; disable TIMER2 by clearing the TMR2ON bit.

The TIMER2 pre-scaler is controlled by the T2CKPS bit of the T2CON register; the TIMER2 postscaler is controlled by the TOUTPS bit of the T2CON register.

The pre-scaler and postscaler counters are cleared under the following conditions:

- When TMR2ON=0
- Any device reset occurs (power-on reset, watchdog timer reset, or undervoltage reset).

Note: Writing T2CON will not clear TMR2. When TMR2ON=0, the TMR2 register cannot be written.



# 10.3 TIMER2 related register

There are two registers related to TIMER2, namely data memory TMR2 and control register T2CON.

### TIMER2 data register TMR2 (11H)

11H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
TMR2								
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	Х	Х	Х	Х	Х	Х	Х	Х

### TIMER2 control register T2CON (12H)

12H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
T2CON		TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0
Read write		R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value		0	0	0	0	0	0	0

Bit7	Not used, read 0.	
Bit6~Bit3	TOUTPS<3:0>:	TIMER2 output frequency division ratio selection bit.
Dito Dito	0000=	1:1;
	0001=	1:2;
	0010=	1:3;
	0011=	1:4:
	0100=	
	0101=	
	0110=	
	0111=	,
	1000=	
	1001=	
	1010=	
	1011=	1:12;
	1100=	1:13;
	1101=	1:14;
	1110=	1:15;
	1111=	1:16.
Bit2	TMR2ON:	TIMER2 enable bit;
	1=	Enable TIMER2;
	0=	Disable TIMER2.
Bit1~Bit0	T2CKPS<1:0>:	TIMER2 clock frequency division ratio selection bit;
	00=	1;
	01=	4;
	1 x=	16.



# 11. Analog to Digital Conversion (ADC)

# 11.1 ADC general

The analog-to-digital converter (ADC) can convert the analog input signal into a 12-bit binary number that represents the signal. The analog input channels used by the device share a sample and hold circuit. The output of the sample and hold circuit is connected to the input of the analog to digital converter. The analog-to-digital converter uses the successive approximation method to generate a 12-bit binary result, and save the result in the ADC result register (ADRESL and ADRESH).

ADC reference voltage is always generated internally. ADC can generate an interrupt after conversion is completed.

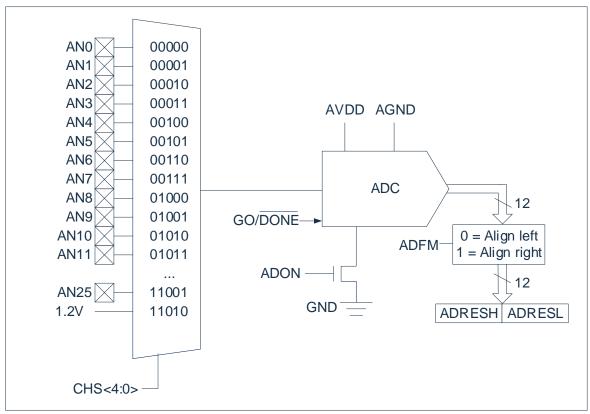


Fig 11-1: ADC structure



## **11.2 ADC configuration**

When configuring and using ADC, the following factors must be considered:

- Port configuration;
- Channel selection;
- ♦ ADC conversion clock source;
- Interrupt control;
- The storage format of the result.

## 11.2.1 Port configuration

ADC can convert both analog signal and digital signal. When converting analog signal, the I/O pin should be configured as analog input pin by setting the corresponding TRIS bit to 1. For more information, please refer to the corresponding port chapter.

Note: Applying analog voltage to pins defined as digital inputs may cause overcurrent in the input buffer.

## 11.2.2 Channel selection

The CHS bit of the ADCON0 register determines which channel is connected to the sample and hold circuit.

If the channel is changed, a certain delay will be required before the next conversion starts. For more information, please refer to the "ADC working principle" chapter.

## 11.2.3 ADC reference voltage

The ADC reference voltage is always provided by the chip's VDD and GND.



### 11.2.4 Converter clock

•

The ADCS bit of the ADCON0 register can be set by software to select the clock source for conversion. There are 4 possible clock frequencies to choose from:

- Fsys/8
   Fsys/32
  - Fsys/16 F<sub>RC</sub> (special internal oscillator)

The time to complete one-bit conversion is defined as TAD. A complete 12-bit conversion requires 49 TAD periods.

Must comply with the corresponding TAD specification to get the correct conversion result. The following table is an example of correct selection of ADC clock.

Note: Unless FRC is used, any change in the system clock frequency will change the ADC clock frequency, which will negatively affect the ADC conversion results.

For different reference voltages and different VDDs, you need to refer to the following table to set a reasonable frequency division.

Reference	Marking voltage ()/)	Fastest divi	Fastest division setting			
voltage	Working voltage (V)	F <sub>SYS</sub> = 16MHz	Fsys = 8MHz	Conversion time (us)		
VDD	4.0~5.5	Fsys/16	Fsys/8	49		
VDD	2.7~4.0	Fsys/32	Fsys/16	98		

### 11.2.5 ADC Interrupt

ADC mod allows an interrupt to be generated after the completion of the analog-to-digital conversion. The ADC interrupt flag bit is the ADIF bit in PIR1register. The ADC interrupt enable bit is the ADIE bit in PIE1register. The ADIF bit must be cleared by software. The ADIF bit after each conversion is completed Will be set to 1, regardless of whether ADC interrupt is allowed.

## 11.2.6 Output Formatting

The result of 12-bit A/D conversion can be in two formats: left-justified or right-justified. The output format is controlled by the ADFM bit in ADCON1register.

When ADFM=0, the AD conversion result is left aligned and the AD conversion result is 12Bit; when ADFM=1, the AD conversion result is right aligned, and the AD conversion result is 10 Bit.



## **11.3 ADC working principle**

### 11.3.1 Start conversion

To enable ADC mod, you must set the ADON bit of the ADCON0 register to 1, and set the GO/ ("DONE")<sup>–</sup> bit of the ADCON0 register to 1 to start analog-to-digital conversion.

Note: It is not possible to set GO/DONE position to 1 with the same instructions that open A/Dmod.

### 11.3.2 Complete conversion

When the conversion is complete, the ADC mod will:

- Clear the GO/DONE bit;
- Set ADIF flag bit to 1;
- Update the ADRESH: ADRESL register with the new conversion result.

### 11.3.3 Stop conversion

If you must terminate the conversion before conversion is completed, you can use software to clear the GO/DONEbit. The ADRESH: ADRESL register will not be updated with the uncompleted analog-to-digital conversion result. Therefore, the ADRESH: ADRESL register will remain on the value obtained by the second conversion. In addition, after the A/D conversion is terminated, a delay of 2 TAD must be passed before the next acquisition can be started. After the delay, the input signal of the selected channel will automatically start to be collected.

Note: Device reset will force all registers to enter the reset state. Therefore, reset will close the ADC mod and terminate any pending conversions.

### 11.3.4 Working principle of ADC in sleep mode

ADC mod can work in sleep mode. This operation requires ADC clock source to be set to FRC option. If FRC clock source is selected, ADC must wait for one more instruction speriod before starting conversion. This allows the execution of STOP instructions to reduce conversion If the ADC interrupt is allowed, the device will wake up from sleep mode when the conversion ends. If the ADC interrupt is disabled, even if the ADON bit remains set, the ADC mod will be closed after the conversion is complete. If the ADC clock source is not FRC, even if the ADON bit remains set, executing the STOP instructions will abort the current conversion and close the A/D mod.



### 11.3.5 A/D conversion procedure

The following steps give an example of using ADC for analog-to-digital conversion:

- 1. port configuration:
  - Configure pin as input pin (see TRIS register).
- 2. configuration ADC mod:
  - Select ADC conversion clock;
  - Select ADC input channel;
  - Choose the format of the result;
  - Start the ADC mod.
- 3. configuration ADC interrupt (optional):
  - Clear ADC interrupt flag bit;
  - Allow ADC interrupt;
  - Allow peripherals interrupt;
  - Allow global interrupt.
- 4. Wait for the required acquisition time.
- 5. Set GO/DONE to 1 to start conversion.
- 6. Wait for the ADC conversion to end by one of the following methods:
  - Query GO/ ("DONE") bit
  - Wait for ADC interrupt (allow interrupt).
- 7. Read ADC results.
- 8. Clear the ADC interrupt flag bit (if interrupt is allowed, this operation is required).

Note: If the user tries to resume sequential code execution after waking the device from sleep mode, the global interrupt must be disabled.

LDIA	B'1000000'	
LD	ADCON1,A	
SETB	TRISA,0	;set PORTA.0 as input
LDIA	B'11000001'	
LD	ADCON0,A	
CALL	DELAY	;delay
SETB	ADCON0,GO	
SZB	ADCON0,GO	;wait ADCto complete
JP	<b>\$-1</b>	
LD	A,ADRESH	;save the highest bit of ADC
LD	RESULTH,A	
LD	A,ADRESL	; save the lowest bit of ADC
LD	RESULTL,A	

#### example: AD conversion



# 11.4 ADC Related Register

There are mainly 4 RAMs related to AD conversion, namely control register ADCON0 and ADCON1, data register ADRESH and ADRESL.

9EH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ADCON0	ADCS1	ADCS0	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON
Read/write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0
Bit7~Bit6	ADCS<1:0>	: A/D conv	version clock	selection bit.				
	00:	= F <sub>SYS</sub> /8						
	01:	= F <sub>SYS</sub> /16						
	10:	= Fsys/32						
	11:	= Frc (A d	edicated inte	rnal oscillato	r generates a	clock with a	a frequency of	up to 32KHz)
Bit5~Bit2	CHS<3:0>			the analog cl	hannel select	tion bit and C	CHS4 form a fi	ve-bit channel
		selection						
	CHS<4:0>							
	00000:							
	00001:							
	00010							
	00011: 00100:							
	00100							
	00101							
	00111:							
	01000:							
		= AN9						
	01010:	= AN10						
	01011:	= AN11						
	10111:	= AN23						
	11000:	= AN24						
	11001:	= AN25						
		= 1.2V (fixe						
Bit1		: A/D conv						
	1:			progress. Set ted, this bit is			onversion. Wł hardware.	nen A/D
	0:	= A/D conv	version comp	lete or not in	progress.			
Bit0	ADON	I: ADC ena	ble bit.					
	1:	= Enable A						
	0:	= Disable /	ADC, not co	nsuming curr	ent.			

### AD control register ADCON0 (9EH)



#### AD data register high bit ADCON1 (9FH)

9FH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ADCON1	ADFM	CHS4						
Read/write	R/W	R/W						
Reset value	0	0						

Bit7

ADFM: A/D conversion result format selection bit

1= Right alignment

0= left alignment

Bit6 CHS4: Channel selection bit

Bit5~Bit0 Not used

### AD data register high bit ADRESH (9DH), ADFM=0

9DH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ADRESH	ADRES11	ADRES10	ADRES9	ADRES8	ADRES7	ADRES6	ADRES5	ADRES4
read/write	R	R	R	R	R	R	R	R
Reset value	Х	Х	Х	Х	Х	Х	Х	Х

Bit7~Bit0

ADRES<11:4>: ADC result register bit.

The higher 8 bits of the 12-bit conversion result.

#### AD data register lower bit ADRESL (9CH), ADFM=0

-			( ):					
9CH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ADRESL	ADRES3	ADRES2	ADRES1	ADRES0				
read/write	R	R	R	R				
Reset value	Х	Х	Х	Х				

Bit7~Bit4 ADRES<3:0>: ADC result register bit.

The lower 4 bits of the 12-bit conversion result.

Bit3~Bit0

Not used

#### AD data registerhigh bit ADRESH (9DH), ADFM=1

9DH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ADRESH							ADRES11	ADRES10
read/write							R	R
Reset value							Х	Х

Bit7~Bit2 Not used.

Bit1~Bit0

ADRES<11:10>: ADC result register bit.

The higher 2 bits of the 12-bit conversion result.



### AD data register lower bits ADRESL (9CH), ADFM=1

9CH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
ADRESL	ADRES9	ADRES8	ADRES7	ADRES6	ADRES5	ADRES4	ADRES3	ADRES2
read/write	R	R	R	R	R	R	R	R
Reset value	Х	Х	х	Х	Х	Х	Х	Х

Bit7~Bit0

ADRES<9:2>: ADC result register bit.

The 2-9 bits of the 12-bit conversion result.

Note: In the case of ADFM=1, the AD conversion result only saves the upper 10 bits of the 12-bit result, where ADRESH saves the upper 2 bits, and ADRESL saves the 2nd to 9th digits.



# 12. LCD/LED Driver Mod

Chip built-in LCD/LED driver mod, they share the control register.

The LED driver mod can support up to driver 11\*8 digital tubes. The program only needs to set the relevant control bits and display data, and the chip pins will automatically output the driver LED waveform (hardware driver).

The LCD driver mod can be driver 1/2 Bias LCD, and the waveform of the output driver LCD must be controlled by a program (software driver).

## 12.1 LCD/LED Function Enable

Set the 7th bit LCDEN of LEDCON0 (113H) to 1, and clear the 6th bit LEDEN to 0 to enable the LCD driver function;

Set the 6th bit LEDEN of LED CON0 (113H) to 1, and clear the 7th bit LCDEN to 0 to enable the LED driver function;

Set both LCDEN and LEDEN to 0, turn off the LCD/LED mod.

Note: Please do not set LCDEN and LEDEN to 1 at the same time.

## **12.2 LCD/LED Pin Function Configuration**

If the LCD driver mod is enabled and the COM port function is enabled, the corresponding I/O port will be forced to be an analog input state, regardless of the state of the corresponding TRIS bit.

If the LED driver function is enabled, the corresponding SEG port and COM port must be set to output state, and output "0", that is, the corresponding TRIS bit and PORT position are "0".



# 12.3 LCD/LED COM Port Configuration

The setting method of LED COM port is as follows:

Set the I/O port direction and data register. The LCD function sets the corresponding pin to input state, and the LED function sets the corresponding pin to output state and output low level.

COMSEL [2:0]	No. of COM port (only LED)
000	4
001	5
010	6
011	8
100	2
101	3
110	3
111	4

Set the COM\_EN register, and set the corresponding pin as the COM port of the LED function.

If the user's COM ports are not arranged in order during use. For example, if COM3-COM6 is used as the COM port for LED function, and COM0-COM2 is used as ordinary I/O port, you can do the following settings:

- Set the number of COM ports to 8 COM, COMSEL="011";
- Set COM3-COM6 of the COM\_EN register to 1, COM0-COM2, COM7 to 0.

At this time, COM3-COM6 is the COM port of LED function, and its output duty cycle is 1/8. And COM0-COM2, COM7 can be used as ordinary I/O ports.

# 12.4 LCD/LED SEG Port Configuration

The SEG port that enables the LCD/LED function must meet the following conditions:

- 1. Set the status of the corresponding pin, the LED function sets the corresponding pin to the output state and output "0";
- 2. Set the corresponding pins in SEGEN0, SEGEN1, and SEGEN2register as LED driver functions;
- 3. Set SEG port output current in SEGEN2 register (LED function only).



# 12.5 LED Data Configuration

Setting the LED display data requires the following steps:

- 1. Set the SEGOUT [1:0] bits of the LEDCON1 register to "1x";
- 2. Set the 7th bit of LEDADD register LEDCS=1 to allow read/write data;
- 3. Set the data address 0-6 of LEDADD;
- 4. Set LEDDATA data (the pin is not used for LED function, and the corresponding LEDDATA bit needs to be set to "0");
- 5. Repeat steps 3-4 to set other address data;
- 6. Turn off the data read/write bit LEDCS=0 after setting.

Relationship of LED address and data:

LEDADD				LEDI	DATA				
00H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SEG0
01H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SEG1
16H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SEG22
17H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	SEG23
	COM7	COM6	COM5	COM4	COM3	COM2	COM1	COM0	-

# 12.6 LCD/LED Related Register

LCD/LED driver related register: control register LEDCON0、LEDCON1; address register LEDADD; data register LEDDATA; register COMEN、SEGEN0、SEGEN1、SEGEN2.LCD/LED control register LEDCON0 (113H)

113H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
LEDCON0	LCDEN	LEDEN	COMS	SEL [1:0]	LEDCLK [3:0]				
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
Reset value	0	0	0	0	0	0	0	0	

Bit7	LCDEN:	LCD mod enab	ble bit;
	0:	Disable LCD n	nod;
	1:	Enable LCD m	iod.
Bit6	LEDEN:	LED mod enat	ble bit;
	0:	Disable LEDm	od;
	1:	Enable LEDmo	od.
Bit5~Bit4	COMSEL [1:0]:	Number of LE	D mod COM port selection (COMSEL[2] in LEDADD register
		5 <sup>th</sup> bit);	
	000:	4COM;	
	001:	5COM;	
	010:	6COM;	
	011:	8COM;	
	100:	2COM;	
	101:	3COM;	
	110:	3COM;	
	111:	4COM.	
Bit3~Bit0	LEDCLK [3:0]:	LED frequency	selection (Select the clock source according to the 6th LEDF
		of LEDCON1);	· · · ·
		LEDF=0	LEDF=1
	0000:	F <sub>HSI</sub> /64	not used;
	0001:	F <sub>HSI</sub> /128	not used;
	0010:	F <sub>HSI</sub> /256	not used;
	0011:	Fнsi /512	External 32.768KHz oscillation/2;
	0100:	F <sub>HSI</sub> /1024	External 32.768KHz oscillation/4;
	0101:	F <sub>HSI</sub> /2048	External 32.768KHz oscillation/8;
	0110:	F <sub>HSI</sub> /4096	External 32.768KHz oscillation/16;
	0111:	F <sub>HSI</sub> /8192	External 32.768KHz oscillation/32;
	1x00:	F <sub>HSI</sub> /16384	External 32.768KHz oscillation/64;
	1x01:	F <sub>HSI</sub> /32768	External 32.768KHz oscillation/128;
	1x10:	F <sub>HSI</sub> /65536	External 32.768KHz oscillation/256;
	1x11:	F <sub>HSI</sub> /131072	External 32.768KHz oscillation/512.



114H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
LEDCON1		LEDF	SEGO	UT[1:0]			LCDISEL[1:0]		
R/W		R/W	R/W	R/W			R/W	R/W	
Reset value		0	0	0			0	0	
Bit7		not used.							
Bit6		LEDF:	LED clock	source selee	ction;				
		0: Internal clock;							
		1:	external32	2.768KHz oso	cillation cloc	xk.			
			(Need to	connect a 32	2.768KHz ci	rystal oscillator	to the OSCI/OS	SCO pin of	
		the chip)							
Bit5~Bit4		SEGOUT:	SEG outp	ut mode sele	ction;				
		00:	SEG outp	ut all 0;					
		01:	SEG outp	ut all 1;					
		1x:	SEG outp	ut as data fro	m LEDDAT	A.			
Bit3~Bit2		not used.							
Bit1~Bit0		LCDISEL[1:0]:	LCD outpu	ut current sel	ection bit;	(only for LCD)			
		00=	100Ua@5	V;					
		01=	200Ua@5	V;					
		10=	400Ua@5	V;					
		11=	800Ua@5	V.					

### LCD control register LEDCON1 (114H)

## LED address register LEDADD (115H)

115H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	
LEDADD	LEDCS		COMSEL[2]	LEDADD[4:0]					
R/W	R/W		R/W	R/W	R/W	R/W	R/W	R/W	
Reset value	0		0	0	0	0	0	0	

Bit7	LEDCS:	LED data read/write enable bit;
	0:	Disable read/write LED data;
	1:	Enable read/write LED data.
Bit6	not used	
Bit5	COMSEL[2]:	COMselect the highest bit, COMSEL[2:0] set number of COM ports and Bias;
Bit4~Bit0	LEDADD[4:0]:	LED address selection;
		LED address range 00H-0AH

## LED data register LEDDATA (116H)

116H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0			
LEDDATA		LEDDATA[7:0]									
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
reset value	Х	Х	Х	Х	Х	Х	Х	Х			

Bit7~Bit0 LEDDATA[7:0]: LED data setting, write the data corresponding to the address of LEDADD.



### LCD/LED COM port control register COMEN (11AH)

11AH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
COMEN	COM7EN	COM6EN	COM5EN	COM4EN	<b>COM3EN</b>	COM2EN	COM1EN	COM0EN
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0

Bit7~Bit0

COMxEN: COM port function configuration ;

- 0: COMx as normal I/O (x=0-7);
- 1: COMx as COM (x=0-7) ofLCD/LED

#### LED SEG port control register SEGEN0 (119H)

119H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
SEGEN0	SEG7EN	SEG6EN	SEG5EN	SEG4EN	SEG3EN	SEG2EN	SEG1EN	SEG0EN
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0

Bit7~Bit0

- SEGxEN: SEG port function configuration ;
  - 0: SEGx as normal I/O (x=0-7);
  - 1: SEGx as COM (x=0-7) ofLCD/LED

### LED SEG port control register SEGEN1 (118H)

118H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
SEGEN1						SEG10EN	SEG9EN	SEG8EN
R/W						R/W	R/W	R/W
Reset value						0	0	0

Bit7~Bit3 not used

Bit2~Bit0

SEGxEN: SEG port function configuration ;

- 0: SEGx as normal I/O (x=8-10);
- 1: SEGx as COM (x=8-10) of LCD/LED

### LED SEG port control register SEGEN2 (117H)

117H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
SEGEN2	SEGDRI[3:0]							
R/W	R/W	R/W	R/W	R/W				
Reset value	0	0	0	0				

Bit7~Bit4	SEGDRI[3:0]:	SEG port driver current configuration;	(仅 LED 功能有效)	
	0000:	SEG port driver current as 0;		
	0001:	SEG port driver current as 2mA;		
	0010:	SEG port driver current as 4mA;		
	0011:	SEG port driver current as 6mA;		
	1110:	SEG port driver current as 28mA;		
	1111:	SEG port driver current as 30mA.		
Bit3~Bit0	not used			
		05 / 405		•

# 13. Capture/Compare/PWM Mod (CCP1 and CCP2)

The chip contains two capture/compare/PWM (CCP1) and (CCP2). The operations of CCP1 and CCP2 mod are basically the same.

Note: CCPRx and CCPx in this document refer to CCPR1 or CCPR2 and CCP1 or CCP2 respectively.

The capture/compare/PWM mod is peripherals that allow users to time and control different events. In capture mode, the peripherals can time the duration of the event. Capture mode allows the user to trigger an external event after the predetermined timing expires. PWM mode can generate pulse width modulation signal with variable frequency and duty cycle.

When CCP is used in capture/compare mode, timer TIMER1 is required.

CCPx control register CCPxCON

	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
CCPxCON			DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0
read/write			R/W	R/W	R/W	R/W	R/W	R/W
Reset value			0	0	0	0	0	0

Bit7~Bit6	not used.	
Bit5~Bit4	DC1B<1:0>:	Lower 2 bits of PWM duty cycle;
	capturemode:	not used;
	comparemode:	not used;
	PWM mode:	These two bits are the lower 2 bits of the 10-bit PWM duty cycle. The higher
		8 bits of the duty cycle are in CCPR1L.
Bit3~Bit0	CCP1M<3:0>:	CCP mode selection bit;
	=0000	capture/compare/PWM off (reset ECCP mod);
	0001=	not used (stay);
	0010=	comparemode, output level turns when matching (CCP1IF set 1);
	0011=	not used (stay);
	0100=	Capture mode, Capture occurs on every falling edge;
	0101=	Capture mode, Capture occurs on every rising edge;
	0110=	Capture mode, Capture occurs for every 4 rising edges;
	0111=	Capture mode, Capture occurs for every 16 rising edges;
	1000=	Compare mode, Output is high when compare matches (CCP1IF set to 1);
	1001=	Compare mode, Output is low when compare matches (CCP1IF set to 1);
	1010=	Compare mode, generate interrupt when compare matches (CCP1IF set 1,
		CCP1 pin not affected);
	1011=	Compare mode, Trigger special events (CCP1IF bit set as 1, CCP1 reset
		TMR1 or TMR2);
	11xx=	PWM mode.



## 13.1 Capture Mode

In capture mode, when an event occurs on the corresponding CCPx pin, CCPRxH: CCPRxL is the 16bit value of the register capture TMR1 register. The event that triggers the capture can be defined as one of the following four, and is determined by the CCPxM in the CCPxCON register<3:0> bit configuration:

- Every falling edge;
- Every rising edge;
- Every 4 rising edges;
- Every 16 rising edges.

The event type is selected by the mode selection bit CCPxM3:CCPxM0 (CCPxCON<3:0>). When a capture occurs, the interrupt request flag bit CCPxIF in the PIRx register is set to 1; it must be cleared by software. If another capture occurs before the value in the register CCPRxH and CCPRxL is read, then the previous capture value will be overwritten by the new capture value (see Figure 13-1).

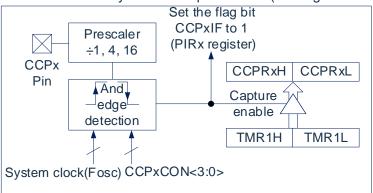


Fig 13-1: capture mode working structure

## 13.1.1 CCP Pin Configuration

In capture mode, the corresponding CCPx pin should be configured as input by setting the corresponding TRIS control bit to 1.

Note: If the CCPx pin is configured as output, a write operation to the port may trigger a capture event.

## 13.1.2 TIMER1 Mode Selection

TIMER1 must run in timer mode or synchronous counter mode CCP mod to use the capture function. Capture operation cannot be performed in asynchronous counter mode.

## 13.1.3 Software interrupt

When the capture mode is changed, a false capture interrupt may occur. The user should keep the CCPxIE interrupt enable bit in the PIEx register cleared to avoid false interrupts. The interrupt flag bit CCPxIF in the PIRx register should also be cleared after any change in the operation mode.



## 13.1.4 CCP pre-scaler

The CCPxM<3:0> bits in the CCPxCON register specify four pre-scaler settings. Whenever the CCP mod is closed or the capture mode is disabled, the pre-scaler counter will be cleared. This means that any reset will clear the pre-scaler. Divide counter.

Switching from one capture prescaler ratio to another capture prescaler ratio will not clear the prescaler counter, but may cause false interrupts. To avoid this unexpected operation, you should change the prescaler Turn off the mod by clearing the CCPxCON register before frequency ratio.

Change the capture prescaler ratio

CLR	CCP1CON	; turn off CCP1
LDIA	B'00000101'	
LD	CCP1CON,A	; pass new value to CCP1



# 13.2 Compare Mode

In compare mode, the value of the 16-bit CCPRx register will be constantly compared with the value of a pair of TMR1 registers. When the two match, the following situations may occur in the CCPx mod:

- CCPx output high level;
- CCPx output low level;
- Generate special events to trigger signal;
- Generate software interrupt

The action on the pin depends on the value of the CCPxM<3:0> control bits in the CCPxCON register, and all capture modes will generate interrupt.

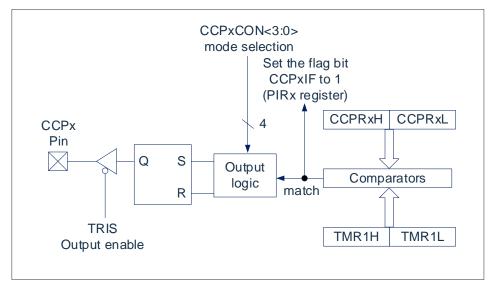


Fig 13-2: compare mode working structure

Special events trigger signal will:

- Clear the TMR1H and TMR1L registers.
- Will not set the TMR1IF flag bit in the PIR1 register.
- Make GO/ ("DONE") bit 1 to start ADC conversion.

## 13.2.1 CCP Pin Configuration

The user must configure the CCPx pin to output by clearing the corresponding TRIS bit.

Note: Clearing the CCPxCON register will force the CCPx compare output latch to the default low level, which is not a port I/O data latch.

### 13.2.2 TIMER1 Mode Selection

In compare mode, TIMER1 must run in timer mode or synchronous counter mode. In asynchronous counter mode, compare operations may not be possible.

### 13.2.3 Software Interrupt Mode

When the software interrupt mode is selected (CCPxM<3:0>=1010), the CCPx mod will not control the CCPx pin (see CCPxCON register).





## 13.2.4 Special Events Trigger Signal

When the special event trigger mode (CCPxM<3:0>=1011) is selected, the CCPx mod will complete the following operations:

- reset TIMER1;
- If ADC is enabled, ADC conversion will also be started.

In this mode, the CCPx mod does not control the CCPx pin (see CCPxCONregister).

When the TMR1H/TMR1L register pair matches the CCPRxH/CCPRxL register pair, the CCP will immediately generate a special event trigger output. The TMR1H/TMR1L register pair will not reset until the next rising edge of the TIMER1 clock. Therefore, the CCPRxH/CCPRxL register pair actually become a 16-bit programmable period register of TIMER1.

### Note:

- 1) The signal triggered by a special event from the CCP mod will not set the TMRxIF interrupt flag bit in the PIR1 register to be 1.
- 2) Change the contents of the CCPRxH and CCPRxL register pair between the edge that triggers the signal when a special event is generated and the clock edge that causes TIMER1 reset to clear the matching condition and prevent reset from occurring.



# 13.3 PWM Mode

PWM mode generates pulse width modulation signal on CCPxpin. Both PWM1 and PWM2 have their own independent period counters. The duty cycle, period and resolution are determined by the following registers:

- PWMCONPWMxCYC
- CCPRxL
   CCPxCON

### PWM control register PWMCON (99H)

	0		,					
99H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PWMCON		CYC2EN	CK2[1:0]			CYC1EN	CK1[1:0]	
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0

Bit7	not used.	
Bit6	CYC2EN:	PWM2 period counter enable bi.
	1=	enable.
	0=	disable.
Bit5-Bit4	CK2[1:0]:	PWM2 period counter clock prescale selection bit.
	00=	Prescale 1.
	01=	Prescale 4.
	1X=	Prescale 16.
Bit3	not used.	
Dive		
Bit2	CYC1EN:	PWM1 period counter enable bit.
Bit2	CYC1EN: 1=	PWM1 period counter enable bit. enable.
Bit2		•
Bit2	1= 0=	enable.
	1= 0=	enable. disable.
	1= 0= CK1[1:0]: 00=	enable. disable. PWM1 period counter clock prescale selection bit.
	1= 0= CK1[1:0]: 00= 01=	enable. disable. PWM1 period counter clock prescale selection bit. Prescale 1.

### PWM1 period data register PWM1CYC (9AH)

9AH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PWM1CYC								
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	1	1	1	1	1	1	1	1

### PWM2 period data register PWM2CYC (9BH)

9BH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
PWM2CYC								
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	1	1	1	1	1	1	1	1



In pulse width modulation (PWM) mode, CCP mod can output a PWM signal with a resolution of up to 10 bits on the CCPx pin. Since the CCPx pin is multiplexed with the port data latch, the corresponding TRIS bit must be cleared to enable CCPx the output driver of the pin.

Note: Clearing the CCPxCON register will give up CCPx's control over the CCPx pin.

The following Figure 13-3 is a simplified block diagram of PWM operation, and Figure 13-4 is a typical waveform of the PWM signal.

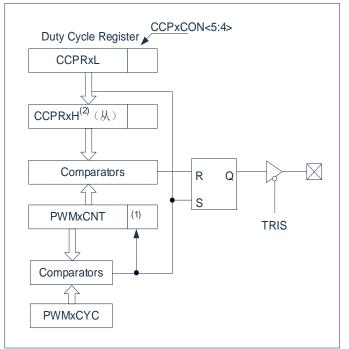


Fig 13-3: simplified PWM block diagram

- Note:
   1) The value of the 8-bit timer PWMxCNT register is combined with a 2-bit internal system clock (F<sub>SYS</sub>) or 2 bits of the pre-scaler to generate a 10-bit time base.
- 2) In PWM mode, CCPRxH is a read-only register.

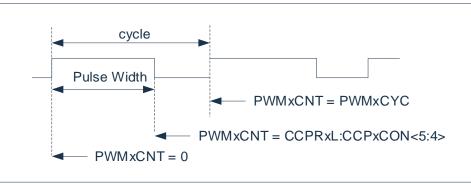


Fig 13-4: CCP PWM output



### 13.3.1 PWM Period

The PWM period is specified by writing the PWMxCYC register of PWMxCNT.

Formula 1: PWM period:

PWM period=[(PWMxCYC)+1]\*4\*Tosc\*(PWMxCNT prescaler value)

Note: Tosc=1/Fsys

When PWMxCNT is equal to PWMxCYC, the following three events will occur in the next up-counting period:

- PWMxCNT is cleared;
- CCPx pin is set to 1 (exception: if PWM duty cycle=0%, CCPx pin will not be set to 1);
- The PWM duty cycle is latched from CCPRxL to CCPRxH.

## 13.3.2 PWM Duty Cycle

The PWM duty cycle can be specified by writing a 10-bit value to the following multiple registers: DCxB<1:0> bits of the CCPRxL register and CCPxCON register. CCPRxL stores the upper 8 bits of the duty cycle, and DCxB<1 of the CCPxCON register: The 0> bit saves the lower 2 bits of the duty cycle. You can write the DCxB<1:0> bits of the CCPRxL and CCPxCON register at any time, but until the values in PWMxCYC and PWMxCNT match (that is, the period ends), the value of the duty cycle It is latched into CCPRxH. In PWM mode, CCPRxH is a read-only register.

Formula 2: Pulse width calculation formula:

pulse width=(CCPRxL:CCPxCON<5:4>)\*Tosc\*(PWMxCNT prescaler value)

Formula 3: PWM duty cycle calculation formula: duty cycle=  $\frac{(CCPRxL:CCPxCON<5:4>)}{4 (PWMxCYC+1)}$ 

The CCPRxH register and a 2-bit internal latch are used to provide double buffering for the PWM duty cycle. This double buffering structure is extremely important to avoid glitches during the PWM operation.

The value of the 8-bit timer PWMxCNT register is combined with a 2-bit internal system clock (Fsys) or 2 bits of the pre-scaler to generate a 10-bit time base. The system clock is used when the PWMxCNT prescaler ratio is 1:1.

When the 10-bit time base matches the combined value of CCPRxH and 2-bit latch, the CCPx pin is cleared (see Figure 13-3).



### 13.3.3 **PWM Resolution**

Resolution determines the number of duty cycles within a given period. For example, a 10-bit resolution will generate 1024 discrete duty cycles, and an 8-bit resolution will generate 256 discrete duty cycles.

When PWMxCYC is 255, the maximum resolution of PWM is 10 bits. As shown in formula 4, resolution is a function of the value of the PWMxCYC register.

formula4: PWM resolution:

resolution=
$$\frac{log[4 (PWMxCYC+1)]}{log (2)}$$

Note: If the pulse width is greater than the period value, the specified PWM pin will remain unchanged.

The following table shows the value of PWM frequency and resolution when FSYS=8MHz.

PWM frequency	1.22KHz	4.90KHz	19.61KHz	76.92KHz	153.85KHz	200.0KHz				
Timer prescaler (1,4 or16)	16	4	1	1	1	1				
PWMxCYC	0x65	0x65	0x65	0x19	0x0C	0x09				
Largest resolution (bit)	8	8	8	6	5	5				

PWM frequency and resolution example (Fsys=8MHz)

### 13.3.4 Operations Under Sleep Mode

In sleep mode, the PWMxCNT register will not increment and the mod state will remain unchanged. If the CCPx pin has an output, it will continue to maintain the output value unchanged. When the device is woken up, PWMxCNT will continue to work from the original state.

### 13.3.5 System Clock Frequency Changes

The PWM frequency is generated by the systemclock frequency. Any change in the system clock frequency will change the PWM frequency.

### 13.3.6 Effect of Reset

Any reset will force all ports to be input mode and force the CCP register to enter its reset state.



## 13.3.7 Configurate PWM

The following steps should be performed when configuring CCP mod to PWM operation mode:

- 1. Disable the output driver of PWMpin (CCPx) by setting the corresponding TRIS bit to 1 to make it an input pin.
- 2. Set the PWM period by loading the PWMxCYC register.
- 3. By loading the CCPxCON register configuration CCP mod PWM mode with appropriate values.
- 4. Set the PWM duty cycle by loading the DCxB<1:0> bits in the CCPRxL register and CCPxCON register.
- 5. Configure and start the PWMxCNT period counter:
  - Clear the TMR2IF interrupt flag bit in the PIR1 register.
  - Set the PWMxCNT prescaler ratio by loading the CK2 or CK1 bit of the PWMCON register.
  - Enable PWMxCNT by setting the CYC2EN and CYC1EN bits in the PWMCON register to 1.
- 6. After the new PWM period starts, enable PWM output:
  - Wait for PWMxCNT overflow.
  - Enable the CCPx pin output driver by clearing the corresponding TRIS bit.

# 14. Universal Synchronous/Asynchronous Transmitter (USART0/USART1)

The universal synchronous/asynchronous transmitter (USART) mod is a serial I/O communication peripheral. This mod includes all the clock generators, shift registers and data buffers necessary to perform input or output serial data transmissions that are not related to device program execution. USART It can also be called a serial communication interface (Serial Communications Interface, SCI), it can be configured as a duplex asynchronous system that can communicate with peripherals such as CRT terminals and personal computers; it can also be configured as an integrated circuit with A/D or D/A, Serial EEPROM and other peripherals or half-duplex synchronous system of other microcontroller communication. The microcontroller with which it communicates usually does not have an internal clock that generates baud rate, it needs a master control synchronous device to provide an external clock signal.

The functions of USART0 and USART1 are exactly the same. The following description is based on USART0 mod.

The USART mod includes the following functions:

- Duplex asynchronous transmit and receive
- Single character output buffer
- Double character input buffer
- Frame error detection from receive to character
- Half-duplex synchronous slave mode

- Character length can be programmed to 8 or 9 bits
- Input buffer overflow error detection
- Half-duplex synchronousmaster controlmode
- In synchronous mode, programmable clock polarity

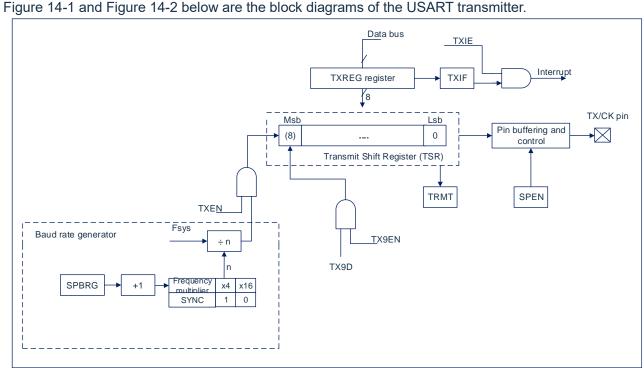


Fig 14-1: USART transmit block diagram



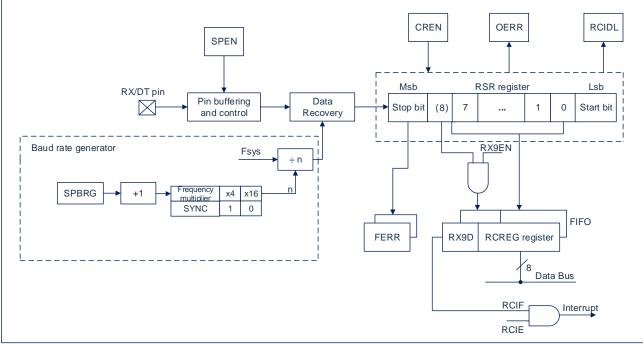


Fig 14-2: USART receive block diagram

The operation of the USART mod is controlled by 3 registers:

- transmit status and control register (TXSTA)
- Receive status and control register (RCSTA)



## 14.1 USART Asynchronous Mode

USART uses the standard non-return-to-zero (NRZ) format for transmit and receive data. Two levels are used to implement NRZ:

It represents the VOH mark state (mark state) of 1data bit, and the VOL space state (space state) of 0 data bit. When using NRZ format to continuously transmit data bits of the same value, the output level will maintain the level of the bit, and It will return the mid-level value after each bit is transmitted. NRZ transmit port is idle in the mark state. The character of each transmit includes a start bit, followed by 8 or 9 data bits and one or more terminations the stop bit of character transmit. The start bit is always in the space state, and the stop bit is always in the mark state. The most commonly used data format is 8 bits. The duration of each transmit bit is 1/ (baud rate). On-chip dedicated 8 Bit/16-bit baud rate generator can be used to generate standard baud rate frequency through system oscillator.

USART first transmit and receive LSb. USART's transmitter and receiver are functionally independent, but use the same data format and baud rate. Hardware does not support parity check, but it can be implemented by software (parity bit is the first 9 data bits).

## 14.1.1 USART Asynchronous Generator

Figure 15-1 shows the block diagram of the USART transmit device. The core of the transmit device is the serial transmit shift register (TSR), which cannot be directly accessed by software. TSR obtains data from the TXREG transmit buffer register.

### 14.1.1.1 Enable Transmit

Enable USART transmit by configuring the following three control bits for asynchronous operation:

- TXEN=1
- SYNC=0
- SPEN=1

It is assumed that all other USART control bits are in their default state.

Set the TXEN bit of the TXSTA register to 1 to enable the USART transmitter circuit. Clear the SYNC bit of the TXSTA register to zero and use the USART configuration for asynchronous operation.

### Note:

- When the SPEN bit and TXEN bit are set to 1, the SYNC bit is cleared, TX/CKI/Opin is automatically configured as an output pin, regardless of the state of the corresponding TRIS bit.
- 2) When the SPEN bit and CREN bit are set to 1, the SYNC bit is cleared, and RX/DTI/Opin is automatically configured as an input pin, regardless of the state of the corresponding TRIS bit.



## 14.1.1.2 Transmit Data

Write a character to the TXREG register to start transmit. If this is the first character, or the previous character has been completely removed from the TSR, the data in TXREG will be immediately transmitted to the TSR register. If all or part of the TSR is still stored. The previous character, the new character data will be stored in TXREG until the stop bit of the previous character is transmitted. Then, after the stop bit is transmitted, after a TCY, the data to be processed in TXREG will be transmitted to TSR. When After data is transmitted from TXREG to TSR, the start bit, data bit, and stop bit sequence are transmitted immediately.

## 14.1.1.3 Transmit Interrupt

As long as the USART transmitter is enabled and there is no data to be transmitted in TXREG, the TXIF interrupt flag bit of the PIR1 register is set to 1. In other words, only when the TSR is busy processing the character and there are new characters queued for transmit in the TXREG, the TXIF bit It is in the cleared state. When writing TXREG, the TXIFflag bit is not cleared immediately. TXIF is cleared at the second instructions period after writing the instructions. Querying TXIF immediately after writing TXREG will return an invalid result. TXIF is a read-only bit and cannot Set or cleared by software.

TXIFinterrupt can be enabled by setting the TXIE interrupt enable bit of PIE1register. However, as long as TXREG is empty, the TXIFflag bit will be set to 1 regardless of the status of the TXIEenable bit.

If you want to use interrupt when transmitting data, set the TXIE bit to 1 only when the data is to be transmitted. After writing the last character to be transmitted to TXREG, clear the TXIE interrupt enable bit.

## 14.1.1.4 TSR Status

The TRMT bit of the TXSTA register indicates the status of the TSR register. The TRMT bit is a readonly bit. When the TSR register is empty, the TRMT bit is set to 1, and when a character is transferred from the TXREG to the TSR register, the TRMT is cleared. The TRMT bit remains Clear the state until all bits are removed from the TSR register. There is no interrupt logic related to this bit, so the user must query this bit to determine the state of the TSR bit.

Note: The TSR register is not mapped to the data memory, so the user cannot directly access it.

## 14.1.1.5 Transmit 9-bit Character

The USART supports 9-bit character transmit. When the TX9EN bit of the TXSTA register is 1, the USART will shift out 9 bits of each character to be transmitted. The TX9D bit of the TXSTA register is the 9th bit, which is the highest data bit. When the 9-bit data is transmitted, it must Before writing the 8 least significant bits to TXREG, write the TX9D data bit. After writing the TXREG register, the 9 data bits will be transferred to the TSR shift register immediately.

# 

## 14.1.1.6 Configure Asynchronous Transmit

- 1. Initialize the SPBRG register to obtain the required baud rate (see "USART baud rate generator (BRG)"
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit to 1.
- 3. If 9-bit transmit is required, set the TX9EN control bit to 1. When the receiver is set for address detection, set the 9th bit of the data bit to 1, indicating that the 8 lowest data bits are address.
- 4. Set the TXEN control bit to 1 to enable transmit; this will cause the TXIF interrupt flag bit to be set to 1.
- 5. If interrupt is required, set the TXIE interrupt enable bit in PIE1register to 1; if the GIE and PEIE bits in the INTCON register are also set to 1, interrupt will occur immediately.
- 6. If you choose to transmit 9-bit data, the 9th bit should be loaded into the TX9Ddata bit.
- 7. Load 8-bit data into TXREG register to start transmitting data.

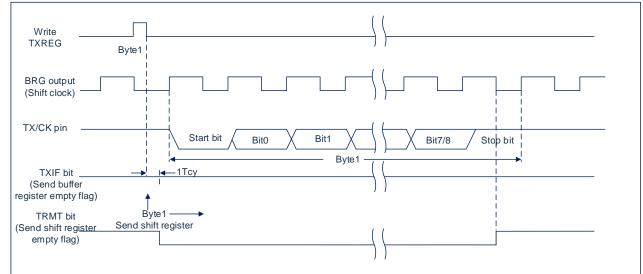


Fig 14-3: asynchronous transmit

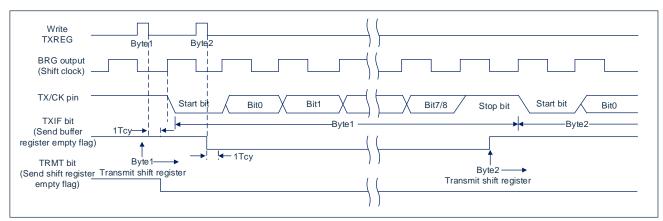


Fig14-4: asynchronous transmit (back to back)

Note: This time series diagram shows two consecutive transmit.



## 14.1.2 USART Asynchronous Receiver

Asynchronous mode is usually used in RS-232 system. Figure 14-2 shows the block diagram of the receiver. Receive data and driver data recovery circuit on RX/DTpin. The data recovery circuit is actually a 16 times baud rate as the operating frequency High-speed shifter, while the serial receive shift register (Receive Shift Register, RSR) works at the bit rate. When all the 8-bit or 9-bit data bits of the character are shifted in, they are immediately transferred to a 2-character FIFO (FIFO) buffer. FIFO buffer allows to receive 2 complete characters and the start bit of the third character, and then software must provide the received data to the USART receiver. FIFO and RSR register cannot be directly accessed by software. The RCREG register accesses the received data.

## 14.1.2.1 Enable Receiver

Enable the USART receiver by configuring the following three control bits for asynchronous operation.

- CREN=1
- SYNC=0
- SPEN=1

Assuming that all other USART control bits are in the default state. Set the CREN bit of the RCSTA register to 1 to enable the USART receiver circuit. Clear the SYNC bit of the TXSTA register to zero and configure the USART for asynchronous operation.

Note:

- 1) When the SPEN bit and TXEN bit are set to 1, the SYNC bit is cleared, and the TX/CKI/O pin is automatically configured as an output pin, regardless of the state of the corresponding TRIS bit.
- 2) When the SPEN bit and CREN bit are set to 1, the SYNC bit is cleared, and the RX/DTI/O pin is automatically configured as an input pin, regardless of the state of the corresponding TRIS bit.

## 14.1.2.2 Receive Data

Receiver data recovery circuit starts the receive character at the falling edge of the first bit. The first bit, usually called the start bit, is always 0. The data recovery circuit counts half a bit time to the center of the start bit. Check whether the bit is still zero. If the bit is not zero, the data recovery circuit will give up receiving the character without error, and continue to look for the falling edge of the start bit. If the zero check of the start bit passes, then the data recovery circuit counts a complete bit time and reaches the center position of the next bit. The majority detection circuit samples the bit and moves the corresponding sampling result 0 or 1 into the RSR. Repeat the process until all data bits are completed Sampling and moving it all into RSRregister. Measure the time of the last bit and sample its level. This bit is the stop bit and is always 1. If the data recovery circuit samples 0 at the stop bit position, the character frame error The flag will be set to 1, otherwise, the frame error flag of the character will be cleared.

When all data bits and stop bits are received, the character in the RSR will be immediately transferred to the receive FIFO of the USART and the RCIF interrupt flag bit of PIR1register is set to 1. The character at the top of the FIFO is moved out of the FIFO by reading the RCREG register.

Note: If you receive FIFO overflow, you cannot continue to receive other characters until the overflow condition is cleared.



## 14.1.2.3 Receive Interrupt

As long as the USART receiver is enabled and there is no unread data in the receive FIFO, the RCIF interrupt flag bit in the PIR1 register will be set to 0. The RCIF interrupt flag bit is read-only and cannot be set or cleared by software.

RCIF interrupt is enabled by setting all of the following bits:

- RCIE interrupt enable bit of PIE1 register;
- PEIE peripherals interrupt enable bit of INTCON register;
- GIE global interrupt enable bit of INTCON register.

If there is unread data in the FIFO, regardless of the state of the interrupt enable bit, the RCIF interrupt flag bit will be set to 1.

## 14.1.2.4 Receive Frame Error

Each character in the Receive FIFO buffer has a corresponding frame error status bit. The frame error indicates that the stop bit was not received within the expected time.

The framing error status is obtained by the FERR bit of the RCSTA register. The FERR bit must be read after reading the RCREG register.

Framing error (FERR=1) will not prevent receiving more characters. There is no need to clear the FERR bit.

Clearing the SPEN bit of the RCSTA register will reset the USART and forcibly clear the FERR bit. Framing error itself will not cause interrupt.

Note: If all characters received in the receive FIFO buffer have framing errors, repeated reading of RCREG will not clear the FERR bit.

## 14.1.2.5 Receive Overflow Error

The receive FIFO buffer can store 2 characters. However, if the third character is received before accessing the FIFO, an overflow error will occur. At this time, the OERR bit of the RCSTA register will be set to 1. The character inside FIFO buffer can be read, but before the error is cleared, no other characters can be received. The error can be cleared by clearing the CREN bit of the RCSTA register or by clearing the SPEN bit of the RCSTA register to make USART reset.

## 14.1.2.6 Receive 9-bit Character

The USART supports 9-bit data receive. When the RX9EN bit of the RCSTA register is set to 1, the USART will shift the 9 bits of each character received into the RSR. You must read the RX9D data bit after reading the lower 8 bits in RCREG.

## 

## 14.1.2.7 Asynchronous Receive Configuration

- Initialize the SPBRG register to obtain the required baud rate. (Please refer to the "USART baud rate generator (BRG)" chapter.
- 2. Set the SPEN bit to 1 to enable the serial port. The SYNC bit must be cleared to perform asynchronous operations.
- 3. If interrupt is required, set the RCIE bit in the PIE1 register and the GIE and PEIE bits in the INTCON register to 1.
- 4. If you need to receive 9 bits of data, set the RX9EN bit to 1.
- 5. Set the CREN bit to 1 to enable receive.
- 6. When a character is transferred from the RSR to the receive buffer, set the RCIF interrupt flag bit to1. If the RCIE interrupt enable bit is also set to 1, an interrupt will also be generated.
- 7. Read the RCREG register and get the received 8 low data bits from the receive buffer.
- 8. Read the RCSTA register to get the error flag bit and the 9th data bit (if 9-bit datareceive is enabled).
- 9. If overflow occurs, clear the OERR flag by clearing the CREN receiver enable bit.

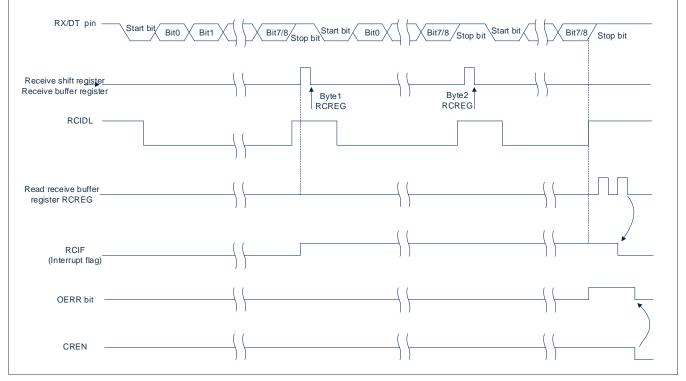


Fig 14-5: Asynchronous receive

Note: This time series diagram shows the situation of three words received in RX input pin. Reading RCREG (receive buffer) after the third word results in OERR (overflow) bit 1.



## 14.2 Clock Precision for Asynchronous Operations

The output of the internal oscillation circuit (INTOSC) is calibrated by the manufacturer. But when VDD or temperature changes, INTOSC will have a frequency shift, which will directly affect the asynchronous baud rate. The baud rate clock can be adjusted by the following methods, but some type of reference is required clock source.

## 14.3 USART Related Register

TXSTA0: transmit status and control register (1EH)								
1EH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
TXSTA0	CSRC0	TX9EN0	TXEN0	SYNC0	SCKP0		TRMT0	TX9D0
read/write	R/W	R/W	R/W	R/W	R/W		R	R/W
Reset value	0	0	0	0	0	0	1	0

Bit7	CSRC0:	clock sourceselection bit;
	Asynchronous mode:	Any value;
	Synchronous mode:	
		1=master control mode (internal BRG generate clock signal);
		0=slave mode (external clock source generate clock).
Bit6	TX9EN0:	9-bit transmit enable bit;
	1=	Select 9-bit transmit;
	0=	Select 8-bit transmit.
Bit5	TXEN0:	Transmit enable bit (1);
	1=	Enable transmit;
	0=	Disable transmit.
Bit4	SYNC0:	USART mode selection bit;
	1=	Synchronous mode;
	0=	Asynchronous mode.
Bit3	SCKP0:	Synchronous clock polarity selection bit.
	Asynchronous mode:	
		1= Invert the level of the data character and transmit to the TX/CK pin;
		0= Directly transmit data character to TX/CK pin.
	Synchronous mode:	
		0= Data is transmitted on the rising edge of clock;
		1= Data is transmitted on the falling edge of clock.
Bit2	not used	
Bit1	TRMT0:	Transmit shift register status bit;
	1=	TSR empty;
	0=	TSR full.
Bit0	TX9D0:	9 <sup>th</sup> bit of Transmit data.
		Can be address/data bit or parity check bit.

Note: In synchronous mode, SREN/CREN will invert the value of TXEN.



RUSTAU. TEU			, <b>,</b> ,			1	T	[
18H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
RCSTA0	SPEN0	RX9EN0	SREN0	CREN0	RCIDL0	FERR0	OERR0	RX9D0
read/write	R/W	R/W	R/W	R/W	R	R	R	R
Reset value	0	0	0	0	1	0	0	0
Bit7	SPE	N0: serial	portenable b	oit:				
	0		-		TX/CK pin cor	nfigured as se	rial portp in):	
				t (hold on res		3	- I - I - <i>I</i>	
Bit6	RX9E		receive enab		,			
		1= Selec	t 9-bit receiv	ve;				
		0= Selec	t 8-bit receiv	′e;.				
Bit5	SRE	N0: Single	e byte receiv	e enable bit.				
	Asynchron	ious any v	alue.					
		ode:						
	Synchron							
	master cor							
	ma	ode:						
			able single b					
			able single b	-				
	Synchron			e completed.				
	slave mo		aiue.					
Bit4	CRE		nuous receiv	ve enable bit.				
DIL	Asynchron							
	-	ode:						
			able receive;					
			able receive					
	Synchron	ious						
	ma	ode:						
		1=ena	able continuo	ous receive u	ntil clear CRE	N enable bit (0	CREN cover	SREN);
		0=dis	able continu	ous receive.				
Bit3	RCIE		ive idle flag b	oit.				
	Asynchron		eiver idle;					
	ma	ode:						
	<u> </u>			e initial bit, re	eceiving data.			
	Synchron	-	alue.					
Dito		ode:	error bit.					
Bit2	FER			he undeted	by rooding the	DODEC roai	stor and road	ive the
			/alid byte);	i be upualeu	by reading the	RCREGiegi		ive the
			ame error.					
Bit1	OFR		low error bit.					
Bitt	02.0				g CREN bit);			
			erflow error.					
Bit0	RX9			it of the data				
					a bit or the pa	rity check bit,	which must b	e
				user firmware	-	- ,		
			-					

## RCSTA0: receive status and control register (18H)





## 14.4 USART Baud Rate Generator (BRG)

The baud rate generator (BRG) is an 8-bit, dedicated to supporting the asynchronous and synchronous working modes of USART.

The SPBRG register determines the period of the free-running baud rate timer.

Table 14-1 contains the formula for calculating baud rate. Formula 1 is an example of calculating baud rate and baud rate error.

Table 14-1 shows the typical baud rate and baud rate error values under various asynchronous modes that have been calculated, which is convenient for you to use.

Writing a new value to the SPBRG register pair will cause the BRG timer to reset (or clear). This can ensure that BRG can output a new baud rate without waiting for a timer overflow.

If the system clock changes during a valid receive process, a receive error may occur or data loss may occur. To avoid this problem, the state of the RCIDL bit should be checked to ensure that the receive operation is idle before changing the system clock.

formula1: claculate baud rate error

For device with F<sub>SYS</sub>=8MHz, target baud rate=9600bps, asynchronous mode is 8-bit BRG:

target baud rate=
$$\frac{Fsys}{16([SPBRG] + 1)}$$

solve SPBRG:

$$X = \frac{\frac{FSYS}{\text{target baud rate}}}{16} - 1 = \frac{\frac{8000000}{9600}}{16} - 1 = [51.08] = 51$$

calculated baud rate = 
$$\frac{8000000}{16(51+1)}$$
 = 9615

error=
$$\frac{\text{claculated baud rate-target baud rate}}{\text{target baud rate}} = \frac{(9615-9600)}{9600} = 0.16\%$$

Table 14-1: baud rateformula

Configuration bit	BRG/USARTmode	baud rateformula		
SYNC				
0	8bit/asynchronous	Fsys/[16 (n+1)]		
1	8bit/synchronous	Fsys/[4 (n+1)]		

Note: n= value of SPBRG register.

Table 14-2: baud rate in asynchronous mode

Target baud rate	SYNC=0								
		F <sub>SYS=8.00</sub> MHz			Fsys=16.00MHz				
5	Real baud rate	error (%)	SPBRG value	Real baud rate	error (%)	SPBRG value			
2400	2404	0.16	207						
9600	9615	0.16	51	9615	0.16	103			
10417	10417	0	47	10417	0	95			
19200	19230	0.16	25	19230	0.16	51			



## 14.5 USART Synchronous Mode

Synchronous serial communication is usually used in a system with a master control device and one or more slave devices. The master control device contains the necessary circuits to generate the baud rate clock and provides clock for all devices in the system. The slave device can use master control clock, so no internal clock generation circuit is needed.

In synchronous mode, there are two signal lines: bi-directional data line and clock line. The slave device uses the external clock provided by the master control device to move the serial data in or out of the corresponding receive and transmit shift register. Because of the use of bi-directional data lines, synchronous operation can only use half-duplex mode. Half-duplex means: master control device and slave device can receive and transmit data, but can not receive or transmit at the same time. USART can be used as a master control device, or as a slave device.

## 14.5.1 Synchronous Master Control Mode

The following bits are used to configure the USART for synchronous master control operation:

- SYNC=1
- CSRC=1
- SREN=0 (to transmit); SREN=1 (to receive)
- CREN=0 (to transmit); CREN=1 (to receive)
- SPEN=1

Set the SYNC bit of the TXSTA register to 1 to use the USART configuration for synchronous operation. Set the CSRC bit of the TXSTA register to 1 to configure the device as a master control device. Clear the SREN and CREN bits of the RCSTA register to zero to ensure that the device is in transmit mode. Otherwise, the device is configured to receive mode. Set the SPEN bit of the RCSTA register to 1, enable USART.

#### 14.5.1.1 Master Control Clock

Synchronous data transmission uses an independent clock line to transmit data synchronously. The device configured as a master control device transmits clock signal on the TX/CK pin. When the USART is configured for synchronous transmit or receive operation, the TX/CK output driver automatically enables. Serial data bits are changed on the rising edge of each clock to ensure that they are valid on the falling edge. The time of each data bit is a clock period, and there can only be as many clock periods as there are data bits.

#### 14.5.1.2 Clock Polarity

The device provides clock polarity options to be compatible with Microwire. The clock polarity is selected by the SCKP bit of the TXSTA register. Set the SCKP bit to 1 to set the clock idle state to high. When the SCKP bit is 1, data on the falling edge of each clock changes. Clear the SCKP bit and set the clock idle state to low. When the SCKP bit is cleared, data changes on each rising edge of the clock.



#### 14.5.1.3 Synchronous Master control Transmit

The RX/DT pin output data of the device. When the USART configuration is synchronous master control transmit operation, the RX/DT and TX/CK output pins of the device are automatically enabled.

Write a character to the TXREG register to start the transmit. If all or part of the previous character is still stored in the TSR, the new character data is stored in TXREG until the stop bit of the previous character is transmitted. If this is the first character, Or the previous character has been completely removed from the TSR, the data in TXREG will be immediately transferred to the TSR register. When the character is transferred from TXREG to TSR, it will immediately begin to transmit data. Each data bit changes on the rising edge of the master control clock and remain effective until the rising edge of the next clock.

Note: The TSR register is not mapped to the data memory, so the user cannot directly access it.

#### 14.5.1.4 Synchronous Master Control Transmit Configuration

1. Initialize the SPBRG register to obtain the required baud rate.

(Please refer to the chapter "USART baud rate generator (BRG)".)

- 2. Set the SYNC, SPEN and CSRC bits to 1, enable synchronous master control serial port.
- 3. Clear the SREN and CREN bits to disable receive mode.
- 4. Set the TXEN bit to 1 to enable transmit mode.
- 5. If you need to transmit a 9-bit character, set TX9EN to 1.
- 6. If interrupt is required, set the TXIE bit in the PIE1 register and the GIE and PEIE bits in the INTCON register to 1.
- 7. If you choose to transmit 9-bit character, you should load the 9th bit of data into the TX9D bit.
- 8. Start transmit by loading data into TXREG register.

RX/DT pinBit0Bit1	_
	•
TX/CK pin	
TX/CK pin         (         )         )         (         (         )	
Write to TXREG ( ( ( ( ( ( ( ( ( ( ( ( ( ( ( ( ( ( (	
TXIF bit         ( (	
TRMT bit	
TXEN bit         1         () ()         1	_

Fig 14-6: synchronous transmit



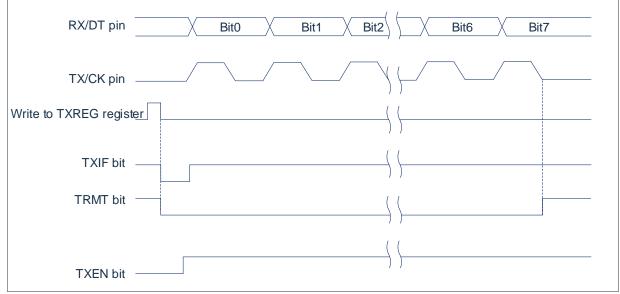


Fig 14-7: synchronous transmit (through TXEN)

## 14.5.1.5 Synchronous Master Control Receive

RX/DT pin receive data. When the USART configuration is synchronous master control receive, the output driver of the RX/DT pin of the device is automatically disabled.

In synchronous mode, set the single word receive enable bit (SREN bit of RCSTAregister) or continuous receive enable bit (CREN bit of RCSTA register) to 1 enable receive. When SREN is set to 1, the CREN bit is cleared, the number of clock period generated is as much as the number of data bit in single character. After a character transmission is over, the SREN bit is automatically cleared. When CREN is set to 1, a continuous clock will be generated until CREN is cleared. If CREN is cleared during a character transmission, The CK clock stops immediately and discards the incomplete character. If both SREN and CREN are set to 1, when the first character transfer is completed, the SREN bit is cleared, and CREN takes precedence.

Set the SREN or CREN bit to 1, start receiving. Sample the data on RX/DT pin at the falling edge of the TX/CK clock pin signal, and shift the sampled data into the receive shift register (RSR). When the RSR receives a complete character, the RCIF bit is set to 1, the character is automatically moved into the 2 byte receive FIFO. The lower 8 bits of the top character in the receive FIFO can be read through RCREG. As long as there are unread characters in the receive FIFO, the RCIF bit remains as 1.

#### 14.5.1.6 Slave Clock

Synchronous data transmission uses an independent clock line synchronous with the data line. Clock signal on the TX/CK line of the slave device is received. When the device is configured to operate synchronously from the transmit or receive, the output driver of the TX/CK pin automatically disable. The serial data bit is changed at the leading edge of the clock signal to ensure that it is valid on the back edge of each clock. Each clock period can only transmit one bit of data, so how many data bits must be received is determined by how many data bits transmitted.



## 14.5.1.7 Receive Overflow Error

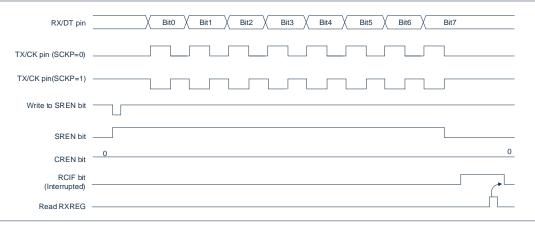
The receive FIFO buffer can store 2 characters. Before reading the RCREG to access the FIFO, if the third character is received completely, an overflow error will occur. At this time, the OERR bit of the RCSTA register will be set to 1. The previous data in the FIFO is not Will be rewritten. Two characters in the FIFO buffer can be read, but before the error is cleared, no other characters can be received. The OERR bit can only be cleared by clearing the overflow condition. If an overflow occurs, the SREN bit is set to 1, the CREN bit is in the cleared state, and the error is cleared by reading the RCREG register. If CREN is set to 1 during overflow, you can clear the CREN bit of the RCSTA register or clear the SPEN bit to reset USART, to clear the error.

## 14.5.1.8 Receive 9-bit Character

The USART supports receive 9-bit characters. When the RX9EN bit of the RCSTA register is 1, the USART moves the 9-bit data of each character received into the RSR. When reading 9-bit data from the receive FIFO buffer, it must read 8 lower bit of RCREG first.

## 14.5.1.9 Synchronous Master Control Receive Configuration

- 1. Initialize the SPBRG register to obtain the required baud rate. (Note: SPBRG>05H must be met)
- 2. Set the SYNC, SPEN and CSRC bits to 1 to enable synchronous master control serial port.
- 3. Make sure to clear the CREN and SREN bits.
- 4. If interrupt is used, set the GIE and PEIE bits of the INTCON register to 1, and set the RCIE bit of the PIE1 register to 1.
- 5. If you need to receive a 9-bit character, set the RX9EN bit to 1.
- 6. Set the SREN bit to 1 to enable receive, or set the CREN bit to 1 to enable continuous receive.
- 7. When the character receive is completed, set the RCIF interrupt flag bit to 1. If the enable bit RCIE is set to 1, an interrupt will also be generated.
- 8. Read the RCREG register to get the received 8-bit data.
- 9. Read the RCSTA register to get the 9th data bit (when 9-bit receive is enabled), and judge whether an error occurs during the receive process.
- 10. If an overflow error occurs, clear the CREN bit of the RCSTA register or clear SPEN to reset USART to clear the error.



#### Fig 14-8: synchronous receive (master control mode, SREN)

Note: The time series diagram illustrates the synchronous master control mode when SREN=1.



## 14.5.2 Synchronous Slave Mode

The following bits are used to configure USART for synchronous slave operation:

- SYNC=1
- CSRC=0
- SREN=0 (to transmit); SREN=1 (to receive)
- CREN=0 (to transmit); CREN=1 (to receive)
- SPEN=1

Set the SYNC bit of the TXSTA register to 1 to configure the device for synchronous operation. Set the CSRC bit of the TXSTA register to 1 to configure the device as a slave device. Clear the SREN and CREN bits of the RCSTA register to zero to ensure that the device is in transmit mode. Otherwise, the device will be configured as receive mode. Set the SPEN bit of the RCSTA register to 1, enable USART.

## 14.5.2.1 USART Synchronous Slave Transmit

The working principle of synchronous master control and slave mode is the same (see chapter "synchronous master control transmission").

## 14.5.2.2 Synchronous Slave Transmit Configuration

- 1. Set the SYNC and SPEN bits and clear the CSRC bit.
- 2. Clear the CREN and SREN bits.
- 3. If interrupt is used, set the GIE and PEIE bits of the INTCON register to 1, and set the TXIE bit of the PIE1 register.
- 4. If you need to transmit 9-bit data, set the TX9EN bit to 1.
- 5. Set the TXEN bit to 1 to enable transmit.
- 6. If you choose to transmit 9-bit data, write the most significant bit to the TX9D bit.
- 7. Write the lower 8 bits of data to the TXREG register to start transmission.

#### 14.5.2.3 USART Synchronous Slave Receive

Except for the following differences, the working principle of synchronous master control and slave mode is the same.

- 1. The CREN bit is always set to 1, so the receiver cannot enter the idle state.
- 2. SREN bit, can be "any value" in slave mode.



## 14.5.2.4 Synchronous Slave Receive Configuration

- 1. Set the SYNC and SPEN bits and clear the CSRC bit.
- 2. If interrupt is used, set the GIE and PEIE bits of the INTCON register to 1, and also set the RCIE bit of the PIE1 register.
- 3. If you need to receive a 9-bit character, set the RX9EN bit to 1.
- 4. Set the CREN bit to 1, enable receive.
- 5. When the receive is completed, set the RCIF bit to 1. If RCIE is set to 1, an interrupt will also be generated.
- 6. Read the RCREG register and get the received 8 low data bits from the receive FIFO buffer.
- 7. If you enable 9-bit mode, get the most significant bit from the RX9D bit of the RCSTA register.

If an overflow error occurs, clear the CREN bit of the RCSTA register or clear the SPEN bit to reset USART to clear the error.



## 15. Master Control Synchronous Serial Port (MSSP)Mod

## 15.1 Master Control SSP (MSSP)Mod General

master control synchronous serial port (Master Synchronous Serial Port, MSSP) mod is a serial interface for communicating with other peripherals or microcontrollers. These peripherals devices can be serial EEPROM, shift register, display driver or A/D converter, etc.

MSSP mod has the following two working modes:

- serial peripherals ports (SPI).
- I<sup>2</sup>C.
  - Full master control mode.
  - Slave mode (Support broadcast address call).

I<sup>2</sup>C interface supports the following modes at hardware level:

- master controlmode.
- Multi master mode.
- Slave mode.

## 15.2 SPI Mode

SPI mode allows simultaneous transmit and receive 8-bit data at the same time.SPI supports 3-wire mode and 4-wire mode communication.

The following three pins are used under 3-wire mod:

- Serial data input (SDIO)——RC7/SDIO
- Serial clock (SCK)——RC6/SCK
- slave selection (SS)——RC4/SS

The following three pins are used under 4-wire mod:

- serial data output (SDO)——RC5/SDO
- serial data input (SDI)——RC7/SDI
- serial clock (SCK)——RC6/SCK
- slave selection (SS)—RC4/SS



## 15.2.1 SPI Related Register

## SSPSTAT: SSP status register (94H)

94H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
SSPSTAT		CKE	MODE					
read/write		R/W	R/W					
Reset value		0	0					

Bit7 Bit 6	Save, write 0 CKE: CKP=	
		0= Transmit data on the rising edge of SCK pin;
		1= Transmit data on the falling edge of SCK pin.
	CKP =	1
		0= Transmit data on the falling edge of SCK pin;
		1= Transmit data on the rising edge of SCK pin.
Bit5	MODE:	Mode selection
		1=3-wire mode (When need to transmit, SDIO port TRIS bit needs to be
		cleared to 0; when need to receive, SDIO port TRIS needs to be set to 1) 0=4-wire mode

Bit4~Bit0 Not used in SPI mode.



		5 (	/					
14H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0
read/write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0
Bit7			Vrite conflict c n the process		eceive data, tr	y to write to t	he SSPBUF	register.
		0= N	lo conflict.			-		
Bit6		SSPOV: F	Receive overfl	ow flag bit.				
		c s t t k	overflow occur lave mode. In SSPBUF to av o 1, because o	s, the data in slave mode oid overflow every time yo	the previous d SSPSR will b , even if transr In master cor ou receive or tr egister (this bi	e lost. Overf nit data only, ntrol mode, th ransmit new	low will only user must ro ne overflow b data, it must	occur in ead it is not set be started
Bit5		SSPEN: S	Synchronous s	serial port en	able bit.			
		1= E	Enable serial p	ort and conf	igure SCK, SD	O, SDI and	SS as serial	port pin.
		0= c	lisable serialp	ort and conf	igure these pir	ns as I/O por	t pins.	
Bit4		CKP: 0	Clock polarity	selection bit.				
		1= 0	Clock is high v	vhen idle.				
		0= 0	Clock is low w	hen idle.				
Bit3~Bit0	S	SPM<3:0>: S	Synchronous s	serial port mo	ode selection b	oit;		
		0000= 9	SPI master co	ntrol mode,	clock= Fsys/4	;		
		0001= 9	SPI master co	ntrol mode,	clock= Fsys/1	6;		
		0010= 5	SPI master co	ntrol mode,	clock= F <sub>SYS</sub> /6	4;		
		0011= g	SPI master co	ntrol mode,	clock= TMR2	output/2;		
		0100= 5	SPI slave mod	e, clock= S	CKpin, enable	e SSpin cont	rol;	
		0101= 5			CKpin, disabl			be used
			Save;					
		0111= \$	Save;					
		1000= j	<sup>2</sup> C master cor	ntrol mode,	clock= Fsys/ (	4 * (SSPADE	D+1));	
		1001= E	Disable load fu	inction;				
		1010= \$	Save;					
		1011= 5	Save;					
		1100= \$	Save;					
			Save;					
			<sup>2</sup> C slave mode	e, 7-bit addre	ess, and allow	start bit and	stop bit interi	rupt;
		1111= s	ave.					

## SSPCON: SSP control register (14H)



## 15.2.2 SPI Working Principle

When initializing the SPI, several options need to be specified. They can be specified by programming the corresponding control bits (SSPCON<5:0> and SSPSTAT<7:6>). These control bits are used to specify the following options:

- master control mode (SCK as clock output)
- clock polarity (SCK idle state)
- clock rate (only in master control mode)
- Slave mode (SCK as clock input)
- Sampling phase of input data (the middle or end of data output time)
- clock edge

   (output data on the rising/falling edge of SCK)
- slave selection mode (only in slave mode)

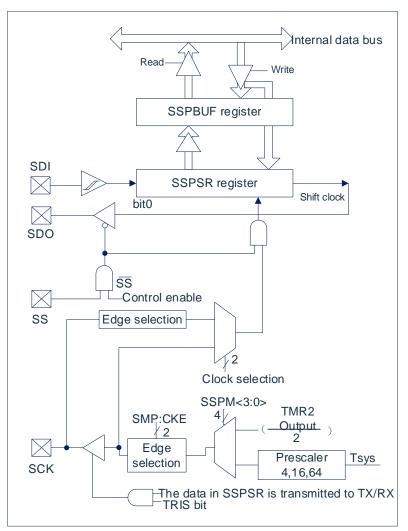


Fig 15-1 MSSP mod block diagram in SPI mode

## Note: I/O pin has diode protection to VDD and VSS.

MSSP mod consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPBUF). SSPSR moves data in and out of the device, with the most significant bit first. SSPBUF saves the data written to the SSPSR last time until the new receive. The data is ready. Once the 8-bit data receive is completed, the byte



is moved into the SSPBUF register. Then, the interrupt flag bit SSPIF of the PIR1 register is set to 1. This double-buffered data receive method (SSPBUF) allows reading the newly received data before starting to receive the next byte. During the data transmit/receive period, any attempt to write to the SSPBUF register will be ignored, and the write conflict detection bit WCOL of the SPCON register will be set to 1. At this time, the user must clear the WCOL bit by software, Otherwise it cannot be judged whether the next write operation to SSPBUF is successfully completed.

When the application software is waiting for the receive valid data, it should read the previous data in the SSPBUF before the next data byte to be transmitted is written into the SSPBUF. The buffer full flag bit BF (SSPSTAT register) is used to indicate when the SSPBUF has been loaded the received data (transmit is completed). If the SPI is only used as a transmitter, you don't need to pay attention to the received data. MSSP interrupt can usually be used to determine when the transmit or receive is completed. If you do not use interrupt to handle the data transmission and reception and use software to query, this method also ensures that no write conflicts occur.



## 15.2.3 Enable SPI I/O

To enable the serial port, the MSSP enable bit SSPEN of the SSPCON register must be set to 1. To reset or reconfigure the SPI mode, first clear the SSPEN bit, reinitialize the SSPCON register, and then set the SSPEN bit to 1. This will set SDI, SDO, The SCK and SS pins are configured as serial port pins. To use these pins as serial ports, the data direction bits (in the TRIS register) must be programmed correctly, as follows:

- SDI controlled by SPI mod;
- TRISC<5> of SDO must be cleared;
- The TRISC<6> bit of SCK (master control mode) must be cleared;
- The TRISC<6> bit of SCK (slave mode) must be set to 1;
- The TRISC<4> of SS (slave mode) must be set to 1.

For any unwanted serial port function, you can skip it by setting the corresponding data direction (TRIS) register to the opposite value.

## 15.2.4 Master Control Mode

The master device controls SCK, so it can start data transmission at any time. The master device determines when the slave device should broadcast data according to the software protocol.

In master control mode, once data is written into the SSPBUF register, it will start to transmit or receive. If SPI is only used as a receiver, you can disable SDO output (program it to input). SSPSR register is connected to the SDI pin at the set clock rate the signal performs continuous shift input. After each byte receive is completed, it will be treated as a normal receive byte and loaded into the SSPBUF register (corresponding to interrupt and status position 1). This can be used as a "line activity monitoring" mode, which is very useful.

The clock polarity can be selected by programming the CKP bit of the SSPCON register accordingly. Figure 15-2, Figure 15-3, Figure 15-4, and Figure 15-5 show the SPI communication waveforms, where MSb is first transmitted. In master control mode, the SPI clock rate (bit rate) can be programmed by the user to one of the following rates:

- F<sub>SYS</sub>/4 (or TCY)
- F<sub>SYS</sub>/16 (or 4.TCY)
- F<sub>SYS</sub>/64 (or 16.TCY)
- TIMER2 output/2



Figure 15-2 shows the waveform of the master control mode. When the CKE bit of the SSPSTAT register is 1, the SDO data is valid before the clock edge appears on the SCK. The figure indicates the time to load the received data into the SSPBUF.

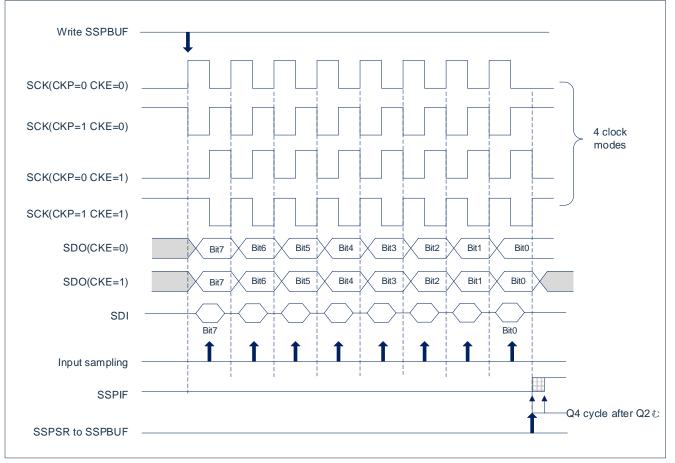


Fig15-2: SPI mode waveform (master control mode)

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## 15.2.5 Slave Mode

In slave mode, when an external clock pulse appears on the SCK pin, transmit and receive data. When the last bit of data is latched, the SSPIF interrupt flag bit of PIR1register is set to 1.

In slave mode, the clock is provided by the external clock source on the SCK pin. The external clock must meet the minimum time requirements for high and low levels specified in the electrical specifications.

In the sleep state, the slave device can still transmit/receive data. When a byte is received, the device is awakened from the sleep state.

## 15.2.6 Slave Synchronous Selection

SS pin allows the device to work in synchronous slave mode. SPI must work in slave mode, and enable SSpin to control SSPxCON1<3:0> = 04h). To use SS pin as input in, the pin driver cannot be set to low level. When the SS pin is low, the transmit and receive of the data are enabled, and the SDO pin is used by the driver. When the SS pin is high, the SDO pin is no longer drived even during the data transmit process. It becomes a floating output. According to the needs of the application, an external pull up/ pull down resistor can be connected.

After SPI mod reset, the bit counter is forced to 0. This can be achieved by forcing the SS pin to be pulled high or clearing the SSPEN bit. Connecting the SDO pin and the SDI pin can simulate a two-wire communication. When SPI When it needs to work as a receiver, SDOpin can be configured as input. This will disable the transmit data from SDO. Because SDI will not cause a bus conflict, it can always be reserved as input (SDI function).

Note: When SPI works in slave mode and SSpin control is enabled (SSPxCON1<3:0> = 0100), if SSpin is set to VDD level, SPI mod will be reset.

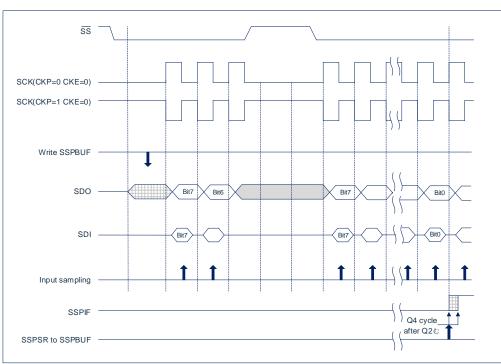


Fig 15-3: Slave synchronous waveform



CMS79FT73x

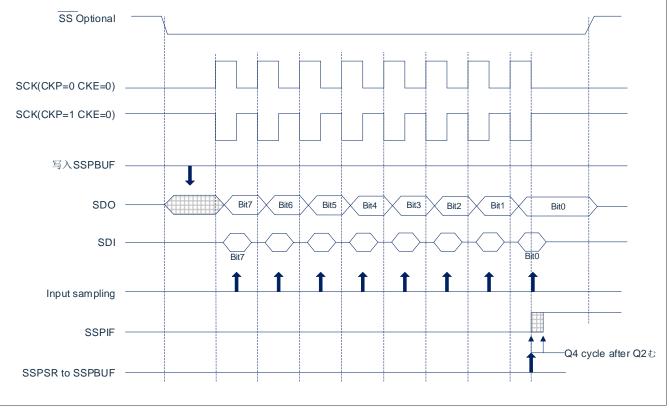


Fig 15-4: SPI mode waveform (slave mode, CKE=0)



## 15.2.7 Sleep Operation

In sleep mode, all mod clocks will stop, and before the device is awakened, transmit/receive will remain in this stagnant state. When the device returns to running mode, the mod will resume to transmit and receive data.

## 15.2.8 Effect of Reset

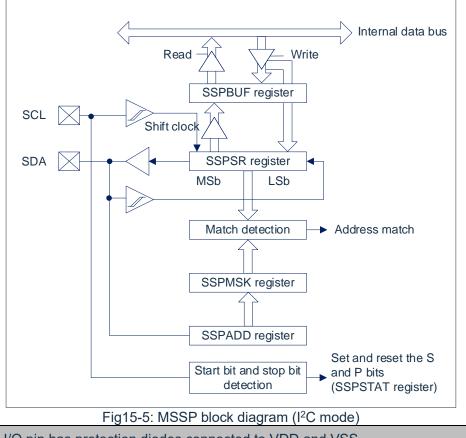
reset will disable MSSP mod and terminate the current transmission.



## 15.3 I<sup>2</sup>C Mod

When MSSP mod works in I<sup>2</sup>C mode, it can realize all master control and slave functions (including broadcast call support), and use hardware to provide interrupts of the start and stop bits to determine when the bus is idle (multi-master function).

There are two pins for data transmission. You can select the I2C function pin assignment in the system configuration register to RD0/RD1 or RC6/RC7. When using the I2C function, the user must pass TRISD<1:0> or TRISC<7:6 > to configure these pins as input pins. By setting the MSSP enable bit, SSPEN, of the SSPCON register to 1, MSSP mod function is enbaled.



Note: The I/O pin has protection diodes connected to VDD and VSS.

MSSP mod has 7 registers for I<sup>2</sup>C operation. They are:

- MSSP control register1 (SSPCON)
- MSSP status register (SSPSTAT)
- MSSP shift register (SSPSR): not directly accessible
- MSSP masking register (SSPMSK)
- MSSP control register2 (SSPCON2)
- serial receive/transmit buffer register (SSPBUF)
- MSSP address register (SSPADD)

You can use SSPCON register to control the operation of I<sup>2</sup>C. You can use the SSPM<3:0> mode selection bit (SSPCON register) to select one of the following I<sup>2</sup>C modes:

- I2C slave mode, 7-bit address, allow start bit and stop bit interrupt
- I2C master control mode, clock=FSYS/ (4\* (SSPADD+1))



If the SCL and SDA pins have been programmed as input pins (set the corresponding TRIS bit to 1), selecting any I<sup>2</sup>C mode and SSPEN bit as 1 will force the SCL and SDA pins to be open drain.



## 15.3.1 Related Register Illustration

## SSPSTAT: SSP status register (94H)

94H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
SSPSTAT		IDLE	D/A	Р	S	R/W		BF
read/write		R	R	R	R	R		R
Reset value		1	0	0	0	0		0
Bit7		not u	and Noture	ed in I <sup>2</sup> C mod	0			
Bit6				control mode				
Dito						valid, all mast	ter control ope	erations ca
			use this	s bit to determ	nine whether t	o terminate)		
			1= No mas	ster control op	peration on the	e bus		
			0= The ma	ster control c	peration is in	progress on th	ne bus	
Bit5		[	D/A: data/ad	ldress bit.				
			1= Indicate	es that the las	t receive or tr	ansmit byte is	data.	
			0= Indicate	es that the las	t receive or tr	ansmit byte is	address.	
Bit4			P: Stop bi	-	eared when N	ISSP mod is d	lisabled (SSP	EN is
					op bit was fina	lly detected (t	he bit is 0 whe	en reset).
			0= Indicate	es that the sto	op bit was not	detected at th	e end.	
Bit3			S: Start bi	t (this bit is cl	eared when d	isable MSSP	mod (SSPEN	is cleared
			1= Indicate	es that the sta	art bit was fina	lly detected (t	he bit is 0 wh	en reset).
			0= The sta	rt bit was not	detected at th	ne end.		
Bit2		R	W: Read/w	rite bit.				
						t information a		
					-	ne address ma	atch to the nex	kt start bit,
	In	l <sup>2</sup> C slave mo		or non-ACK	bit.			
	In	I I-C Slave mo	1= read	4				
			0 = write					
	I <sup>2</sup> C mast	ter control mo		5.				
	r o mas			smiting.				
				transmiting.				
				-	R operation b	etween this bi	t and SEN, R	SEN, PEN
				-	-	ther MSSP is		
Bit1		not u	sed					
Bit0			BF buffer f	ull status bit.				
		rece	eive:					
			1= rece	eive complete	, SSPBUF full			
			0= rece	eive not comp	lete, SSPBU	F empty.		
		trans	mit:					
			1 = data	a transmiting	(not including	ACK and stop	o bit), SSPBU	F full.
			0 = data	a transmit cor	mplete (not ind	cluding ACK a	nd stop bit), S	SPBUF
			empty.					



Bit0

14H	Bit7	BItb	BIt5	BIt4	BIt3	BIt2	BIT	BItU			
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0			
read/write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
Reset value	0	0	0	0	0	0	0	0			
Bit7		WCO		Write conflict detection bit.							
	maste	r control mode		to write to the		gister when I <sup>2</sup>	C does not m	eet the			
				of starting tra	nsmit data.						
			0= no coi	nflict.							
		Slave mode				1 10 11		., .			
				transmitting the	-	ord, write the	SSPBUF reg	ister again			
			0= no co	ar through sof	itware).						
Bit6		SSPO\		overflow flag b	hit (only valid	in slave recei	ve mode)				
Bito				SSPBUF rec				receives a			
				. In the transm							
			-	clear through :				,			
		0	= No overfl	OW.							
Bit5		SSPEN	I: Synchror	nous serial por	rt enable bit ( <sup>-</sup>	These pins mu	ust be correct	ly			
			configure	d as input or o	output pins).						
		1		erial port and	-	-	-	t pin.			
		-		erial port and	-	se pins as I/C	) port pins.				
Bit4		CKF		arity selection	ı bit.						
	In I	<sup>2</sup> C slave mode		ase control.							
			1 = enabl			vacion) (upod	to opouro dot				
			=	clock line is lo ment time).	JW (CIUCK EXIE	ension) (usea	to ensure dat	a			
	In I <sup>2</sup> C maste	r control mode									
Bit3~Bit0		SSPM<3:0>	-	nous serial por							
		0000		er control mod							
		0001		er control mod							
		0010 0011		er control moder control moder							
		0100		er control mode mode, clock		-					
		0100		e mode, clock	-	-		can be			
		0101	used as I					ban bo			
		0110									
		0111	= save.								
		1000	<ul> <li>I<sup>2</sup>C master</li> </ul>	er control mod	le, clock=Fs	sys/ (4 * (SSP	ADD+1)).				
		1001	= Disable lo	bad function.							
		1010	= save.								
		1011	= save.								
		1100									
		1101									
		1110		mode, 7-bit	address, allo	w start bit and	d stop bit inter	rupt.			
		1111	= save.								

## SSPCON: SSP control register (14H)

Bit6

Bit5

Bit4

Bit3

Bit2

Bit1

Bit7

14H



001 00112.00	. sona si rog		- (5.11)	,					
91H	Bit7	I	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
SSPCON2	GCEN	AC	KSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN
read/write	R/W	F	R/W	R	R/W	R/W	R/W	R/W	R/W
Reset value	0		1	0	0	0	0	0	0
	· ·			-				-	
Bit7	GCE	EN:	Broadca	ast call enable	bit (only in I <sup>2</sup>	<sup>2</sup> C slave mod	le).		
		1=		wed to genera			-	eral call add	ress
				in SSPSR.	-		-		
		0=		broadcast cal					
Bit6	ACKST	AT:	ACK sta	atus bit (only ir	n I <sup>2</sup> C master	control mode	e).		
	In master con								
	transmit mo	de:	4 D:-I	not rocalize -	roopores free	m the classes	lovico		
				not receive a	-				
Bit5	ACKI	ידח		esponse from t ta bit (only in l					
DIG	In master con			ue of the user'				e is complete	be
	receive mo				0 10000100 0				
		- '	1 = not	respond.					
			0 = resp	-					
Bit4	ACKE	EN:	ACK en	able bit (only i	n I <sup>2</sup> C master	control mod	e).		
	In master con	trol							
	receive mo	de:							
				rt the response	-		nd SCL pin,	transmit AC	KDT data
				omatically clea	-	vale.			
Bit3	RCE	EN:		e enable bit (or		ster control n	node).		
		1=		I <sup>2</sup> C receive me	-				
		0=	Receive						
Bit2	PE	EN:	stop en	able bit (only i	n I <sup>2</sup> Cmaster o	controlmode)			
			1 = Sta	rt stop conditio	on on SDA ar	nd SCL pin. A	utomatically	/ cleared by	
			hardwa	re.					
			0 =idle.						
Bit1	RSE			enable bit (on	-				
		1=		repeated start	conditions or	n the SDA ar	nd SCL pins.	Automatical	ly
		0=	idle.	by hardware.					
Bit0	SF	EN:		able bit.					
Dito	In master con		e.art of						
		de:							
				rt the start con	ditions on the	e SDA and S	CL pins. Aut	tomatically c	eared by
			hardwa						
			0 = idle						
	In slave mo	de:							
				h transmit and	receive will e	enable clock	extension (e	enable clock	
			extension 0 = disc	able clock exte	ntion.				
			0 – 0136						

## SSPCON2: SSP control register2 (91H)



## 15.3.2 Master Control Mode

The master control mode works by generating interrupt when the start and stop conditions are detected. The stop (P) bit and the start (S) bit are cleared when reset or disable MSSP mod. When the P bit is set to 1, the control of I2C bus can be obtained; otherwise the bus is idle, and both the P and S bits are zero.

In master control mode, the SCL line is manipulated by the MSSP hardware, and SDA pin must be configured as input (the corresponding pin TRIS bit is set to 1). The following events will set the MSSP interrupt flag bit SSPIF to 1 (if MSSP interrupt is allowed, interrupt will be generated):

Start condition

Stop condition

٠

Reply to transmit

- Data transmission byte has been transmitted/received
- Repeated start conditions

## 15.3.3 I<sup>2</sup>C Master Control Mode Support

The master control mode can be enabled by setting the corresponding SSPM bit in SSPCON to 1 or clearing it and setting the SSPEN bit to 1. Once the master control mode is enabled, the user can select the following 6 operations:

- 1. Issue a start condition on SDA and SCL.
- 2. Issue a repeated start condition on SDA and SCL.
- 3. Write the SSPBUF register to start data/address transmit.
- 4. Generate a stop condition on SDA and SCL.
- 5. Configure the I2C port to receive data.
- 6. The response condition is generated after the data byte is received.

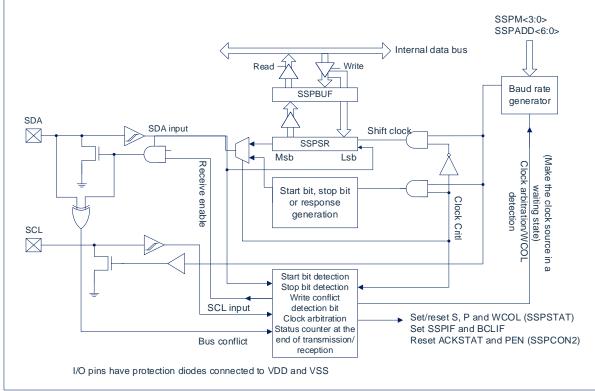


Fig 15-6: MSSP block diagram (I<sup>2</sup>C<sup>™</sup> master control mode)



Note: When configured as I<sup>2</sup>C master mode, MSSP module does not allow event queuing. For example, before the end of the start condition, the user is not allowed to issue another start condition and write to the SSPBUF register immediately to initiate the transfer. In this case, SSPBUF will not be written and the WCOL bit will be set to 1, which indicates that no write operation to SSPBUF has occurred.

## 15.3.3.1 I<sup>2</sup>C Master Control Mode Opeartion

All serial clock pulses and start/stop conditions are generated by the master device. The stop condition or the repeated start condition can end the transmission. Because the repeated start condition is also the beginning of the next serial transmission, the I<sup>2</sup>C bus will not be released. In the master control transmit mode, the serial data is output through SDA, and the serial clock is output by SCL. The first byte of the transmit includes the address (7 bits) and read/write (R/W) bits of the receiver. In this case, R/W bit will be logic 0. Serial data transmits 8 bits each time. Every time a byte is transmitted, an acknowledge bit will be received. The output of the start and stop conditions indicates the start and end of the serial transmission.

In master control receive mode, the first byte of transmit includes the address (7 bits) of the transmit device and the R/W bit. In this case, the R/W bit will be logic 1. Therefore, the first byte of transmit byte is a 7-bit slave device address, followed by 1 to indicate receive. The serial data is received through SDA, while the serial clock is output by SCL. Every time 8 bits of serial data are received. Every time a byte is received, an answer bit will be transmitted. Start and stop conditions indicate the start and end of transmit, respectively.

In I<sup>2</sup>C mode, the baud rate generator used in SPI mode is used to set the SCL clock frequency to 100KHz, 400KHz or 1MHz. The reload value of the baud rate generator is located in the lower 7 bits of the SSPADD register. When a write to SSPBUF occurs during operation, the baud rate generator will automatically start counting. If the specified operation is completed (ie, the last data bit of transmit is followed by ACK), the internal clock will automatically stop counting, and the SCL pin will remain in its last state.

The following is a typical transmit event sequence:

- The user generates a start condition by setting the start enable bit SEN (SSPCON2 register) to 1.
- SSPIF set to 1. Before performing any other operations, MSSP mod will wait for the required startup time.
- The user will load the SSPBUF from the device address to transmit.
- The address is moved out of the SDA pin until all 8 bits are transmitted.
- MSSP mod shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSPCON2 register.
- MSSP mod sets the SSPIF bit to 1 at the end of the 9th clock period, generating an interrupt.
- The user loads 8-bit data into SSPBUF.
- Data is moved out from the SDA pin until all 8 bits are transmitted.
- MSSP mod shifts in the ACK bit from the slave device and writes its value into the ACKSTAT bit of the SSPCON2 register.
- MSSP mod sets the SSPIF bit to 1 at the end of the 9th clock, generating an interrupt.
- The user generates a stop condition by setting the stop enable bit (PEN) bit (SSPCON2 register) to 1.
- Once the stop condition is completed, an interrupt will be generated.



## 15.3.4 Baud Rate Generator

In I<sup>2</sup>C master control mode, the baud rate generator reloaded value is located in the lower 7 bits of the SSPADD register (Figure 15-7). When the value is loaded, the baud rate generator will automatically start counting and decrement to 0, and then stop until the next reload. BRG will count down twice on the Q2 and Q4 clock periods in each instructions period (TCY). In I<sup>2</sup>C master control mode, BRG will be automatically reloaded. For example, when clock arbitration occurs, BRG will be reloaded when SCL pin is sampled to high level (Figure 15-8).

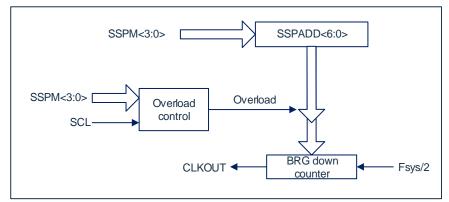


Fig 15-7: baud rate generator block diagram

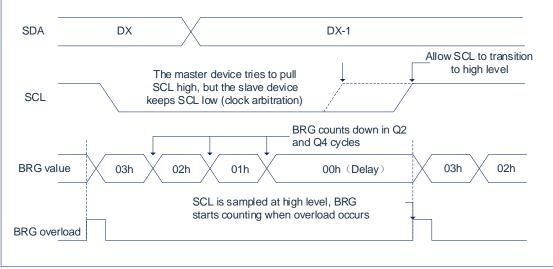


Fig 15-8: Time seires of baud rate generator with clock arbitration



## 15.3.5 I<sup>2</sup>C Master Control Mode Transmit

Transmit a data byte and a 7-bit address can be achieved directly by writing a value to the SSPBUF register. This operation will set the buffer full flag bit BF to 1, and the baud rate generator will start counting, and at the same time start the next transmit. After the falling edge of SCL is valid, each bit of address/data will be shifted out to the SDA pin. In a baud rate generator full return count period (TBRG), SCL remains low. Data should be released to high level in SCL when SCL pin is released to high level, it will remain high for the entire TBRG. During this period and a period of time after the next SCL falling edge, the data on the SDA pin must remain stable. After the 8th bit is shifted out (the falling edge of the 8th clock period), the BF flag bit is cleared, and the master device releases SDA.

At this time, if an address match occurs or data is correctly received, the addressed slave device will respond with an ACK bit at the 9th bit time. The ACK status is written to the ACKDT bit at the falling edge of the 9th clock period. After master device receiving the response, the response status bit ACKSTAT will be cleared; if the response is not received, the bit will be set to 1. After the 9th clock, the SSPIF bit will be set to 1, and the master control clock (baud rate generator) will be suspended until until the next data byte is loaded into SSPBUF, SCLpin remains low and SDA remains unchanged.

After writing SSPBUF, each bit of address is shifted out on the falling edge of SCL until all 7 bits of address and R/W bit are shifted out. At the falling edge of the eighth clock, the master device pulls the SDA pin to high level to allow the slave device to send an acknowledgment response. At the falling edge of the 9th clock, the master device determines whether the address is recognized by the slave device by sampling the SDA pin. The status of the ACK bit is loaded into the ACKSTAT status bit (SSPCON2 register). After the 9th clock falling edge of the transmit address, SSPIF is set to 1, the BF flag bit is cleared, the baud rate generator is turned off until the next write operation to SSPBUF, and the SCL pin remains low, allowing the SDA pin to suspend.

## 15.3.5.1 BF Status Indication

In transmit mode, the BF bit (SSPSTAT register) is set to 1 when the CPU writes SSPBUF, and is cleared after all 8 bits of data are shifted out.

#### 15.3.5.2 WCOL Status Indication Bit

If the user writes SSPBUF during the transmit process (that is, when the SSPSR is still moving out of the data byte), WCOL is set to 1 and the contents of the buffer remain unchanged (no write operation has occurred). WCOL must clear through software.

#### 15.3.5.3 ACKSTAT Status Indication

In transmit mode, when the slave device transmits a response (ACK=0), the ACKSTAT bit (SSPCON2 register) is cleared; when the slave device does not respond (ACK=1), the bit is 1. The slave device recognizes its address (Including the broadcast call address) or after receiving the data correctly, a response will be transmitted.



## 15.3.6 I<sup>2</sup>C Master Control Mode Receive

By programming receive enable bit RCEN (SSPCON2 register) to enable master control mode receive. The baud rate generator starts counting, and each time the count returns, the state of the SCL pin changes (from high to low or from low to high), and data is shifted into SSPSR. After the falling edge of the eighth clock, the receive enable flag bit is automatically cleared, the content of SSPSR is loaded into SSPBUF, the BF flag bit is set to 1, the SSPIF flag bit is set to 1, the baud rate generator pauses counting, and the SCL remains at low level. At this time, MSSP is in idle state, waiting for the next command. When the CPU reads the buffer, the BF flag bit will be automatically cleared. By setting the response sequence enable bit ACKEN (SSPCON2 register) to 1, the user can end the receive transmit response bit.

#### 15.3.6.1 BF Status Indication

When receiving, when the address or data byte is loaded from SSPSR into SSPBUF, the BF bit is set to 1, and the BF bit is cleared when reading the SSPBUF register.

## 15.3.6.2 WCOL Status Indication

If the user writes SSPBUF during the receive process (that is, when the SSPSR is still moving into the data byte), the WCOL bit is set to 1, and the buffer content remains unchanged (no write operation has occurred).

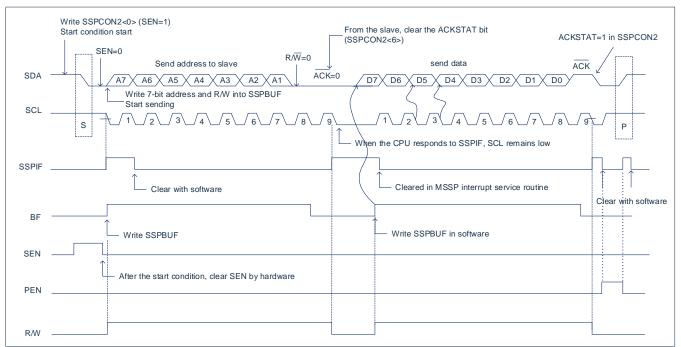


Fig15-9: time seires of I<sup>2</sup>C<sup>™</sup> master control mode transmit



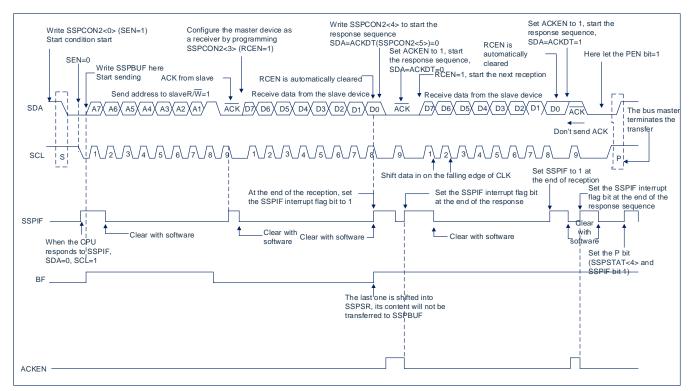


Fig 15-10: Time series of I<sup>2</sup>C<sup>™</sup> master control mode receive (7-bit address)



## 15.3.7 I<sup>2</sup>C Master Control Mode Start Condition Time Series

To initiate a start condition, the user should set the start condition enable bit SEN of the SSPCON2 register to 1. When both SDA and SCL pins are sampled as high, the baud rate generator reloads the contents of SSPADD<6:0> and starts counting. When the baud rate generator timeout (TBRG) occurs, if both SCL and SDA are sampled as high level, the SDA pin is low level by the driver. When SCL is high level, setting the SDA driver to low level is the startup condition. Set the S bit (SSPSTAT register) to 1. Then the baud rate generator reloads the contents of SSPADD<6:0> and resumes counting. When the baud rate generator times out (TBRG), the SEN bit of the SSPCON2 register will be automatically cleared by hardware. The baud rate generator is pasued, the SDA line remains low, and the start condition ends.

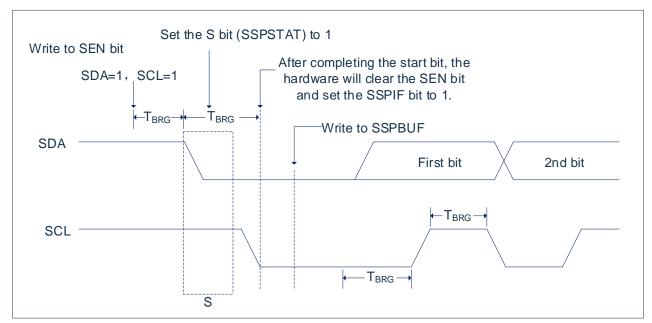


Fig 15-11: time series for the frist starting bit

## 15.3.7.1 WCOL Status Indication

When the startup sequence is in progress, if the user writes SSPBUF, WCOL is set to 1, and the buffer content remains unchanged (no write operation has occurred).

Note: Since event queues are not allowed, the lower 5 bits of SSPCON2 cannot be written before the start condition ends.



### 15.3.8 I<sup>2</sup>C Master Control Mode Repeat Condition Time Series

When the RSEN bit (SSPCON2 register) is programmed to be high and the I2C logic mod is in an idle state, a repeated start condition will occur. When the RSEN bit is 1, the SCL pin is pulled low. When the SCL pin is sampled low, baud rate generator loads the contents of SSPADD<6:0> and starts counting. In a baud rateg enerator counting period (TBRG), the SDA pin is released (its pin level is pulled high). When baud rate generator timeout, if SDA is sampled as high, SCL pin will be pulled high. When SCL pin is sampled as high, the baud rate generator will be reloaded into the contents of SSPADD<6:0> and start counting. SDA and SCL must be in one count period TBRG and sampled as high level. Then the SDA pin is pulled low (SDA = 0) and keeps a count period TBRG while SCL is high level. Then the RSEN bit (SSPCON2 register) will be automatically cleared, The baud rate generator will not be reloaded, and the SDA pin remains low. Once the start condition is detected on the SDA and SCL pins, the S bit (SSPSTAT register) will be set to 1. The SSPIF bit will not be set to 1 until the baud rate generator times out.

Once the SSPIF bit is set to 1, the user can write the 7-bit address into SSPBUF. When the first 8 bits are transmitted and an ACK is received, the user can transmit 8-bit data.

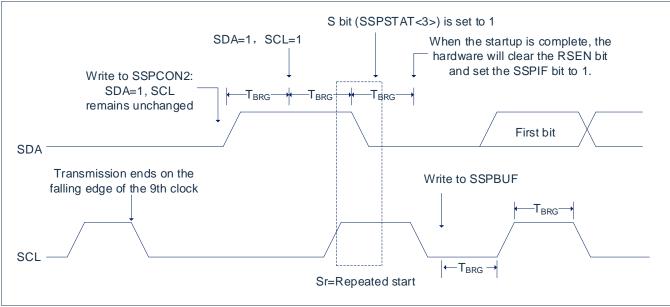


Fig 15-12: time series of repreat condition

# 15.3.8.1 WCOL Status Indication

When the repeated start sequence is in progress, if the user writes SSPBUF, WCOL is set to 1, and the buffer content remains unchanged (no write operation has occurred).

Note: As events are not allowed to be queued, the lower 5 bits of SSPCON2 cannot be written until the repeated start condition ends.



# 15.3.9 ACK Time Series

Set the ACK enable bit ACKEN (SSPCON2register) to 1 to enable the acknowledgement. When this bit is set to 1, the SCL pin is pulled low, and the content of the ACK data bit appears on the SDA pin. If the user wants to generate a response, it should clear the ACKDT bit to zero; otherwise, the user should set the ACKDT bit to 1 before the start of the ACK. Then the baud rate generator counts the full return period (TBRG), and then the SCL pin level is pulled high. When the SCL pin is sampled as at high level (clock arbitration), baud rate generator counts for another TBRG period. Then SCLpin is pulled low. After that, the ACKEN bit is automatically cleared, baud rate generator is turned off, and MSSP mod enters idle mode.

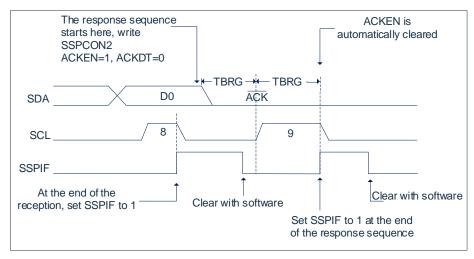
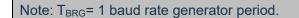


Fig 15-13: times series for ACK



# 15.3.9.1 WCOL Status Indication Bit

If the user writes SSPBUF while the ACK sequence is in progress, WCOL will be set to 1 and the contents of the buffer will remain unchanged (no write operation has occurred).



# 15.3.10 Stop Condition

At the end of receive/transmit, by setting the enable bit of the stop sequence, PEN (SSPCON2 register), the SDA pin will generate a stop bit. At the end of receive/transmit, the SCL pin will remain low after the falling edge of the 9th clock Level. When the PEN bit is 1, the master control device sets SDA low. When the SDA line is sampled low, the baud rate generator is reloaded with the value and counts down to 0. When the baud rate generator times out, The SCL pin is pulled to a high level, and after a TBRG (baud rategenerator counts back to zero), SDA pin is pulled to a high level again. When SDA pin is cleared and the SSPIF bit is set to 1. After a TBRG period, the PEN bit is cleared and the SSPIF bit is set to 1.

# 15.3.10.1 WCOL Status Indication

If the user attempts to write SSPBUF during the stop sequence, the WCOL bit will be set to 1, and the contents of the buffer will not change (no write operation has occurred).

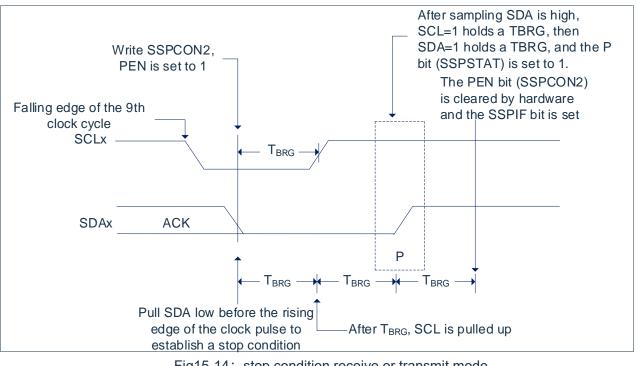


Fig15-14: stop condition receive or transmit mode

Note: T<sub>BRG</sub>=1 baud rate generator period.



### 15.3.11 Clock Arbitration

If during any receive, transmit, or repeated start/stop conditions, the master device pulls up the SCL pin (allowing the SCL pin to float high), clock arbitration will occur. If the SCL pin is allowed to float high, the baud rate generator (BRG) will pause counting until the SC L pin is actually sampled high. When the SCL pin is sampled high, the baud rate generator will be reloaded with the contents of SSPADD<6:0> and start counting. This can ensure that when the external device pulls the clock low, the SCL always maintains high for at least one BRG full return period.

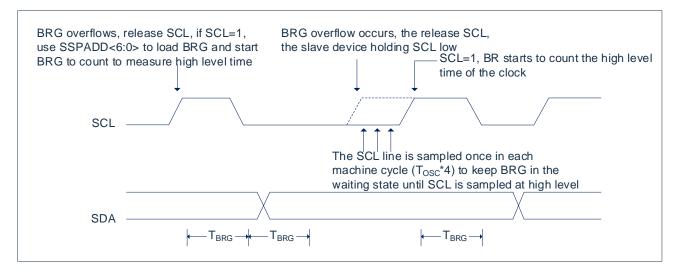


Fig 15-15: clock arbitration in master control transmit mode

#### 15.3.12 Multi Master Mode

In multi-master mode, it can be determined when the bus is free by generating interrupt when the start and stop conditions are detected. The stop (P) bit and the start (S) bit are cleared when reset or disable MSSP mod. When the P bit is set to 1, you can get control of the I<sup>2</sup>C bus; otherwise, the bus is in an idle state, and the P and S bits are cleared. When the bus is busy, if a stop condition occurs, an interrupt will be generated (if MSSP interrupt is allowed).

When working in multi-master mode, you must monitor the SDA line for arbitration to see if the signal level is the expected output level. This check is done by hardware, and the result is placed in the BCLIF bit.

Arbitration may fail under the following conditions:

- address transmission
- Start condition

- data transmission
- Repeated start condition

ACK conditions



# 15.3.13 Multi Master Communication, Bus Conflict and Bus Arbitration

Multi-master mode is supported by bus arbitration. When the master device outputs the address/data bit to the SDA pin, if one master device outputs 1 on SDA by floating the SDA pin to high level, and the other master device outputs 0, bus arbitration will occur. If the expected data on the SDA pin is 1, and the data actually sampled on the SDA pin is 0, a bus conflict has occurred. The master device will set the bus conflict interrupt flag bit BCLIF to 1, and reset the I<sup>2</sup>C port to idle state.

If a bus conflict occurs during the transmit process, the transmit stops, the BF flag bit is cleared, the SDA and SCL lines are pulled high, and SSPBUF is allowed to be written. After the bus conflict interrupt service program is executed, if the I<sup>2</sup>C bus is free, user can resume communication by issuing a start condition. If a bus conflict occurs during the start, repeated start, stop, or response condition, the condition is aborted, the SDA and SCL lines are pulled high, and the corresponding control bit in the SSPCON2 register is cleared. After executing the bus conflict interrupt service program, if the I<sup>2</sup>C bus is free, the user can resume communication by issuing a start condition. The master device will continue to monitor SDA and SCL pin. If a stop condition occurs, the SSPIF bit will be set to 1. No matter what bus occurs What is the progress of the transmit during conflict, writing SSPBUF will start transmitting data from the first data bit.

In multi-master mode, the interrupt can be generated when the start and stop conditions are detected to determine when the bus is free. When the P bit is set to 1, you can obtain control of the I2C bus, otherwise the bus is free and the S and P bits are cleared.

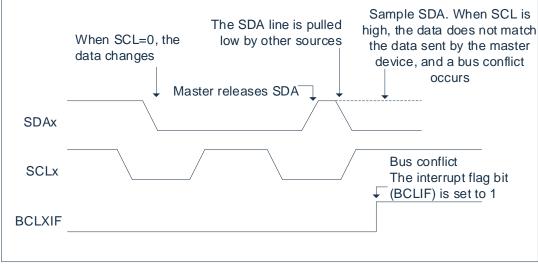


Fig 15-16: time series for transmit and ACK bus conflict



### 15.3.14 Slave Mode

In slave mode, SCL pin and SDA pin must be configured as input (TRISC<7:6> is set to 1). When needed (such as from the transmitter), the MSSP mod will use output data to rewrite the input state.

When the address matches or the data transmitted after the address matches is received, the hardware will automatically generate an acknowledge (ACK) pulse, and load the data received in the SSPSR register at the time into the SSPBUF register.

As long as one of the following conditions is met, MSSP mod will not generate this ACK pulse:

-The buffer full flag bit BF (SSPCON register) is 1 before the received data to be transmitted.

-Before receiving the transmitted data, the overflow flag bit SSPOV (SSPCON register) has been set 1.

In this case, the value of SSPSR register will not be loaded into SSPBUF, but the SSPIF bit of PIR1 register will be set to 1. The BF bit is cleared by reading the SSPBUF register, and the SSPOV bit is cleared by software.

To ensure normal operation, SCL clock input must meet the minimum high-level time and minimum lowlevel time requirements.



#### 15.3.14.1 Addressing

Once MSSP mod is enabled, it will wait for the start condition to be generated. After the start condition occurs, 8 bits of data are shifted into the SSPSR register. All input bits are sampled on the rising edge of the clock (SCL) line. RegisterSSPSR<7:1> The value will be compared with the value of the SSPADD register. The comparison is performed on the falling edge of the 8th clock pulse (SCL). If the address matches and the BF bit and SSPOV bit are zero, the following events will occur:

- The value of SSPSR register is loaded into SSPBUF register.
- The buffer full flag bit BF is set to 1.
- Generate ACK pulse.
- On the falling edge of the 9th SCL pulse, the MSSP interrupt flag bit SSPIF of the PIR1 register is set to 1 (interrupt is generated if interrupt is allowed)

# 15.3.14.2 Receive

When the R/W bit of the address byte is cleared and an address match occurs, the R/W bit of the SSPSTAT register is cleared. The received address is loaded into the SSPBUF register.

When there is an address byte overflow condition, an acknowledge pulse (ACK) will not be generated. The overflow condition means that the BF bit (SSPSTAT register) is set to 1, or the SSPOV bit (SSPCON register) is set to 1. Each data transmission byte will generate an MSSP interrupt. The interrupt flag bit SSPIF of the PIR1 register must be cleared by software. The SSPSTAT register is used to determine the status of the byte.



#### 15.3.14.3 Transmit

When the R/W bit of the received addressbyte is 1 and an address match occurs, the R/W bit of the SSPSTAT register is 1. The received address is loaded into the SSPBUF register. The ACK pulse is transmitted on the 9th bit while the SDA pin remains low. The transmitted data must be loaded into the SSPBUF register and also into the SSPSR register. Then the CKP bit (SSPCON register) should be set to 1 to enable the SCL pin. Before transmitting another clock pulse, the master control device must monitor the SCL pin. The slave device can suspend the data transmission with the master control device by extending the clock. 8 data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high level.

Each byte of data transmission will generate an MSSP interrupt. The SSPIF flag bit must be clear through software, and the SSPSTAT register is used to determine the status of the byte. The SSPIF bit is set at the falling edge of the 9th clock pulse. The ACK pulse from the main receiver will be latch on the rising edge of the 9th pulse of SCL input. If the SDA line is high (no ACK), then the data transfer has been completed. In this case, if the slave device latches the ACK, reset the slave logic (Reset SSPSTAT register), while the slave device monitors the appearance of the next start bit. If the SDA line is low (ACK), then the data to be transmitted must be loaded into the SSPBUF register, which will also load the SSPSR register. CKP should be set 1 to enable SCL.

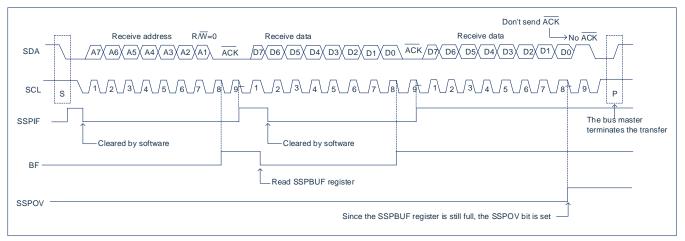


Fig 15-17: Time series for I<sup>2</sup>C<sup>™</sup> slave mode receive (7-bit address)



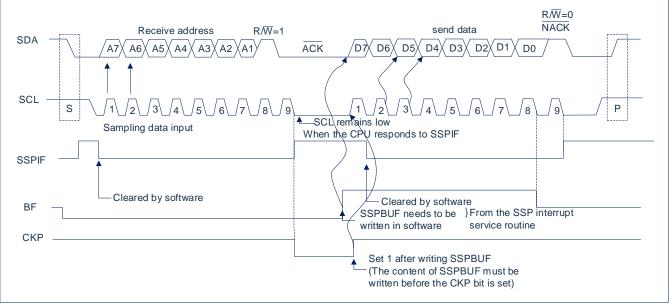


Fig 15-18: I<sup>2</sup>C<sup>™</sup> slave mode transmit (7-bit address)



# 15.3.15 SSP Masking Register

In I<sup>2</sup>C slave mode, the SSP mask (SSPMSK) register is used to mask the value in the SSPSR register under the address compare operation. A bit of 0 in the SSPMSK register will make the corresponding bit in the SSPSR register a "don't care".

This register is reset to all 1s when any reset condition occurs. Therefore, it has no effect on the standard SSP operation before writing the mask value. The register must be initialized before selecting the I<sup>2</sup>C slave mode by setting the SSPM<3:0> bits. This register can only be accessed after the appropriate mode is selected through the SSPM<3:0> bits of SSPCON.

The SSP masking register is valid in the following situations:

- 7-bit address mode: perform address compare with A<7:1>.
- 10-bit address mode: only perform address compare with A<7:0>

SSP masking is invalid during the period from receive to the first (high) byte of address.

93H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
SSPMSK	MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0 <sup>(2)</sup>
read/write	R/W							
Reset value	1	1	1	1	1	1	1	1

SSPMSK: SSP masking register (93H)<sup>(1)</sup>

MSK<7:1>: Mask bit.

not used.

 1= Bit n of the received address is compared with SSPADD<n> to detect the match of the I<sup>2</sup>C address.
 0= Bit n of the received address is not used to detect I<sup>2</sup>C address matching.

Bit 0

Bit7~Bit1

Note:

1) When the SSPCON bit SSPM<3:0> = 1001, any read or write operation to the SSPADDSFR

- address is performed through the SSPMSK register.
- 2) In all other SSP modes, this bit is invalid

# 15.3.16 Operation Under Sleep Mode

In sleep mode, I<sup>2</sup>C mod cannot be used.

# 15.3.17 Effect of Reset

Reset will disable MSSP mod and terminate the current transmission.



# 16. Program EEPROM and Program Memory Control

# 16.1 General

The devices in this series have 8K words of program memory, the address range is from 000h to 1FFFh, which is read-only in all address ranges; the device has a 128-byte program EEPROM, and the address range is 0h to 07Fh, which is available in all address ranges. It can be read/write.

These memories are not directly mapped to the register file space, but indirectly addressed through the special function register (SFR). A total of 6 SFR registers are used to access these memories:

- EECON1
- EECON2
- EEDAT
- EEDATH
- EEADR
- EEADRH

When accessing the program EEPROM, the EEDAT register stores 8-bit read/write data, and the EEADR register stores the address of the program EEPROM unit being accessed.

When accessing the program memory of the device, the EEDAT and EEDATH register form a double byte word to save the 16-bit data to be read, and the EEADR and EEADRH register form a double byte word to save the 13-bit EEPROM cell address to be read.

Program memory allows reading in units of bytes. Program EEPROM allows byte read/write. A byte write operation can automatically erase the target cell and write new data (erase before writing).

The writing time is controlled by the on-chip timer. The writing and erasing voltages are generated by the on-chip charge pump, which is rated to work within the voltage range of the device for byte or word operations.

When the device is protected by code, the CPU can still continue to read/write the program EEPROM and program memory. When the code is protected, the device programmer will no longer be able to access the program EEPROM or program memory.

#### Note:

- Program memory refers to ROM space, that is, the space where instructions code is stored, which can only be read; Program EEPROM is a space for storing user data, which can be read/write.
- 2) The normal writing voltage range of program EEPROM is 3.0V~5.5V, writing current is 20mA@VDD=5V.



# 16.2 Related Register

# 16.2.1 EEADR and EEADRH Register

The EEADR and EEADRH registers can address up to 128 bytes of program EEPROM or up to 8K bytes of program memory.

When the program memory address value is selected, the high byte of the address is written into the EEADRH register and the low byte is written into the EEADR register. When the program EEPROM address value is selected, only the low byte of the address is written into the EEADR register.

# 16.2.2 EECON1 and EECON2 Register

EECON1 is the control register to access the program EEPROM.

The control bit EEPGD determines whether to access program memory or program EEPROM. When this bit is cleared, as with reset, any subsequent operations will be performed on the program EEPROM. When this bit is set to 1, any subsequent operations will be performed on the program memory. Program memory is read-only.

The control bits RD and WR start reading and writing respectively. Software can only set these bits to 1 and cannot be cleared. After the read or write operation is completed, they are cleared by hardware. Since the WR bit cannot be cleared by software, it can be used to avoid accidentally terminating write operations prematurely.

-When WREN is set to 1, the program EEPROM is allowed to be written. When power is on, the WREN bit is cleared. When the normal write operation is LVR reset or WDT timeout reset interrupt, the WRERR bit will be set to 1. In these cases, after reset, the user can check the WRERR bit and rewrite the corresponding unit.

-When the write operation is completed, the interrupt flag bit EEIF in the PIR1 register is set to 1. This flag bit must be cleared by software.

EECON2 is not a physical register. Reading result of EECON2 is all 0s.

The EECON2 register is only used when executing the program EEPROM write sequence.

10CH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
EEDAT	EEDAT7	EEDAT6	EEDAT5	EEDAT4	EEDAT3	EEDAT2	EEDAT1	EEDAT0
read/write	R/W							
Reset value	Х	Х	Х	Х	Х	Х	Х	Х

# EEPROM data register EEDAT (10CH)

Bit7~Bit0 EEDAT<7:0>: To read or write the lower 8 bits of data from the program EEPROM, or read the lower 8 bits of data from the program memory.

#### EEPROM address register EEADR (10DH)

	-		```					
10DH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
EEADR	EEADR7	EEADR6	EEADR5	EEADR4	EEADR3	EEADR2	EEADR1	EEADR0
read/write	R/W							
Reset value	0	0	0	0	0	0	0	0

#### Bit7~Bit0

EEADR<7:0>: Specify the lower 8 bits of address for program EEPROM read/write operations, or the lower 8 bits of address for program memory read operations.



EEPROM data register EEDATH (*	10EH)	

10EH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
EEDATH	EEDATH7	EEDATH6	EEDATH5	EEDATH4	EEDATH3	EEDATH2	EEDATH1	EEDATH0
read/write	R/W							
Reset value	Х	Х	Х	Х	Х	Х	Х	Х

Bit7~Bit0 EEDATH<7:0>: The upper 8 bits of data read from the program EEPROM/program memory.

### EEPROM address register EEADRH (10FH)

10FH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
EEADRH				EEADRH4	EEADRH3	EEADRH2	EEADRH1	EEADRH0
read/write				R/W	R/W	R/W	R/W	R/W
reset value				0	0	0	0	0

Bit7~Bit5 not used, read 0.

Bit4~Bit0 EEADRH<4:0>: Specify the upper 5 address of the program memory read operation.

## EEPROM control register EECON1 (11BH)

11BH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
EECON1	EEPGD		EETIME1	EETIME0	WRERR	WREN	WR	RD
read/write	R/W		R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0		0	0	Х	0	0	0

Bit7	EEPGD:	Program/program EEPROMselection bit;
	1=	Operate program memory;
	0=	Operate program EEPROM.
Bit6	not used	
Bit5~Bit4	EETIME[1:0]	Maximum programming waiting time; (For more EETIME information, please refer
	00=	1.25ms
	01=	2.5ms (VDD=4.0~5.5V, suggested TEMP=0~85℃)
	10=	5ms
	11=	10ms (suggested other than 2.5ms)
Bit3	WRERR:	EEPROM error flag bit;
	1=	Write error (any WDT reset or undervoltage reset during normal operation, or the time
		set by EETIME is up but the self-check has not been successful);
	0=	Write complete.
Bit2	WREN:	EEPROM write enable bit;
	1=	Enable write period;
	0=	Disable write memory.
Bit1	WR:	Write control bit;
	1=	Start write period (Once the write operation is completed, this bit is cleared by
		hardware, and the WR bit can only be set to 1, but not cleared by software);;
	0=	Write period complete.
Bit0	RD:	Read control bit;
2.10	1=	Start the memory read operation (the RD is cleared by hardware, and the RD bit can
		only be set to 1, but not cleared by software);
	0	
	0=	Not start memory read operation.



# 16.3 Read Program EEPROM

To read the program EEPROM cell, the user must write the address to the EEADR register, clear the EEPGD control bit of the EECON1 register, and then set the control bit RD to 1. Once the read control bit is set, the program EEPROM controller will use the second instruction period to read data. This will cause the second instruction following the "SETB EECON1, RD" instruction to be ignored (1). In the next clock period, the corresponding address value of the program EEPROM will be latched into the EEDAT register In, the user can read these two registers in subsequent instructions. EEDAT will save this value until the next time the user reads or writes data to the unit.

Note: The two instructions after the program memory read operation must be NOP. This prevents the user from executing dual period instructions on the next instruction after the RD position is 1.

#### example: read program EEPROM

EEPDATA_READ:		
LD	A,RADDR	; Put the address to be read into the EEADR register
LD	EEADR,A	
CLRB	EECON1,EEPGD	;access data memory
SETB	EECON1,RD	;start reading
NOP		
NOP		
LD	A,EEDAT	;read and load data to ACC
LD	RDATA,A	
EEPDATA_READ_BACK:		
RET		



# **16.4 Write Program EEPROM**

To write a program EEPROM storage unit, the user should first write the unit's address to the EEADR register and write data to the EEDAT register. Then the user must start writing each byte in a specific order.

If you do not follow the following instructions exactly (that is, first write 55h to EECON2, then write Aah to EECON2, and finally set the WR bit to 1) to write each byte, the write operation will not be started. Interrupt should be disabled in this code.

In addition, the WREN bit in EECON1 must be set to 1 to enable write operations. This mechanism can prevent EEPROM from being written by mistake due to code execution errors (abnormal) (ie program runaway). When not updating EEPROM, the user should always keep the WREN bit cleared. The WREN bit cannot be cleared by hardware.

After a write process is started, clearing the WREN bit will not affect the write period. Unless the WREN bit is set, the WR bit will not be set to 1. When the write period is completed, the WR bit is cleared by hardware and the EE write is completed interrupt flag bit (EEIF) is set to 1. user can allow this interrupt or query this bit. EEIF must be cleared by software.

Note: During the writing of the program EEPROM, the CPU will stop working, the CLRWDT command must be executed before the writing operation starts to avoid WDT overflow to reset the chip during this period.

EEPDATA_V	VRITE:		
	LD	A,WADDR	; Put the address to be written into the EEADR register
	LD	EEADR,A	
	LD	A,WDATA	; put the data to be written to the EEDAT register
	LD	EEDAT,A	
	CLRWDT		
	CLR	EECON1	
	SETB	EECON1,EETIME0	
	SETB	EECON1,EETIME1	;EE programming time 10ms, user-defined
	CLRB	EECON1,EEPGD	;access data memory
	SETB	EECON1,WREN	;enable write period
	CLRB	F_GIE_ON	;save interrupt enabled status
	SZB	INTCON,GIE	
	SETB	F_GIE_ON	
	CLRB	INTCON,GIE	;disable interrupt
	SZB	INTCON,GIE	;ensure interrupt is disabled
	JP	\$-2	
	LDIA	055H	
	LD	EECON2,A	
	LDIA	0AAH	
	LD	EECON2,A	
	SETB	EECON1,WR	;start writing
	NOP		,
	NOP		
	-	159 / 195	V/1 E 2

#### example: write program EEPROM



CLRWDT		
CLRB	EECON1,WREN	;write complete, turn off write enable bit
SZB	F_GIE_ON	;restore interrupt enabled status
SETB	INTCON,GIE	
SNZB	EECON1,WRERR	;check EEPROM write
JP	EEPDATA_WRITE_BACH	<
SZDECR	WERR_C	; Exit when the count expires, user-defined
JP	EEPDATA_WRITE	;rewrite when EEPROM write error
EEPDATA_WRITE_BACK		
RET		



# 16.5 Read Program Memory

To read the program memory unit, the user must write the high and low bits of the address to the EEADR and EEADRH registers respectively, set the EEPGD bit of EECON1register to 1, and then set the control bit RD to 1. Once the read control bit is set, the program memory controller will use the second instructionsperiod to read data. This will cause the second instructions following the "SETB EECON1, RD" instructions to be ignored. In the next clock period, the value of the corresponding address of the program memory will be latched to EEDAT In the EEDATH register, the user can read these two registers in the subsequent instructions. The EEDAT and EEDATH register will save this value until the next time the user reads or writes data to the unit.

#### Note:

1) The two instructions after the program memory read operation must be NOP. This prevents the user from executing double period instructions in the next instruction after the RD position is 1.

2) If the WR bit is 1 when EEPGD=1, it will reset to 0 immediately without performing any operation.

example: read flash program	n memory	
LD	A,RADDRL	; Put the address to be read into the EEADR register
LD	EEADR,A	
LD	A,RADDRH	; Put the high bit of the address to be read into EEADRHregister
LD	EEADRH,A	
SETB	EECON1,EEPGD	;select to operate on program memory
SETB	EECON1,RD	;enable read
NOP		
NOP		
LD	A,EEDAT	;save read data
LD	RDATL,A	
LD	A,EEDATH	
LD	RDATH,A	

#### Write Program Memory 16.6

program memoryis read only, cannot be written.



# **16.7 Precautions on Program EEPROM**

# 16.7.1 **Programming Time for Program EEPROM**

The program EEPROM programming time is not fixed. The time required to program different data varies from 100us to 10ms. The EETIME bit of the EECON1 register determines the maximum time for program EEPROM programming. Program EEPROM mod built-in self-calibration during the programming process, if the self-verification is successful or the time set by EETIME has expired, the write operation will be terminated when one of the conditions is met. During the programming, the CPU stops working, the peripherals mod works normally, and the program needs to be well dealt with accordingly.

# 16.7.2 Number of Times for Programming EEPROM

The number of times of the program EEPROM are related to the programming time set by EETIME, as well as voltage and temperature. For details, please refer to the following diagram.

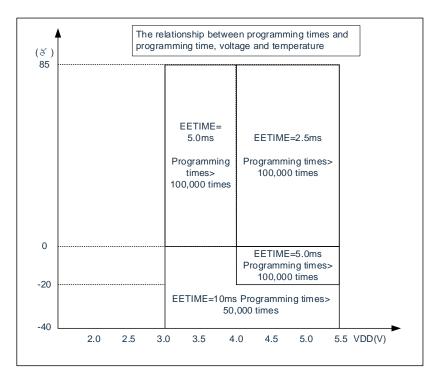


Fig 16-1 The relationship between program EEPROM programming times and programming time, voltage and temperature

# 16.7.3 Write Verification

According to specific applications, good programming habits generally require verification of the value written into the program EEPROM against the expected value.



# 16.7.4 **Protection to Avoid Writing Wrongly**

In some cases, the user may not want to write data to the program EEPROM. In order to prevent accidental writing of EEPROM, various protection mechanisms are embedded in the chip. The WREN bit is cleared when the power is turned on. Moreover, the power-on delay timer (the delay time is 18ms) Will prevent writing to the EEPROM.

The start sequence of the write operation and the WREN bit will work together to prevent false write operations in the following situations:

- Undervoltage
- Power glitch
- Software failure



# 17. LVD Low Voltage detection

# 17.1 LVD Mod General

CMS79FT73x series of MCUhave a low-voltage detection function, which can be used to monitor the power supply voltage. If the power supply voltage is lower than the set value, an interrupt signal can be generated; the program can read the LVD output flag bit in real time.

# 17.2 LVD Related Register

There is 1 register related to LVD mod.

LVD control register LVDCON (185H)

185H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
LVDCON	LVD_RES	_	—	_	l	_VD_SEL[2:0	]	LVDEN
R/W	R	—	—	_	R/W	R/W	R/W	R/W
Reset value	Х	_	_	_	0	0	0	0

Bit7	LVD_RES: 0= 1=	I
Bit6~Bit4	not used	
Bit3~Bit1	LVD_SEL[2:0]:	LVD voltage selection
	=000	2.2V;
	001=	2.4V;
	010=	2.7V;
	011=	3.0V;
	100=	3.3V;
	101=	3.7V;
	110=	4.0V;
	111=	4.3V;
Bit0	LVDEN:	LVD enable bit
	0=	disable;
	1=	enable;

# 17.3 LVD Operation

By setting the LVD voltage value in the LVDCON register, after enabling LVDEN, when the power supply voltage is lower than the set voltage value, the LVD\_RES bit in the LVDCON register is set high. After LVD mod is enabled, it takes a delay of 1ms to be able to read the LVD\_RES bit, because the internal has done filtering processing to reduce the frequent fluctuation of the LVD output result when the VLVD voltage is near.

LVD mod has its own interrupt flag bit. When the relevant interrupt enable bit is set, and the power supply voltage is lower than the set voltage value, LVD interrupt will be generated, the interrupt flag bit LVDIF will be set to 1, and interrupt generation. LVD is also possible used for interrupt wake up mode.



# 18. DIV Hardware Divider

# **18.1 Hardware Divider General**

CMS79FT73x series of MCU have a built-in hardware divider, 32-bit dividend, 16-bit divisor, and no remainder output.

Set the dividend through DIVE3, DIVE2, DIVE1 and DIVE0 register. These four registers can only be written but cannot be read. The divisors can be set through DIVS1 and DIVS0 register. These two registers can be read/write. The quotient of the operation is stored in DIVQ3, DIVQ2 In DIVQ1 and DIVQ0 register, these four registers can only be read but cannot be written. DIVEx and DIVQx share a register address, enableD IVEN, and wait for the CAL\_END bit to be 1 before reading the quotient.

# **18.2 Hardware Divider Related Register**

There are 11 registers related to the divider mod, namely DIVCON, DIVE3, DIVE2, DIVE1, DIVE0, DIVS1, DIVS0, DIVQ3, DIVQ2, DIVQ1 and DIVQ0.

		(	-)					
189H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVCON	DIVEN	CAL_END	_	_	_	_	_	DIV_CLK
R/W	R/W	R	_	—	_	_	—	R/W
Reset value	0	1	_	_	_	_	—	0

DIV control register DIVCON (189H)

Bit7	DIVEN:	DIV enable bit
	0=	Disable;
	1=	enable.
Bit6	CAL_END:	Result flag bit
	0=	Calculation in progress;
	1=	Calculation end already or not started yet.
Bit5~Bit1	not used	
Bit0	DIV_CLK:	DIV clock scaling selection bit
	0:	Fsys/2;
	1:	Fsys/4;

#### Divider register DIVE3 (18CH)

18CH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVEH1								
R/W	W	W	W	W	W	W	W	W
reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Divider DIVE[32:24]



### Divider register DIVE2 (18DH)

18DH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVEH0								
R/W	W	W	W	W	W	W	W	W
reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Divider DIVE[23:16]

### Divider register DIVE1 (18EH)

18EH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVEM								
R/W	W	W	W	W	W	W	W	W
reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Divider DIVE[15:8]

# Divider register DIVE0 (18FH)

18FH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVEL								
R/W	W	W	W	W	W	W	W	W
reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Divider DIVE[7:0]

### Divisor register DIVS1 (187H)

187H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVSH								
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Divisor DIVS[15:8]

#### Divisor register DIVS0 (188H)

188H	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVSL								
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Divisor DIVS[7:0]



### Quotient register DIVQ3 (18CH)

18CH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVQH1								
R/W	R	R	R	R	R	R	R	R
reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Quotient DIVQ[32:24]

# Quotient register DIVQ2 (18DH)

18DH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVQH0								
R/W	R	R	R	R	R	R	R	R
reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Quotient DIVQ[23:16]

# Quotient register DIVQ1 (18EH)

18EH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVQM								
R/W	R	R	R	R	R	R	R	R
reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Quotient DIVQ[15:8]

### Quotient register DIVQ0 (18FH)

18FH	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
DIVQL								
R/W	R	R	R	R	R	R	R	R
Reset value	0	0	0	0	0	0	0	0

Bit7~Bit0 Quotient DIVQ[7:0]



# 19. Touch Button

# **19.1 Touch Button Mod General**

The touch detection mod is an integrated circuit designed to realize a human touch interface. It can replace mechanical touch buttons to achieve a waterproof and dustproof, sealed and isolated, sturdy and beautiful operation interface.

technical parameter:

- 1-16 buttons are optional, all I/O can be configured as touch channels
- No need for external touch capacitance
- High anti-interference performance, can easily pass static 10V, dynamic 3V conduction test

# **19.2 Precautions for Touch Button Mod**

- The ground wire of the detection part of the touch button should be separately connected to an independent ground, and another point is connected to the common ground of the whole machine.
- Avoid high-voltage, high-current, high-frequency operation of the motherboard and the touch circuit board. If it is unavoidable, try to stay away from the area of the high-voltage current or add shielding on the motherboard.
- The connection between the sensor pad and the touch chip should be as short and thin as possible. If the PCB process allows it, try to use a line width of 0.1mm.
- The connection between the sensor panel and the touch chip should not cross the signal line with strong interference and high frequency.
- Do not use other signal lines around 0.5mm from the sensor panel to the touch chip.



# 20. Electrical Parameter

# 20.1 Limit Parameter

Supplying voltage	GND-0.3V~GND+6.0V
storage temperature	50°C~125°C
working temperature	40°C~85°C
port input voltage	GND-0.3V~VDD+0.3V
Maximum source current for all ports	200mA
Maximum sink current for all ports	150mA

Note: If the device operating conditions exceed the above "limit parameters", it may cause permanent damage to the device. The above values are only the maximum value of the operating conditions. We do not recommend that the device operate outside the range specified in this specification. The device works for a long time. Under extreme conditions, its stability will be affected.



# 20.2 DC Feature

(VDD=5V,  $T_A$ = 25°C, Unless otherwise indicated)

Symbol	parameter	٦	est condition	Min.	Typical	Max.	unit
Symbol	parameter	VDD	condition	value	value	value	um
VDD	Working volates		F <sub>SYS</sub> =16MHz	2.6		5.5	V
VDD	Working volatge		Fsys=8MHz	2.0		5.5	V
		5V	Fsys=16MHz		3.8		mA
DD	Working current	3V	F <sub>SYS</sub> =16MHz		3.0		mA
IDD	Working current	5V	Fsys=8MHz		3.0		mA
		3V	Fsys=8MHz		2.3		mA
ISTB	Static current	5V			0.1	2	μA
ISTB		3V			0.1	1	μA
VIL	Low level input voltage					0.3VDD	V
VIH	High level input voltage			0.7VDD			V
Vон	High level output voltage		No load	0.9VDD			V
Vol	Low level output voltage		No load			0.1VDD	V
VEEPROM	EEPROM mod w/r voltage			3.0		5.5	V
R <sub>PH</sub>	pull up resistor	5V	V <sub>O</sub> =0.5VDD		35		KΩ
INPH	resistance	3V	Vo=0.5VDD		63		KΩ
RPD	pull down resistor	5V	Vo=0.5VDD		35		KΩ
INPD	resistance	3V	Vo=0.5VDD		63		KΩ
Let 1	Output port source	5V	V <sub>OL</sub> =0.3VDD		60		mA
IOL1	current (normal I/O port)	3V	Vol=0.3VDD		25		mA
	Output port drain	5V	V <sub>OH</sub> =0.7VDD		-20		mA
Іон1	current (normal I/O port)	3V	V <sub>OH</sub> =0.7VDD		-9		mA
	Output port source	5V	V <sub>OL</sub> =0.3VDD		150		mA
IOL2	current (LED COM port)	3V	Vol=0.3VDD		70		mA
	Output port drain	5V	VOH=0.7VDD		-30		mA
Іон2	current (LED SEG port max. current)	3V	V <sub>OH</sub> =0.7VDD		-12		mA
	<i>(</i>	VDD=	2.5~5.5V TA=25°C	-1.5%	1.2	1.5%	V
$V_{BG}$	Internal reference voltage 1.2V		DD=2.5~5.5V ″ <sub>A</sub> = - 40~85°C	-2.0%	1.2	2.0%	V



# 20.3 ADC Feature

 $(T_A = 25^{\circ}C, Unless otherwise indicated)$ 

Symbol	parameter	Test condition	min. value	typical value	max. value	unit
VADC	ADC working voltage	F <sub>ADC</sub> =500kHz	2.7		5.5	V
lun e	ADC current	V <sub>ADC</sub> =5V,F <sub>ADC</sub> =500kHz			500	μA
IADC	ADC current	Vadc =3V,Fadc=500kHz			200	μA
Vadi	ADC inputvoltage	Vadc=5V,Fadc=250kHz	0		VADC	V
DNL	Differential nonlinearity error	Vadc=5V,Fadc=250kHz		±3		LSB
INL	Integral nonlinearity error	V <sub>ADC</sub> =5V,F <sub>ADC</sub> =250kHz		±4		LSB
TADC	ADC conversion time	-		49		TADCCLK

# 20.4 Power on Reset Feature

(T<sub>A</sub>= 25°C, Unless otherwise indicated)

symbol	parameter	Test condition	min. value	typical value	max. value	unit
tvdd	VDD increase rate	-	0.05			V/ms
VLVR2	LVR =2.0V	VDD=1.8~5.5V	1.9	2.0	2.1	V
V <sub>LVR3</sub>	LVR=2.6V	VDD=2.4~5.5V	2.5	2.6	2.7	V

# 20.5 AC Electrical Characteristics

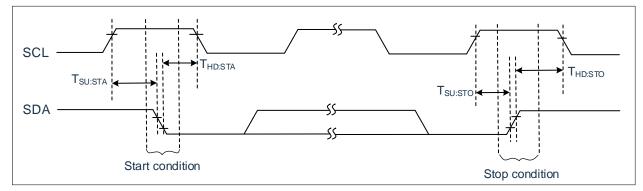
(T<sub>A</sub>=25°C, Unless otherwise indicated)

aymbol	parameter	Test	condition	min.	typical	max.	unit
symbol	parameter	VDD	condition	value	value	value	unit
ł	T <sub>WDT</sub> WDT reset time	5V	-		16		ms
IWDT		3V	-		32		ms
	EEPROM programming time	5V	F <sub>HSI</sub> =8MHz/16MHz			10	ms
TEEPROM		3V	F <sub>HSI</sub> =8MHz/16MHz			10	ms
		VDD=4.5~5.5V	Ta=25°C	-1.5%	8	+1.5%	MHz
		VDD=2.0~5.5V	Ta=25°C	-2%	8	+2%	MHz
		VDD=4.5~5.5V	Ta= - 40~85°C	-2.5%	8	+2.5%	MHz
F	Internal frequency	VDD=2.0~5.5V	Ta= - 40~85°C	-3.5%	8	+3.5%	MHz
F <sub>RC</sub>	stability	VDD=4.5~5.5V	Ta=25°C	-1.5%	16	+1.5%	MHz
		VDD=2.6~5.5V	Ta=25°C	-2%	16	+2%	MHz
		VDD=4.5~5.5V	Ta= - 40~85°C	-2.5%	16	+2.5%	MHz
		VDD=2.6~5.5V	Ta= - 40~85°C	-3.5%	16	+3.5%	MHz

Remark: Low-temperature specification values are not tested in mass production.



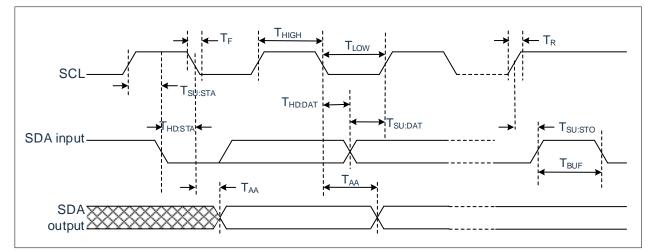
# 20.6 IIC Electrical Characteristics



Symbol	Charac	teristic	Condition	Min	Тур	Max	Unit
-	Start condition	100kHz Mode	Only relevant for repeated	4700	-	-	ns
T <sub>SU:STA</sub>	set-up time	400kHz Mode	start conditions	600	-	-	ns
Ŧ	Start condition	100kHz Mode	The first clock pulse is	4000	-	-	ns
Thd:sta	hold time	400kHz Mode	generated after this cycle	600	-	-	ns
Ŧ	Stop condition	100kHz Mode		4700	-	-	ns
Tsu:sto	set-up time	400kHz Mode		600	-	-	ns
Ŧ	Stop condition	100kHz Mode		4000	-	-	ns
T <sub>HD:STO</sub>	hold time	400kHz Mode		600	-	-	ns

Note: These parameters are characteristic values only and have not been tested.





# Figure 20-2: I<sup>2</sup>C<sup>™</sup> Bus Data Timing

Symbol	Character	istic	Condition	Min	Max	Unit
		100kHz Mode	The operating frequency of the device must not be lower than 4MHz	4.0	-	us
Тнідн	Clock high level time	400kHz Mode	The operating frequency of the device must not be lower than 16MHz	0.6	-	us
_		100kHz Mode	The operating frequency of the device must not be lower than 4MHz	4.7	-	us
TLOW	.ow Clock low level time	400kHz Mode	The operating frequency of the device must not be lower than 16MHz	1.3	-	us
		100kHz Mode		-	1000	ns
T <sub>R</sub>	SDA and SCL rising time	400kHz Mode	C <sub>B</sub> values are specified in the range of 10-400pF	20+0.1C <sub>B</sub>	300	ns
		100kHz Mode		-	300	ns
T <sub>F</sub>	SDA and SCL falling time	400kHz Mode	C <sub>B</sub> values are specified in the range of 10-400pF	20+0.1C <sub>B</sub>	300	ns
<b>-</b>	Start condition set-up	100kHz Mode	Only relevant for repeated start	4.7	-	us
Tsu:sta	time	400kHz Mode	conditions	0.6	-	us
T <sub>HD:STA</sub>	Start condition hold	100kHz Mode	The first clock pulse is	4.0	-	us
THD:STA	time	400kHz Mode	generated after this cycle	0.6	-	us
T <sub>HD:DAT</sub>	Data input hold time	100kHz Mode		2/F <sub>sys</sub>	-	us
I HU:DAI		400kHz Mode		2/F <sub>sys</sub>	1-2/F <sub>sys</sub>	us
Tours	Data input set-up time	100kHz Mode		2/F <sub>sys</sub>	-	us
Tsu:dat	Data input set-up time	400kHz Mode		2/F <sub>sys</sub>	-	us
Tsu:sto	Stop condition set-up	100kHz Mode		4.7	-	us



	time	400kHz Mode		0.6	-	us
т	Clock output valid	100kHz Mode		-	3.7-2/F <sub>sys</sub>	us
T <sub>AA</sub>	time	400kHz Mode		-	-	us
		100kHz Mode	The duration that the bus shall	4.7	-	us
TBUF	Bus idle time	400kHz Mode	remain idle for before a new transmission begins.	1.3	-	us
Св	Bus capaciti	ve load		-	400	pF

Note: These parameters are characteristic values only and have not been tested.



# 21. Instructions

# 21.1 Instructions Table

mnemonic		operation	instructio nsperiod	symbol
control-3		•		
NOP		Empty operation	1	None
STOP		Enter sleep mode	1	TO,PD
CLRWDT		Clear watchdog timer	1	TO,PD
Data trans	sfer-4	•		
LD	[R],A	Transfer content to ACC to R	1	NONE
LD	A,[R]	Transfer content to R to ACC	1	Z
TESTZ	[R]	Transfer the content of data memory data memory	1	Z
LDIA	i	Transfer I to ACC	1	NONE
logic ope	ration -16	5		
CLRA		Clear ACC	1	Z
SET	[R]	Set data memory R	1	NONE
CLR	[R]	Clear data memory R	1	Z
ORA	[R]	Perform 'OR' on R and ACC, save the result to ACC	1	Z
ORR	[R]	Perform 'OR' on R and ACC, save the result to R	1	Z
ANDA	[R]	Perform 'AND' on R and ACC, save the result to ACC	1	Z
ANDR	[R]	Perform 'AND' on R and ACC, save the result to R	1	Z
XORA	[R]	Perform 'XOR' on R and ACC, save the result to ACC	1	Z
XORR	[R]	Perform 'XOR' on R and ACC, save the result to R	1	Z
SWAPA	[R]	Swap R register high and low half byte, save the result to ACC	1	NONE
SWAPR	[R]	Swap R register high and low half byte, save the result to R	1	NONE
COMA	[R]	The content of R register is reversed, and the result is stored in ACC	1	Z
COMR	[R]	The content of R register is reversed and the result is stored in R	1	Z
XORIA	i	Perform 'XOR' on i and ACC, save the result to ACC	1	Z
ANDIA	i	Perform 'AND' on i and ACC, save the result to ACC	1	Z
ORIA	i	Perform 'OR' on i and ACC, save the result to ACC	1	Z
Shift oper	ration-8		II	
RRCA	[R]	Data memory rotates one bit to the right with carry, the result is stored in ACC	1	С
RRCR	[R]	Data memory rotates one bit to the right with carry, the result is stored in R	1	С
RLCA	[R]	Data memory rotates one bit to the left with carry, the result is stored in ACC	1	С
RLCR	[R]	Data memory rotates one bit to the left with carry, the result is stored in R	1	С
RLA	[R]	Data memory rotates one bit to the left without carry, and the result is stored in ACC	1	NONE
RLR	[R]	Data memory rotates one bit to the left without carry, and the result is stored in R	1	NONE
RRA	[R]	Data memory does not take carry and rotates to the right by one bit, and the result is stored in ACC	1	NONE
RRR	[R]	Data memory does not take carry and rotates to the right by one bit, and the result is stored in $\ensuremath{R}$	1	NONE
Increae/d	ecrease-	4		
INCA	[R]	Increment data memory R, result stored in ACC	1	Z
INCR	[R]	Increment data memory R, result stored in R	1	Z



mnemonic		operation	instructio nsperiod	symbol
DECA	[R]	Decrement data memory R, result stored in ACC	1	Z
DECR	[R]	Decrement data memory R, result stored in R	1	Z
Bit operati	ion-2			
CLRB	[R],b	Clear some bit in data memory R	1	NONE
SETB	[R],b	Set some bit in data memory R 1	1	NONE
look-up ta	ble-2			
TABLE	[R]	Read FLASH and save to TABLE_DATAH and R	2	NONE
TABLEA		Read FLASH and save to TABLE_DATAH and ACC	2	NONE
Math oper	ation-16			
ADDA	[R]	ACC+[R]→ACC	1	C,DC,Z,OV
ADDR	[R]	ACC+[R]→R	1	C,DC,Z,OV
ADDCA	[R]	ACC+[R]+C→ACC	1	Z,C,DC,OV
ADDCR	[R]	ACC+[R]+C→R	1	Z,C,DC,OV
ADDIA	i	ACC+i→ACC	1	Z,C,DC,OV
SUBA	[R]	[R]-ACC→ACC	1	C,DC,Z,OV
SUBR	[R]	[R]-ACC→R	1	C,DC,Z,OV
SUBCA	[R]	[R]-ACC-C→ACC	1	Z,C,DC,OV
SUBCR	[R]	[R]-ACC-C→R	1	Z,C,DC,OV
SUBIA	i	i-ACC→ACC	1	Z,C,DC,OV
HSUBA	[R]	ACC-[R]→ACC	1	Z,C,DC,OV
HSUBR	[R]	ACC-[R]→R	1	Z,C,DC,OV
HSUBCA	[R]	ACC-[R]- C →ACC	1	Z,C,DC,OV
HSUBCR	[R]	ACC-[R]- C →R	1	Z,C,DC,OV
HSUBIA	i	ACC-i→ACC	1	Z,C,DC,OV
Unconditio	onal tran	sfer -5	•	
RET		Return from subroutine	2	NONE
RET	i	Return from subroutine, save I to ACC	2	NONE
RETI		Return from interrupt	2	NONE
CALL	ADD	Subroutine call	2	NONE
JP	ADD	Unconditional jump	2	NONE
Condition	al transfe	er-8		
SZB	[R],b	If the b bit of data memory R is "0", skip the next instruction	1 or 2	NONE
SNZB	[R],b	If the b bit of data memory R is "1", skip the next instruction	1 or 2	NONE
SZA	[R]	data memory R is sent to ACC, if the content is "0", skip the next instruction	1 or 2	NONE
SZR	[R]	If the content of data memory R is "0", skip the next instruction	1 or 2	NONE
SZINCA	[R]	Add "1" to data memory R and put the result into ACC, if the result is "0", skip the next oneinstructions	1 or 2	NONE
SZINCR	[R]	Add "1" to data memory R, put the result into R, if the result is "0", skip the next instruction	1 or 2	NONE
SZDECA	[R]	Data memory R minus "1", the result is put into ACC, if the result is "0", skip the next instruction	1 or 2	NONE
SZDECR	[R]	Data memory R minus "1", put the result into R, if the result is "0", skip the next oneinstructions	1 or 2	NONE



# 21.2 Instructions Illustration

ADDA	[R]			
operation:	Add ACC to R, save the result to ACC			
period:	1			
Affected flag bit:	C, DC, Z, OV			
example:				
	LDIA	09H	;load 09H to ACC	
	LD	R01,A	;load ACC (09H) to R01	
	LDIA	077H	;load 77H to ACC	
	ADDA	R01	;execute: ACC=09H + 77H =80H	

ADDR	[R]				
operation:	Add ACC to	Add ACC to R , save the result to R			
period:					
Affected flag bit: example:	° C, DC, Z, OV				
	LDIA	09H	;load 09H to ACC		
	LD	R01,A	; load ACC (09H) to R01		
	LDIA	077H	; load 77H to ACC		
	ADDR	R01	;execute: R01=09H + 77H =80H		

ADDCA	

operation: period: affected fla bit: example:

Add ACC to C, save the result to ACC

lag	С,	DC,	Z,	OV	

1

[R]

LDIA	09H	; load 09H to ACC	
LD	R01,A	; load ACC (09H) to R01	
LDIA	077H	; load 77H to ACC	
ADDCA	R01	;execute: ACC= 09H + 77H + C=80H	(C=0)
		ACC= 09H + 77H + C=81H	(C=1)

ADDCR	[R]				
operation:	Add ACC to (	Add ACC to C, save the result to R			
period:	1				
affected flag bit:	C, DC, Z,	OV			
example:					
	LDIA	09H	; load 09H to ACC		
	LD	R01,A	; load ACC (09H) to R01		
	LDIA	077H	; load 77H to ACC		
	ADDCR	R01	;execute: R01 = 09H + 77H + C=80H (C=0) R01 = 09H + 77H + C=81H (C=1)		



ADDIA operation: period: affected flag bit: example:	i Add i to ACC, save the result to ACC 1 C, DC, Z, OV			
	LDIA	09H	; load 09H to ACC	
	ADDIA	077H	;execute: ACC = ACC (09H) + i (77H)=80H	
ANDA	[R]			
operation:	Perform'AN	ND'on register R and A	ACC, save the result to ACC	
period:	1			
affected flag	Z			
example:				
	LDIA	0FH	;load 0FH to ACC	
	LD	R01,A	;load ACC (0FH) to R01	
	LDIA	77H	;load 77H to ACC	
	ANDA	R01	;execute: ACC= (0FH and 77H)=07H	
ANDR	[R]			
operation:	Perform'AN	ND'on register R and A	ACC, save the result to R	
period:	1			
affected flag bit:	Z			
example:				
·	LDIA	0FH	; load 0FH to ACC	
	LD	R01,A	; load ACC (0FH) to R01	
	LDIA	77H	; load 77H to ACC	
	ANDR	R01	;execute:R01= (0FH and 77H)=07H	
ANDIA	i			
operation:	Perform'AN	ND'on i and ACC, save	e the result to ACC	
period:	1			
affected flag	Z			
example:				
	LDIA	0FH	; load 0FH to ACC	
	ANDIA	77H	;execute:ACC = (0FH and 77H)=07H	
CALL	add			
operation:	Call subrou	ıtine		
period:	2			
affected flag	none			
bit: example:	-			
overnpro.	CALL	LOOP	; Call the subroutine address whose name is defined as "LOOP"	
	UALL	2001	, can the subroutine address whose name is defined as LOOP	



# CLRA

operation: period: affected flag bit: example:	ACC clear 1 Z CLRA		;execute: ACC=0
CLR	[R]		
operation:	Register R cl	ear	
period: affected flag	1 Z		
example:	CLR	R01	;execute: R01=0
CLRB	[R],b		
operation:	Clear b bit or	n register R	
period: affected flag bit: example:	1 none		
	CLRB	R01,3	;execute: 3 <sup>rd</sup> bit of R01 is 0
CLRWDT			
operation: period: affected flag bit: example:	Clear watcho 1 TO, PD	log timer	
·	CLRWDT		;watchdog timer clear
СОМА	[R]		
operation:	Reverse regi	ster R, save the res	ult to ACC
period: affected flag bit: example:	1 Z		
	LDIA	0AH	;load 0AH to ACC
	LD	R01,A	;load ACC (0AH) to R01
	COMA	R01	;execute: ACC=0F5H



COMR operation: period: affected flag bit: example:	<b>[R]</b> Reverse register R, save the result to R 1 Z			
	LDIA	0AH	; load 0AH to ACC	
	LD	R01,A	; load ACC (0AH) to R01	
	COMR	R01	;execute: R01=0F5H	
DECA operation: period: affected flag bit: example:	<b>[R]</b> Decrement va 1 Z	alue in register , save	the result to ACC	
·	LDIA	0AH	;load 0AH to ACC	
	LD	R01,A	; load ACC (0AH) to R01	
	DECA	R01	;execute: ACC= (0AH-1)=09H	
DECR operation: period: affected flag bit: example:	1 Z LDIA	alue in register , save 0AH	; load 0AH to ACC	
	LD	R01,A	; load ACC (0AH) to R01	
	DECR	R01	;execute: R01= (0AH-1)=09H	
HSUBA operation: period: affected flag bit:	<b>[R]</b> ACC subtract 1 C,DC,Z,OV	t R, save the result to	ACC	
example:				
	LDIA LD LDIA HSUBA	077H R01,A 080H R01	; load 077H to ACC ; load ACC (077H) to R01 ; load 080H to ACC ;execute : ACC= (80H-77H)=09H	



#### HSUBR [R]

operation:	ACC subtract R, save the result to R		
period:	1		
affected flag bit:	C,DC,Z,OV		
example:			
	LDIA	077H	; load 077H to ACC
	LD	R01,A	; load ACC (077H) to R01
	LDIA	080H	; load 080H to ACC
	HSUBR	R01	;execute: R01= (80H-77H)=09H

HSUBCA	[R]		
operation:	ACC subtract C, save the result to ACC		
period:	1		
affected flag bit:	C,DC,Z,OV		
example:			
	LDIA	077H	; load 077H to ACC
	LD	R01,A	; load ACC (077H) to R01
	LDIA	080H	; load 080H to ACC
	HSUBCA	R01	;execute: ACC= (80H-77H-C)=09H (C=0) ACC= (80H-77H-C)=08H (C=1)

HSUBCR	[R]		
operation:	ACC subtrac	ct C, save the result to	R
period:	1		
affected flag bit:	C,DC,Z,OV		
example:			
	LDIA	077H	; load 077H to ACC
			: load ACC (077H) to P01

LD	R01,A	; load ACC (077H) to R01
LDIA	080H	; load 080H to ACC
HSUBC R	R01	;execute: R01= (80H-77H-C)=09H (C=0) R01= (80H-77H-C)=08H (C=1)

INCA	[R]		
operation:	Register R increment 1, save the result to ACC		
period:	1		
affected flag bit:	Z		
example:			
	LDIA	0AH	; load 0AH to ACC
	LD	R01,A	; load ACC (0AH) to R01
	INCA	R01	;execute: ACC= (0AH+1)=0BH



INCR	[R]		
operation:	Register R increment 1, save the result to R		
period:	1		
affected flag	Z		
bit: example:			
oxumpio.	LDIA	0AH	; load 0AH to ACC
	LD	R01,A	; load ACC (0AH) to R01
	INCR	R01	;execute: R01= (0AH+1)=0BH
JP	add		
operation:	Jump to ac	ld address	
period:	2		
affected flag bit:	none		
example:			
esternip te t	JP	LOOP	; jump to the subroutine address whose name is defined as "LOOP"
LD	A,[R]		
operation:	Load the v	alue of R to ACC	
period:	1		
affected flag	Z		
bit: example:			
oxumpio.	LD	A,R01	;load R01 to ACC
	LD	R02,A	;load ACC to R02, achieve data transfer from R01 $\rightarrow$ R02
		···-,··	
LD	[R],A		
operation:		alue of ACC to R	
period:	1		
affected flag	none		
bit: example:			
oxumpio.	LDIA	09H	;load 09H to ACC
	LD	R01,A	;execute: R01=09H
LDIA	i		
operation:	Load in to	ACC	
period:	1		
affected flag	none		
bit: example:			
example.	LDIA	0AH	;load 0AH to ACC
	2001	0, 11	



#### NOP

operation: period: affected flag bit: example:	Empty instru 1 none NOP NOP	ictions		
ORIA	i	i		
operation:		?' on I and ACC, save	the result to ACC	
period: affected flag	1			
bit:	Z			
example:	LDIA	0AH	; load 0AH to ACC	
	ORIA	030H	;execute: ACC = (0AH or 30H)=3AH	
ORA	[R]			
operation: period:	Perform 'OR	c' on R and ACC, save	e the result to ACC	
affected flag	Z			
bit: example:	2			
champic.	LDIA	0AH	; load 0AH to ACC	
	LD	R01,A	;load ACC (oAH) to R01	
	LDIA	30H	;load 30H to ACC	
	ORA	R01	;execute: ACC= (0AH or 30H)=3AH	
ORR	[R]			
operation:		c' on R and ACC, save	e the result to R	
period:	1			
affected flag bit:	Z			
example:				
	LDIA	0AH	; load 0AH to ACC	
	LD	R01,A	; load ACC (oAH) to R01	
	LDIA ORR	30H R01	; load 30H to ACC ;execute:R01= (0AH or 30H)=3AH	
			, one off ( of the off of the off) - on the off	



R	E	т
	_	

RET			
operation:	Return from	m subroutine	
period:	2		
affected flag			
bit:	none		
example:			
	CALL	LOOP	; Call subroutine LOOP
	NOP		; This statement will be executed after RET instructions return
			; others
LOOP:			
			;subroutine
	RET		;return
			,
DET			
RET	i		
operation:		n parameter from t	he subroutine, and put the parameter in ACC
period:	2		
affected flag bit:	none		
example:			
example:	CALL	LOOP	; Call subroutine LOOP
	NOP	2001	; This statement will be executed after RET instructions return
			; others
LOOP:			,001013
LOOP.			
	 DET	0511	;subroutine
	RET	35H	;return,ACC=35H
RETI			
operation:	Interrupt re	eturn	
period:	2		
affected flag bit:	none		
example:			
INT_START			;interrupt entrance
INT_START			
			;interrupt procedure
	RETI		;interrupt return
RLCA	[R]		
operation:	Register R	R rotates to the left	with C and savethe result into ACC
period:	1		
affected flag bit:	С		
example:			
-	LDIA	03H	;load 03H to ACC
	LD	R01,A	;load ACC to R01,R01=03H
	RLCA	R01,A R01	
	NECA	Rot	;operation result:ACC=06H (C=0); ACC=07H (C=1) C=0



RLCR	[R]	[R]		
operation:	Register R rotates one bit to the left with C, and save the result into R			
period:	1			
affected flag	С			
example:				
	LDIA	03H	; load 03H to ACC	
	LD	R01,A	; load ACC to R01,R01=03H	
	RLCR	R01	;operation result: R01=06H (C=0); R01=07H (C=1); C=0	
RLA	[R]			
operation:	Register R v	vithout C rotates to th	e left, and save the result into ACC	
period:	1			
affected flag bit:	none			
example:				
	LDIA	03H	; load 03H to ACC	
	LD	R01,A	; load ACC to R01,R01=03H	
	RLA	R01	;operation result: ACC=06H	
RLR	[R]			
<b>RLR</b> operation:		vithout C rotates to th	e left, and save the result to R	
		vithout C rotates to th	e left, and save the result to R	
operation: period: affected flag bit:	Register R v	vithout C rotates to th	e left, and save the result to R	
operation: period: affected flag	Register R v 1 none			
operation: period: affected flag bit:	Register R v 1 none LDIA	03H	; load 03H to ACC	
operation: period: affected flag bit:	Register R v 1 none LDIA LD	03H R01,A	; load 03H to ACC ; load ACC to R01,R01=03H	
operation: period: affected flag bit:	Register R v 1 none LDIA	03H	; load 03H to ACC	
operation: period: affected flag bit:	Register R v 1 none LDIA LD	03H R01,A	; load 03H to ACC ; load ACC to R01,R01=03H	
operation: period: affected flag bit: example:	Register R v 1 none LDIA LD RLR [ <b>R</b> ]	03H R01,A R01	; load 03H to ACC ; load ACC to R01,R01=03H	
operation: period: affected flag bit: example: <b>RRCA</b>	Register R v 1 none LDIA LD RLR [ <b>R</b> ]	03H R01,A R01	; load 03H to ACC ; load ACC to R01,R01=03H ;operation result:R01=06H	
operation: period: affected flag bit: example: <b>RRCA</b> operation:	Register R v 1 none LDIA LD RLR <b>[R]</b> Register R r	03H R01,A R01	; load 03H to ACC ; load ACC to R01,R01=03H ;operation result:R01=06H	
operation: period: affected flag bit: example: <b>RRCA</b> operation: period: affected flag	Register R v 1 none LDIA LD RLR <b>[R]</b> Register R r 1	03H R01,A R01	; load 03H to ACC ; load ACC to R01,R01=03H ;operation result:R01=06H	
operation: period: affected flag bit: example: <b>RRCA</b> operation: period: affected flag bit:	Register R v 1 none LDIA LD RLR <b>[R]</b> Register R r 1	03H R01,A R01	; load 03H to ACC ; load ACC to R01,R01=03H ;operation result:R01=06H	
operation: period: affected flag bit: example: <b>RRCA</b> operation: period: affected flag bit:	Register R v 1 none LDIA LD RLR <b>[R]</b> Register R r 1 C	03H R01,A R01 otates one bit to the r	; load 03H to ACC ; load ACC to R01,R01=03H ;operation result:R01=06H ight with C, and puts the result into ACC	



RRCR	[R]			
operation:		Register R rotates one bit to the right with C, and save the result into R		
period:	1			
' affected flag bit:	С			
example:				
example.	LDIA	03H	; load 03H to ACC	
	LDIA	R01,A	; load ACC to R01,R01=03H	
	RRCR	R01,A		
			;operation result:R01=01H (C=0); R01=81H (C=1); C=1	
RRA	[R]			
operation:		without C rotates one	bit to the right, and save the result into ACC	
period:	1		bit to the right, and save the result into ACC	
affected flag bit:	none			
example:				
	LDIA	03H	; load 03H to ACC	
	LD	R01,A	; load ACC to R01,R01=03H	
	RRA	R01	;operation result: ACC=81H	
			,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,	
RRR	[R]			
operation:	Register R	without C rotates one	bit to the right, and save the result into R	
period:	1			
affected flag bit:	none			
example:				
	LDIA	03H	; load 03H to ACC	
	LD	R01,A	; load ACC to R01,R01=03H	
	RRR	R01	;operation result: R01=81H	
SET	[R]			
operation:	Set all bits i	n register R as 1		
period:	1			
affected flag bit:	none			
example:	QET	<b>D</b> 01	concretion result: D01-0554	
	SET	R01	;operation result:R01=0FFH	
SETB	[R],b			
operation:	Set b bit in I	register R 1		
period:	1			
affected flag bit:	none			
example:				
	CLR	R01	;R01=0	
	SETB	R01,3	;operation result:R01=08H	



#### STOP

operation:	Enter sleep	
period:	1	
affected flag <sup>hit: example:</sup>	TO, PD	
	STOP	; The chip enters the power saving mode, the CPU and oscillator stop working, and the IO port keeps the original state

SUBIA	i			
operation:	ACC minus I	ACC minus I, save the result to ACC		
period:	1			
affected flag bit:	C,DC,Z,OV	C,DC,Z,OV		
example:				
	LDIA	077H	;load 77H to ACC	
	SUBIA	80H	;operation result: ACC=80H-77H=09H	

SUBA operation: period: affected flag bit: example:	<b>[R]</b> Register R minus ACC, save the result to ACC 1 C,DC,Z,OV		
	LDIA	080H	;load 80H to ACC
	LD	R01,A	;load ACC to R01, R01=80H
	LDIA	77H	;load 77H to ACC
	SUBA	R01	;operation result: ACC=80H-77H=09H
SUBR	[R]		
operation:	Register R m	inus ACC, save the re	esult to R
period:	1		
affected flag bit:	C,DC,Z,OV		
example:			
	LDIA	080H	; load 80H to ACC
	LD	R01,A	; load ACC to R01, R01=80H
	LDIA	77H	; load 77H to ACC

SUBR

R01

;operation result: R01=80H-77H=09H



SUBCA	[R]			
operation:	Register R	Register R minus ACC minus C, save the result to ACC		
period:	1			
affected flag	C,DC,Z,OV	/		
bit: example:				
oxampio.	LDIA	080H	; load 80H to ACC	
	LD	R01,A	; load ACC to R01, R01=80H	
	LDIA	77H	; load 77H to ACC	
	SUBCA	R01	;operation result: ACC=80H-77H-C=09H (C=0);	
			ACC=80H-77H-C=08H (C=1);	
SUBCR	[R]			
operation:	Register R	minus ACC minus	s C, save the result to ACC	
period:	1			
affected flag bit:	C,DC,Z,OV	/		
example:				
	LDIA	080H	; load 80H to ACC	
	LD	R01,A	; load ACC to R01,R01=80H	
	LDIA	77H	; load 77H to ACC	
	SUBCR	R01	;operation result: R01=80H-77H-C=09H (C=0)	
			R01=80H-77H-C=08H (C=1)	
	(D)			
SWAPA	[R]	high and low half	but our the course and thinks ACC	
operation:	-	nign and low hait	byte swap, the save result into ACC	
period: affected flag	1			
bit:	none			
example:				
	LDIA	035H	;load 35H to ACC	
	LD	R01,A	; load ACC to R01,R01=35H	
	SWAPA	R01	;operation result: ACC=53H	
SWAPR	[R]			
operation:		high and low half	byte swap, the save result into R	
period:	1	0		
affected flag	none			
bit: oxomplo:				
example:		025	· load 25H to ACC	
	LDIA LD	035H R01 A	; load 35H to ACC	
		R01,A	; load ACC to R01, R01=35H	

SWAPR

R01

;operation result: R01=53H



SZB operation: period: affected flag bit: example:	<b>[R],b</b> Determine the bit b of register R, if it is 0 then jump, otherwise execute in sequence 1 or 2 none		
·	SZB	R01,3	;determine 3 <sup>rd</sup> bit of R01
	JP	LOOP	;if is 1, execute, jump to LOOP
	JP	LOOP1	; if is 0, jump,execute, jump to LOOP1
SNZB	[R],b		
operation:		ne bit b of register R,	if it is 1 then jump, otherwise execute in sequence
period:	1 or 2		
affected flag bit:	none		
example:			
	SNZB	R01,3	; determine 3 <sup>rd</sup> bit of R01
	JP	LOOP	; if is 0, execute, jump to LOOP
	JP	LOOP1	; if is 1, jump,execute, jump to LOOP1
SZA	[R]		
operation:	Load the val	ue of R to ACC, if it is	s 0 then jump, otherwise execute in sequence
period:	1 or 2		
affected flag bit:	none		
example:			
	SZA	R01	;R01→ACC
	JP	LOOP	;if R01 is not 0, execute,jump to LOOP
	JP	LOOP1	;if R01 is 0, jump, execute, jump to LOOP1
SZR	[R]		
operation:	Load the val	ue of R to R, if it is 0	then jump, otherwise execute in sequence
period:	1 or 2		
affected flag bit:	None		
example:			

SZR	R01	;R01→R01
JP	LOOP	; if R01 is not 0, execute,jump to LOOP
JP	LOOP1	; if R01 is 0, jump,execute,jump to LOOP1



#### SZINCA [R]

SZINCA	[R]		
operation:	Increment r	egister by 1, save the	e result to ACC, if it is 0 then jump, otherwise execute in sequence
period:	1 or 2		
affected flag bit:	none		
example:			
	SZINCA	R01	;R01+1→ACC
	JP	LOOP	; if ACC is not 0, execute,jump to LOOP
	JP	LOOP1	; if ACC is 0, jump, execute, jump to LOOP1
SZINCR	[R]		
operation:	Increment r	egister by 1, save the	e result to R,if it is 0 then jump, otherwise execute in sequence
period:	1 or 2		
affected flag bit:	none		
example:			
·	SZINCR	R01	;R01+1→R01
	JP	LOOP	; if R01 is not 0, execute,jump to LOOP
	JP	LOOP1	; if R01 is 0, jump, execute, jump to LOOP1
SZDECA	[R]		
SZDECA operation:		register by 1,save th	ne result to ACC, if it is 0 then jump, otherwise execute in sequence
		register by 1,save th	ne result to ACC, $$ if it is 0 then jump, otherwise execute in sequence
operation: period: affected flag	decrement 1 or 2	register by 1,save th	ne result to ACC, if it is 0 then jump, otherwise execute in sequence
operation: period: affected flag bit:	decrement	register by 1,save th	ne result to ACC, if it is 0 then jump, otherwise execute in sequence
operation: period: affected flag	decrement 1 or 2 none		
operation: period: affected flag bit:	decrement 1 or 2 none SZDECA	R01	;R01-1→ACC
operation: period: affected flag bit:	decrement 1 or 2 none SZDECA JP	R01 LOOP	;R01-1→ACC ; if ACC is not 0, execute,jump to LOOP
operation: period: affected flag bit:	decrement 1 or 2 none SZDECA	R01	;R01-1→ACC
operation: period: affected flag bit: example:	decrement 1 or 2 none SZDECA JP JP	R01 LOOP	;R01-1→ACC ; if ACC is not 0, execute,jump to LOOP
operation: period: affected flag bit: example: SZDECR	decrement 1 or 2 none SZDECA JP JP [ <b>R</b> ]	R01 LOOP LOOP1	;R01-1→ACC ; if ACC is not 0, execute, jump to LOOP ; if ACC is 0, jump, execute, jump to LOOP1
operation: period: affected flag bit: example: <b>SZDECR</b> operation:	decrement 1 or 2 none SZDECA JP JP <b>[R]</b> Decrement	R01 LOOP LOOP1	;R01-1→ACC ; if ACC is not 0, execute,jump to LOOP
operation: period: affected flag bit: example: <b>SZDECR</b> operation: period:	decrement 1 or 2 none SZDECA JP JP <b>[R]</b> Decrement 1 or 2	R01 LOOP LOOP1	;R01-1→ACC ; if ACC is not 0, execute, jump to LOOP ; if ACC is 0, jump, execute, jump to LOOP1
operation: period: affected flag bit: example: <b>SZDECR</b> operation:	decrement 1 or 2 none SZDECA JP JP <b>[R]</b> Decrement	R01 LOOP LOOP1	;R01-1→ACC ; if ACC is not 0, execute, jump to LOOP ; if ACC is 0, jump, execute, jump to LOOP1
operation: period: affected flag bit: example: <b>SZDECR</b> operation: period: affected flag	decrement 1 or 2 none SZDECA JP JP <b>[R]</b> Decrement 1 or 2	R01 LOOP LOOP1	;R01-1→ACC ; if ACC is not 0, execute, jump to LOOP ; if ACC is 0, jump, execute, jump to LOOP1
operation: period: affected flag bit: example: <b>SZDECR</b> operation: period: affected flag bit:	decrement 1 or 2 none SZDECA JP JP <b>[R]</b> Decrement 1 or 2	R01 LOOP LOOP1	;R01-1→ACC ; if ACC is not 0, execute, jump to LOOP ; if ACC is 0, jump, execute, jump to LOOP1
operation: period: affected flag bit: example: <b>SZDECR</b> operation: period: affected flag bit:	decrement i 1 or 2 none SZDECA JP JP <b>[R]</b> Decrement 1 or 2 none	R01 LOOP LOOP1 register by 1, save th	;R01-1→ACC ; if ACC is not 0, execute, jump to LOOP ; if ACC is 0, jump, execute, jump to LOOP1 the result to R, if it is 0 then jump, otherwise execute in sequence
operation: period: affected flag bit: example: <b>SZDECR</b> operation: period: affected flag bit:	decrement i 1 or 2 none SZDECA JP JP <b>[R]</b> Decrement 1 or 2 none SZDECR	R01 LOOP LOOP1 register by 1, save th	;R01-1→ACC ; if ACC is not 0, execute, jump to LOOP ; if ACC is 0, jump, execute, jump to LOOP1 the result to R, if it is 0 then jump, otherwise execute in sequence ;R01-1→R01



[R]

#### TABLE

operation: Look-up table, the lower 8 bits of the look-up table result are placed in R, and the high bits are placed in the dedicated register TABLE\_DATAH

period:	2		
affected flag bit: example:	none		
	LDIA	01H	;load 01H to ACC
	LD	TABLE_SPH,A	;load ACC to higher bits of table address,TABLE_SPH=1
	LDIA	015H	;load 15H to ACC
	LD	TABLE_SPL,A	; load ACC to lower bits of table address, TABLE_SPL=15H
	TABLE	R01	;look-up table 0115H address , operation result : TABLE_DATAH=12H, R01=34H
	ORG	0115H	
	DW	1234H	

#### TABLEA

operation: period:		Look-up table, the lower 8 bits of the look-up table result are placed in ACC, and the high bits are placed in the dedicated register TABLE_DATAH 2		
affected flag bit:	none			
example:				
	LDIA	01H	; load 01H to ACC	
	LD	TABLE_SPH,A	; load ACC to higher bits of table address, TABLE_SPH=1	
	LDIA	015H	; load 15H to ACC	
	LD	TABLE_SPL,A	; load ACC to lower bits of table address,TABLE_SPL=15H	
	TABLEA		;look-up table 0115H address , operation result :	
			TABLE_DATAH=12H, ACC=34H	
	ORG	0115H		
	DW	1234H		
TESTZ	[R]			
operation:	Pass the R	to R, as affected Z flag	g bit	
period:	1			
affected flag bit:	Z			
example:				
	TESTZ	R0	;	
	SZB	STATUS,Z	;check Z flag bit,if it is 0 then jump	
	JP	Add1	;if R0 is 0, jump to address Add1	
	JP	Add2	;if R0 is not 0, jump to address Add2	

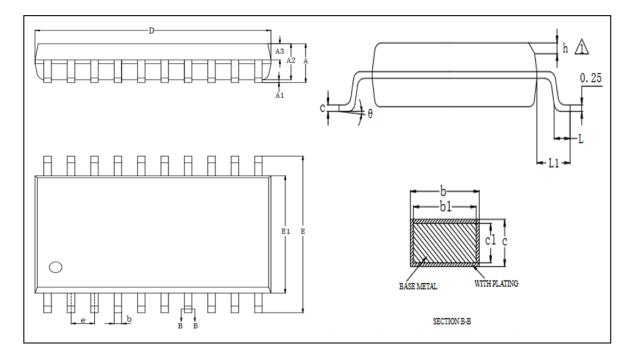


XORIA operation: period: affected flag bit: example:	1 Z		save the result to ACC
	LDIA	0AH	;load 0AH to ACC
	XORIA	0FH	;execute: ACC=05H
XORA	[R]		
operation:		R' on I and ACC,	save the result to ACC
period:	1	·	
affected flag bit:	Z		
example:			
	LDIA	0AH	; load 0AH to ACC
	LD	R01,A	;load ACC to R01,R01=0AH
	LDIA	0FH	;load 0FH to ACC
	XORA	R01	;execute: ACC=05H
XORR	[R]		
operation:		P' on Land ACC	save the result to R
period:	1	IN OIT AND ACC,	
affected flag bit:	Z		
example:			
	LDIA	0AH	; load 0AH to ACC
	LD	R01,A	; load ACC to R01,R01=0AH
	LDIA	0FH	; load 0FH to ACC
	XORR	R01	;execute: R01=05H



## 22. Packaging

### 22.1 SOP20

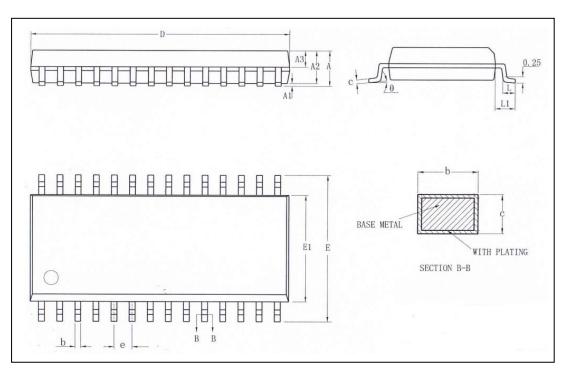


Currente el	Millimeter		
Symbol	Min	Nom	Max
A	-	-	2.65
A1	0.10	-	0.30
A2	2.24	-	2.44
A3	0.97	1.02	1.07
b	0.39	-	0.47
b1	0.38	0.41	0.44
с	0.25	-	0.30
c1	0.24	0.25	0.26
D	12.65	-	12.90
E	10.10	10.30	10.50
E1	7.40	7.50	7.60
е	1.27BSC		
h		0.50REF	
L	0.70	-	1.01
L1		1.40REF	
θ	0	-	8°

Caution: Package dimensions do not include mold flash or gate burrs.



### 22.2 SOP28



Sympol	Millimeter		
Symbol	Min	Nom	Max
A	-	-	2.65
A1	0.10	-	0.30
A2	2.24	-	2.44
A3	0.97	1.02	1.07
b	0.39	-	0.47
с	0.25	-	0.30
D	17.90	18.00	18.10
E	10.10	10.30	10.50
E1	7.40	7.50	7.60
е	1.27BSC		
L	0.70	-	1.01
L1		1.40REF	
θ	0	-	8°

Caution: Package dimensions do not include mold flash or gate burrs.



# 23. Version Revision

Version number	Date	Revised content
V1.0	Aug,2019	Original version
V1.1	May,2020	1.fix the description error in PORTC procedure 2.add 16MHz convertor clock illustration in ADC chapter
V1.2	June,2020	Add 32.768KHz crystal oscillator description
V1.3	June,2020	Update program EEPROM read/write example
V1.4	Mar,2022	<ol> <li>Corrected that RC0IF and TX0IF in PIR1 register are read- only and RC1IF and TX1IF in PIR1 register are read-only</li> <li>Corrected TX0IF description in PIR1 register and TX1IF description in PIR2 register</li> <li>Figs 14-3 and 14-4 corrected for USART asynchronous sending</li> <li>Corrected the sequence number of MSSP chapters</li> </ol>
V1.5.0	May,2022	<ol> <li>Corrected incorrect description of I2CTM master mode receive timing diagram</li> <li>Corrected 14.1.2.3 Receive Interrupt description</li> <li>Corrected FERR frame error bits in RCSTA register are read-only</li> <li>Add clock block diagram</li> <li>Revised the internal high-speed oscillation frequency to F<sub>HSI</sub>, and revised the clock sources of other modules according to the clock block diagram</li> <li>Revised sleep wake-up waiting time and WDT reset time</li> <li>Revised reset value of LED DATA register</li> <li>Delete description of sleep wake-up in ADC interrupt</li> <li>Modify the description of ANS15</li> <li>Added content in Section 20.6 IIC Electrical Characteristics</li> </ol>
V1.5.1	Aug,2024	1) Modified SOP20/SOP28 package dimensions
V1.5.2	Sep 2024	<ul> <li>2) Corrected the cover page</li> <li>1) Revised the cover page</li> <li>2) Added remarks to section 20.5</li> </ul>